CC253x System-on-Chip Solution for 2.4-GHz IEEE 802.15.4 and ZigBee® Applications

CC2540/41 System-on-Chip Solution for 2.4-GHz Bluetooth® low energy Applications

User's Guide



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Contents

Prefa	ce		15
1	Introc	duction	19
	1.1	Overview	20
		1.1.1 CPU and Memory	23
		1.1.2 Clocks and Power Management	23
		1.1.3 Peripherals	23
		1.1.4 Radio	25
	1.2	Applications	25
2	8051	СРИ	27
	2.1	8051 CPU Introduction	28
	2.2	Memory	28
		2.2.1 Memory Map	28
		2.2.2 CPU Memory Space	30
		2.2.3 Physical Memory	30
		2.2.4 XDATA Memory Access	
		,	36
	2.3	5	37
		2.3.1 Data Pointers	
		2.3.2 Registers R0–R7	
		2.3.3 Program Status Word	
		2.3.4 Accumulator	
		2.3.5 B Register	
	~ (2.3.6 Stack Pointer	
	2.4	Instruction Set Summary	
	2.5	Interrupts	
			43
			47
_			49
3		g Interface	
	3.1	Debug Mode	
	3.2	Debug Communication	
	3.3	Debug Commands	
		3.3.1 Debug Configuration	
		3.3.2 Debug Status	
	~ .	3.3.3 Hardware Breakpoints	
	3.4		60
	0.5	3.4.1 Lock Bits	
	3.5		60 60
	3.6	Registers	
4		5	63
	4.1		64
		4.1.1 Active and Idle Mode	
		4.1.2 PM1	
		4.1.3 PM2	65

		4.1.4 PM3	65
	4.2	Power-Management Control	65
	4.3	Power-Management Registers	66
	4.4	Oscillators and Clocks	69
		4.4.1 Oscillators	69
		4.4.2 System Clock	69
		4.4.3 32 kHz Oscillators	70
		4.4.4 Oscillator and Clock Registers	70
	4.5	Timer Tick Generation	72
	4.6	Data Retention	72
5	Reset		73
•	5.1	Power-On Reset and Brownout Detector	
	5.2	Clock-Loss Detector	
c			
6		Controller	
	6.1	Flash Memory Organization	
	6.2	Flash Write	
		6.2.1 Flash-Write Procedure	
		6.2.2 Writing Multiple Times to a Word	
		6.2.3 DMA Flash Write	
		6.2.4 CPU Flash Write	
	6.3	Flash Page Erase	
		6.3.1 Performing Flash Erase From Flash Memory	
	0 4	6.3.2 Different Flash Page Size on CC2533	
	6.4	Flash DMA Trigger	
	6.5	Flash Controller Registers	
7		orts	
	7.1	Unused I/O Pins	
	7.2	Low I/O Supply Voltage	
	7.3	General-Purpose I/O	
	7.4	General-Purpose I/O Interrupts	
	7.5	General-Purpose I/O DMA	
	7.6	Peripheral I/O	
		7.6.1 Timer 1	-
		7.6.2 Timer 3	84
		7.6.3 Timer 4	
		7.6.4 USART 0	85
		7.6.5 USART 1	85
		7.6.6 ADC	86
		7.6.7 Operational Amplifier and Analog Comparator	86
	7.7	Debug Interface	
	7.8	32-kHz XOSC Input	86
	7.9	Radio Test Output Signals	86
	7.10	Power-Down Signal MUX (PMUX)	86
	7.11	I/O Registers	87
8	DMA	Controller	95
-	8.1	DMA Operation	
	8.2	DMA Configuration Parameters	
		8.2.1 Source Address	
		8.2.2 Destination Address	
		8.2.3 Transfer Count	
		8.2.4 VLEN Setting	
		8.2.5 Trigger Event	
			~~~

www.t	ti.com		
		8.2.6 Source and Destination Increment	. 99
		8.2.7 DMA Transfer Mode	
		8.2.8 DMA Priority	100
		8.2.9 Byte or Word Transfers	
		8.2.10 Interrupt Mask	100
		8.2.11 Mode 8 Setting	100
	8.3	DMA Configuration Setup	100
	8.4	Stopping DMA Transfers	101
	8.5	DMA Interrupts	101
	8.6	DMA Configuration Data Structure	101
	8.7	DMA Memory Access	101
	8.8	DMA Registers	104
9	Timer	⁻ 1 (16-Bit Timer)	107
	9.1	16-Bit Counter	108
	9.2	Timer 1 Operation	
	9.3	Free-Running Mode	
	9.4	Modulo Mode	
	9.5	Up/Down Mode	
	9.6	Channel-Mode Control	
	9.7	Input Capture Mode	
	9.8	Output Compare Mode	
	9.9	IR Signal Generation and Learning	
		9.9.1 Introduction	
		9.9.2 Modulated Codes	
		9.9.3 Non-Modulated Codes	
		9.9.4 Learning	
		9.9.5 Other Considerations	
	9.10	Timer 1 Interrupts	117
	9.11	Timer 1 DMA Triggers	
	9.12	Timer 1 Registers	
	9.13	Accessing Timer 1 Registers as Array	
10	Timer	3 and Timer 4 (8-Bit Timers)	
	10.1	8-Bit Timer Counter	
	10.1	Timer 3/Timer 4 Mode Control	
	10.2	10.2.1 Free-Running Mode	126
		10.2.2 Down Mode	
		10.2.3 Modulo Mode	
		10.2.4 Up/Down Mode	
	10.3	Channel Mode Control	
	10.4	Input Capture Mode	
	10.1	Output Compare Mode	
	10.6	Timer 3 and Timer 4 Interrupts	
	10.7	Timer 3 and Timer 4 DMA Triggers	
	10.8	Timer 3 and Timer 4 Registers	
11		u de la construcción de la const	133
	11.1		
	11.1	General Timer Compare	
	11.2	Timer Compare	
	11.3 11.4		
		Sleep Timer Registers	
12	ADC		
	12.1	ADC Introduction	
	12.2	ADC Operation	138



www	tı.	com

12.2.2       ADC Conversion Sequences       139         12.2.3       Single ADC Conversion       139         12.2.4       ADC Operating Modes       139         12.2.5       ADC Conversion Results       140         12.2.6       ADC Cereference Voltage       140         12.2.7       ADC Conversion Timing       140         12.2.8       ADC Interrupts       140         12.2.9       ADC DMA Triggers       140         12.2.10       ADC Registers       141         13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147         14.1       Introdom-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         15.4       AES Operation       154         15.5       Mode Input Data       154         15.6       AES Operation       154
12.2.4       ADC Operating Modes       139         12.2.5       ADC Conversion Results       140         12.2.6       ADC Conversion Timing       140         12.2.7       ADC Conversion Timing       140         12.2.7       ADC Conversion Timing       140         12.2.8       ADC Interrupts       140         12.2.9       ADC DMA Triggers       140         12.2.10       ADC Registers       141         13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.2       Battery Monitor Registers       147         14       Random-Number Generator Operation       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.3       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Coperation       154         15.2       Key and IV       154         15.3       Pading of Input Data       15
12.2.5       ADC Conversion Results       140         12.2.6       ADC Reference Voltage       140         12.2.7       ADC Conversion Timing       140         12.2.8       ADC Interrupts       140         12.2.9       ADC MA Triggers       140         12.2.10       ADC Registers       141         13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147         14.       Random-Number Generator       149         14.1       Introduction       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Rendom-Number-Generator Registers       151         15.1       AES Coprocessor       153         15.1       AES Coprocessor       153         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       AES Operation       154         15.5       AES Legisters       157         15.6       CBC-MAC       154         15.7
12.2.6       ADC Reference Voltage       140         12.2.7       ADC Conversion Timing       140         12.2.8       ADC Interrupts       140         12.2.9       ADC DMA Triggers       140         12.2.10       ADC Registers       140         12.2.10       ADC Registers       141         13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147         14       Random-Number Generator       149         14.1       Introduction       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.2.3       CRC16       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         15.4       AES Coprocessor       153         15.5       Key and IV       154         15.4       Key and IV       154         15.5       Modes Of Operation       154         15.6       CBC-MAC       155
12.2.7 ADC Conversion Timing       140         12.2.8 ADC Interrupts       140         12.2.9 ADC DMA Triggers       140         12.2.10 ADC Registers       141         13 Battery Monitor       145         13.1 Functionality and Usage of the Battery Monitor       146         13.2 Using the Battery Monitor for Temperature Monitoring       146         13.3 Battery Monitor Registers       147 <b>Random-Number Generator</b> 149         14.1 Introduction       150         14.2.1 Seadorandom Sequence Generation       150         14.2.2 Seeding       150         14.2.3 CRC16       150         14.3 Random-Number-Generator Registers       151         15 <b>AES Coprocessor</b> 153         15.1 AES Operation       154         15.2 Key and IV       154         15.3 Padding of Input Data       154         15.4 CBC-MAC       154         15.5 AES Interrupts       157         15.6 CBC-MAC       154         15.7 CCM Mode       155         15.8 AES Interrupts       157         15.9 AES DMA Triggers       157         15.0 AES Registers       157         15.1 Mates Operation       158
12.2.8       ADC Interrupts       140         12.2.10       ADC Registers       141         13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147         14       Random-Number Generator       149         14.1       Introduction       150         14.2       Seatory Monitor Sequence Generation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         15.4       AES Coprocessor       153         15.1       AES Coprocessor       153         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       AES Interrupts       154         15.5       Modes of Operation       154         15.6       CBC-MAC       155         15.8       AES Interrupts       157         15.9       AES Inte
12.2.9 ADC DMA Triggers       140         12.2.10 ADC Registers       141         13 Battery Monitor       145         13.1 Functionality and Usage of the Battery Monitor       146         13.2 Using the Battery Monitor for Temperature Monitoring       146         13.3 Battery Monitor Registers       147         13.3 Battery Monitor Registers       147         14.1 Introduction       149         14.1 Introduction       150         14.2.1 Pseudorandom Sequence Generation       150         14.2.2 Seeding       150         14.2.3 CRC16       150         14.3 Random-Number-Generator Registers       151         15       ABS Coprocessor       153         15.1 AES Operation       154         15.2 Key and IV       154         15.4 Interface to CPU       154         15.5 Mode of Operation       154         15.6 GBC-MAC       155         15.8 AES Interrupts       157         15.9 AES DMA Triggers       157         15.1 Watchdog Mode       160         16.1 Watchdog Mode       160         16.2 Timer Mode       160         16.3 Watchdog Timer       169         16.1 Watchdog Mode       160
12.2.10 ADC Registers       141         13 Battery Monitor       145         13.1 Functionality and Usage of the Battery Monitor       146         13.2 Using the Battery Monitor for Temperature Monitoring       146         13.3 Battery Monitor Registers       147         14       Random-Number Generator       149         14.1 Introduction       150         14.2 Random-Number-Generator Operation       150         14.2.1 Pseudorandom Sequence Generation       150         14.2.2 Seeding       150         14.2.3 CRC16       150         14.3 Random-Number-Generator Registers       151 <b>AES Coprocessor</b> 153         15.1 AES Operation       154         15.2 Key and IV       154         15.3 Padding of Input Data       154         15.4 Mode       154         15.5 Modes of Operation       154         15.6 CBC-MAC       154         15.7 CCM Mode       155         15.8 AES Interrupts       157         15.9 AES DMA Triggers       157         15.10 AES Registers       157         15.10 AES Registers       157         15.10 AES Registers       157         15.10 AES Registers       157      <
13       Battery Monitor       145         13.1       Functionality and Usage of the Battery Monitor       146         13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147         14       Random-Number Generator       149         14.1       Introduction       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.2.4       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Itse Key and IV       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES MAA Triggers       157         15.10       AES Registers
13.1       Functionality and Usage of the Battery Monitor
13.2       Using the Battery Monitor for Temperature Monitoring       146         13.3       Battery Monitor Registers       147 <b>Random-Number Generator</b> 149         14.1       Introduction       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.3.8       Random-Number-Generator Registers       151 <b>AES Coprocessor</b> 153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       Mode of Operation       154         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.0       AES Registers       157         15.0       AES Register       160         16.1       Watchdog Timer Register       163 <t< th=""></t<>
13.3       Battery Monitor Registers       147         14       Random-Number Generator       149         14.1       Introduction       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.2.3       CRC16       150         14.3.8       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.9       AES BOMA Triggers       157         15.0       AES Registers       157         16       Watchdog Mode       160         16.1       Watchdo
14       Random-Number Generator       149         14.1       Introduction       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.2.4       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         16.1       Watchdog Timer       159         16.1       Watchdog Timer Register       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       161         17.1
14.1       Introduction       150         14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         14.3       Random-Number-Generator Registers       151         15.1       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes       155         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES Interrupts       157         15.0       AES Registers       157         16       Watchdog Mode       160         16.2       Timer Mode       160         16.2       Timer Mode       160         16.1       Watchdog Timer       163         17.1       UART Mode
14.2       Random-Number-Generator Operation       150         14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.0       AES Registers       157         15.10       AES Registers       157         15.4       Watchdog Timer       159         16.1       Watchdog Timer Register       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.4       17.1.1
14.2.1       Pseudorandom Sequence Generation       150         14.2.2       Seeding       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151 <b>15 AES Coprocessor</b> 153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.0       AES Registers       157         15.0       AES Registers       157         15.10       AES Registers       157         16.1       Watchdog Timer       159         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         16.4       17.1.1       UART Transmit
14.2.2       Seeding       150         14.2.3       CRC16       150         14.3       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES Registers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17       USART       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164
14.2.3 CRC16       150         14.3 Random-Number-Generator Registers       151         15 AES Coprocessor       153         15.1 AES Operation       154         15.2 Key and IV       154         15.3 Padding of Input Data       154         15.4 Interface to CPU       154         15.5 Modes of Operation       154         15.6 CBC-MAC       154         15.7 CCM Mode       154         15.8 AES Interrupts       157         15.9 AES DMA Triggers       157         15.10 AES Registers       157         16       Watchdog Timer       159         16.1 Watchdog Mode       160         16.2 Timer Mode       160         16.3 Watchdog Timer Register       160         16.3 Watchdog Timer Register       160         16.4 UART Mode       160         16.7 ULART Mode       164         17.1 UART Mode       164         17.1.2 UART Receive       164         17.1.3 UART Hardware Flow Control       164         17.1.4 UART Character Format       165         17.2 SPI Master Operation       165         17.2.4 SPI Master Operation       165         17.2.5 SPI Slave Operation       165
14.3       Random-Number-Generator Registers       151         15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       154         15.8       AES Interrupts       157         15.9       AES Interrupts       157         15.0       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Transmit       164         17.1.3       UART Thardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation
15       AES Coprocessor       153         15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         16.1       Watchdog Timer       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Transmit       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.1       AES Operation       154         15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       1557         15.9       AES DMA Triggers       157         15.0       AES Registers       157         16.1       Watchdog Timer       159         16.1       Watchdog Timer Register       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Transmit       164         17.1.1       UART Transmit       164         17.1.2       UART Hardware Flow Control       164         17.1.3       UART Hardware Flow Control       164         17.2.1       SPI Master Operation       165         17.2.2       SPI Master Operation       165         17.2.1       SPI Master Operation       165         17.2.2       SPI S
15.2       Key and IV       154         15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       155         15.9       AES DMA Triggers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.3       Padding of Input Data       154         15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.2       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2       SPI Mode       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165
15.4       Interface to CPU       154         15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17.1       UART Mode       163         17.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.5       Modes of Operation       154         15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
15.6       CBC-MAC       154         15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2       SPI Mode       165         17.2.2       SPI Slave Operation       165
15.7       CCM Mode       155         15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.0       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.8       AES Interrupts       157         15.9       AES DMA Triggers       157         15.10       AES Registers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.4       17.1       UART Mode       163         17.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.9       AES DMA Triggers       157         15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
15.10       AES Registers       157         16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
16       Watchdog Timer       159         16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.4       T.1.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
16.1       Watchdog Mode       160         16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       165
16.2       Timer Mode       160         16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Receive       164         17.1.5       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
16.3       Watchdog Timer Register       160         17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Receive       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17       USART       163         17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1       UART Mode       164         17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Receive       164         17.1.4       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1.1       UART Transmit       164         17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1.2       UART Receive       164         17.1.3       UART Hardware Flow Control       164         17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.1.4       UART Character Format       165         17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.2       SPI Mode       165         17.2.1       SPI Master Operation       165         17.2.2       SPI Slave Operation       166
17.2.2 SPI Slave Operation
·
·
17.3 SSN Slave-Select Pin
17.4 Baud-Rate Generation
17.5 USART Flushing
17.6 USART Interrupts
17.7 USART DMA Triggers

www.t	i.com			
	18.1	Descrip	tion	174
	18.2	•	ion	
	18.3		ource	
	18.4		rs	
19	Analo	•	parator	
15	19.1	•	tion	
	19.2		r	
00	-	0		
20				
	20.1		on I ² C Initialization and Reset	
			I ² C Serial Data	
			I ² C Addressing Modes	
			I ² C Module Operating Modes	
			I ² C Clock Generation and Synchronization	
			Bus Error	
			I ² C Interrupt	
			I ² C Pins	
	20.2		isters	
24	-	0		
21			ler	
	21.1 21.2		roduction	
	21.2		able USB PLL	
	21.3		errupts	
	21.4		enupis	
	21.6		nt-O Interrupts	
	21.0	21.6.1	Error Conditions	
		21.6.2		
		21.6.3	IN Transactions (TX State)	
			OUT Transactions (RX State)	
	21.7		nts 1–5	
		21.7.1	FIFO Management	195
		21.7.2	Double Buffering	
		21.7.3	FIFO Access	
		21.7.4	Endpoint 1–5 Interupts	197
		21.7.5	Bulk/Interrupt IN Endpoint	198
		21.7.6	Isochronous IN Endpoint	198
		21.7.7	Bulk/Interrupt OUT Endpoint	198
		21.7.8	Isochronous OUT Endpoint	198
	21.8	DMA		199
	21.9	USB Re	eset	199
	21.10	Suspen	d and Resume	199
	21.11		Wake-Up	
	21.12	USB Re	egisters	200
22	Timer	⁻ 2 (MAC	C Timer)	<b>207</b>
	22.1		Dperation	
		22.1.1	General	208
		22.1.2	Up Counter	208
		22.1.3	Timer Overflow	208
		22.1.4	Timer Delta Increment	
		22.1.5	Timer Compare	
			Overflow Count	
		22.1.7	Overflow-Count Update	209

		22.1.8 Overflow-Count Overflow	209
		22.1.9 Overflow-Count Compare	
		22.1.10 Capture Input	
		22.1.10 Captule input	
	22.2	• · · · ·	
		Interrupts Event Outputs (DMA Trigger and Radio Events)	
	22.3		
	22.4	Timer Start/Stop Synchronization	
		22.4.1 General	
		22.4.2 Timer Synchronous Stop	
		22.4.3 Timer Synchronous Start	
	22.5	Timer 2 Registers	212
23	CC25	3x Radio	<b>219</b>
	23.1	RF Core	220
		23.1.1 Interrupts	220
		23.1.2 Interrupt Registers	220
	23.2	FIFO Access	224
	23.3	DMA	224
	23.4	Memory Map	224
		23.4.1 RXFIFO	
		23.4.2 TXFIFO	
		23.4.3 Frame-Filtering and Source-Matching Memory Map	
	23.5	Frequency and Channel Programming	
	23.6	IEEE 802.15.4-2006 Modulation Format	
	23.7	IEEE 802.15.4-2006 Frame Format	
	23.7	23.7.1 PHY Layer	
		23.7.2 MAC Layer	
	<u></u>	•	
	23.8	Transmit Mode	
		23.8.1 TX Control	
		23.8.2 TX State Timing	
		23.8.3 TXFIFO Access	
		23.8.4 Retransmission	
		23.8.5 Error Conditions	
		23.8.6 TX Flow Diagram	
		23.8.7 Transmitted Frame Processing	
		23.8.8 Synchronization Header	232
		23.8.9 Frame-Length Field	232
		23.8.10 Frame Check Sequence	232
		23.8.11 Interrupts	233
		23.8.12 Clear-Channel Assessment	233
		23.8.13 Output Power Programming	233
		23.8.14 Tips and Tricks	
	23.9	Receive Mode	
	_0.0	23.9.1 RX Control	
		23.9.2 RX State Timing	
		23.9.3 Received Frame Processing	
		23.9.4 Synchronization Header and Frame-Length Fields	
		23.9.5 Frame Filtering	
		23.9.6 Source Address Matching	
		23.9.7 Frame-Check Sequence	
	00.40	23.9.8 Acknowledgement Transmission	
	23.10	RXFIFO Access	
		23.10.1 Using the FIFO and FIFOP	
		23.10.2 Error Conditions	244

www.ti	.com			
		23.10.3	RSSI	244
			Link Quality Indication	
	23.11		ontrol State Machine	
	23.12		n-Number Generation	
	23.13		Sniffing and Radio Test Output Signals	
	23.14		nd Strobe/CSMA-CA Processor	
		23.14.1	Instruction Memory	
		23.14.2	Data Registers	
		23.14.3	Program Execution	250
		23.14.4	Interrupt Requests	
		23.14.5	Random Number Instruction	
		23.14.6	Running CSP Programs	
		23.14.7	Registers	
		23.14.8	Instruction Set Summary	
		23.14.9	Instruction Set Definition	
	23.15	Register	7S	
		23.15.1	Register Settings Update	
		23.15.2	Register Access Modes	
		23.15.3	Register Descriptions	
24	CC25			287
27	24.1			-
		-		
25		•	rietary Mode Radio	
	25.1			
	25.2	-	S	
			Interrupt Registers	
	25.3		Data Memory	
			FIFOs	
			DMA	
			RAM-Based Registers	
			Variables in RAM Page 5	
	25.4		am Processor	
			Whitening	
			CC2500-Compatible PN9 Whitening	
			CRC	
	05 F		Coprocessor Mode	308
	25.5		cy and Channel Programming	
	25.6		ion Formats	
	25.7		r	
	25.8		Format	
			Rx FIFO Packet Organization	
			Tx FIFO Packet Organization	
	25.0		Tx Buffers for ACK Payload	
	25.9	•	er Engine	
			Command Register	
			Radio Tasks	
	25 10		RF Test Commands	
	25.10		Number Generation	
	25.11		Sniffing	
	25.12	-	rs	
		25.12.1	Register Overview	
			Register Settings Update	
26	Volta	ge Regu	lator	355



27	Available Software				
	27.1	SmartRF™ Software for Evaluation (www.ti.com/smartrfstudio)	358		
	27.2	RemoTI™ Network Protocol (www.ti.com/remoti)	358		
	27.3	SimpliciTI™ Network Protocol (www.ti.com/simpliciti)	359		
	27.4 TIMAC Software (www.ti.com/timac)				
	27.5	Z-Stack [™] Software (www.ti.com/z-stack)	360		
	27.6	BLE Stack Software	360		
Α	Abbre	eviations	361		
В	Addit	ional Information	365		
	B.1	Texas Instruments Low-Power RF Web Site	366		
	B.2	Low-Power RF Online Community	366		
	B.3	Texas Instruments Low-Power RF Developer Network	366		
	B.4	Low-Power RF eNewsletter	366		
С	Refer	ences	367		
Revis	Revision History				



## List of Figures

1-1.	CC253x Block Diagram	
1-2.	CC2540 Block Diagram	
1-3.	CC2541 Block Diagram	22
2-1.	XDATA Memory Space (Showing SFR and DATA Mapping)	29
2-2.	CODE Memory Space	
2-3.	CODE Memory Space for Running Code From SRAM	29
2-4.	Interrupt Overview	
3-1.	External Debug Interface Timing	
3-2.	Transmission of One Byte	
3-3.	Typical Command Sequence—No Extra Wait for Response	
3-4.	Typical Command Sequence. Wait for Response	
3-5.	Burst Write Command (First 2 Bytes)	
4-1.	Clock System Overview	
6-1.	Flash Write Using DMA	
8-1.	DMA Operation	
8-2.	Variable Length (VLEN) Transfer Options	
9-1.	Free-Running Mode	108
9-2.	Modulo Mode	109
9-3.	Up/Down Mode	109
9-4.	Output Compare Modes, Timer Free-Running Mode	112
9-5.	Output Compare Modes, Timer Modulo Mode	113
9-6.	Output Compare Modes, Timer Up/Down Mode	114
9-7.	Block Diagram of Timers in IR-Generation Mode	116
9-8.	Modulated Waveform Example	116
9-9.	IR Learning Board Diagram	117
11-1.	Sleep Timer Capture (Example Using Rising Edge on P0_0)	135
12-1.	ADC Block Diagram	138
14-1.	Basic Structure of the Random-Number Generator	150
15-1.	Message Authentication Phase Block B0	155
15-2.	Authentication Flag Byte	155
15-3.	Message Encryption Phase Block	156
15-4.	Encryption Flag Byte	156
19-1.	Analog Comparator	176
20-1.	Block Diagram of the I ² C Module	178
20-2.	I ² C Bus Connection Diagram	179
20-3.	I ² C Module Data Transfer	179
20-4.	Bit Transfer on I ² C Bus	180
20-5.	I ² C Module 7-Bit Addressing Format	180
20-6.	I ² C Module Addressing Format With Repeated START Condition	180
20-7.	Arbitration Procedure Between Two Master Transmitters	186
20-8.	Synchronization of Two I ² C Clock Generators During Arbitration	187
21-1.	USB Controller Block Diagram	192
21-2.	IN/OUT FIFOs	196
23-1.	Modulation	227
23-2.	I/Q Phases When Transmitting a Zero-Symbol Chip Sequence, $t_c = 0.5 \ \mu s$	227
23-3.	Schematic View of the IEEE 802.15.4 Frame Format [1]	228
23-4.	Format of the Frame Control Field (FCF)	228



www	

23-5.	Frame Data Written to the TXFIFO	230
23-6.	TX Flow	231
23-7.	Single Transmitted Frame	232
23-8.	Transmitted Synchronization Header	232
23-9.	FCS Hardware Implementation	233
23-10.	Single Received Frame and Transmitted Acknowledgement Frame	234
23-11.	SFD Signal Timing	235
23-12.	Filtering Scenarios (Exceptions Generated During Reception)	237
23-13.	Matching Algorithm for Short and Extended Addresses	239
23-14.	Interrupts Generated by Source Address Matching	240
23-15.	Data in RXFIFO for Different Settings	241
23-16.	Acknowledge Frame Format	241
23-17.	Acknowledgment Timing	242
23-18.	Command Strobe Timing	242
23-19.	Behavior of FIFO and FIFOP Signals	244
23-20.	Main FSM	246
23-21.	FFT of the Random Bytes	247
23-22.	Histogram of 20 Million Bytes Generated With the RANDOM Instruction	247
23-23.	Running a CSP Program	251
23-24.	Example Hardware Structure for the R* Register Access Mode	267
25-1.	Mapping of Radio Memory to MCU XDATA Memory Space	294
25-2.	FIFO Pointers	294
25-3.	PN7 Whitening	305
25-4.	CC2500-Compatible Whitening	306
25-5.	CRC Module	307
25-6.	Air Interface Packet Format for Basic Mode	310
25-7.	Air Interface Packet Format for Auto Mode	311
25-8.	Bits of 9-Bit Header	311
25-9.	Bits of 10-Bit Header	311
25-10.	Structure of Packets in the Rx FIFO	312
25-11.	Structure of Packets in the Tx FIFO	313
25-12.	Timing of Packets in Rx Tasks	328
25-13.	Timing of Packets in Tx Tasks	329
25-14.	Complete Appended Packet	331



### List of Tables

0-1.	CC253x Family Overview	16
0-2.	Register Bit Conventions	17
2-1.	SFR Overview	31
2-2.	Overview of XREG Registers	35
2-3.	Instruction Set Summary	40
2-4.	Instructions That Affect Flag Settings	43
2-5.	Interrupts Overview	44
2-6.	Priority Level Setting	<b>50</b>
2-7.	Interrupt Priority Groups	50
2-8.	Interrupt Polling Sequence	51
3-1.	Debug Commands	56
3-2.	Debug Configuration	<b>58</b>
3-3.	Debug Status	58
3-4.	Relation Between PCON_IDLE and PM_ACTIVE	59
3-5.	Flash Lock-Protection Bit Structure Definition	<b>60</b>
4-1.	Power Modes	<b>64</b>
6-1.	Example Write Sequence	77
7-1.	Peripheral I/O Pin Mapping	83
8-1.	DMA Trigger Sources	101
8-2.	DMA Configuration Data Structure	102
9-1.	Initial Compare Output Values (Compare Mode)	111
9-2.	Frequency Error Calculation for 38-kHz Carrier	115
10-1.	Initial Compare Output Values (Compare Mode)	127
13-1.	Values Showing How Different Temperatures Relate to BATTMON_VOLTAGE for a Typical Device	146
13-2.	Values for A and B (for a Typical Device) When Using the Battery monitor for Temperature Monitoring	147
17-1.	Commonly Used Baud-Rate Settings for 32 MHz System Clock	166
20-1.	Slave Transmitter Mode	181
20-2.	Slave Receiver Mode	182
20-3.	Master Transmitter Mode	184
20-4.	Master Receiver Mode	185
20-5.	Miscellaneous States	187
20-6.	Clock Rates Defined at 32 MHz	188
21-1.	USB Interrupt Flags Interrupt-Enable Mask Registers	193
21-2.	FIFO Sizes for EP 1–5	196
22-1.	Internal Registers	213
23-1.	Frame Filtering and Source Matching Memory Map	225
23-2.	IEEE 802.15.4-2006 Symbol-to-Chip Mapping	227
23-3.	FSM State Mapping	247
23-4.	Instruction Set Summary	253
23-5.	Register Overview	266
23-6.	Registers That Require Update From Their Default Value	267
23-7.	Register-Bit Access Modes	267
25-1.	Radio RAM Pages	293
25-2.	Commands to FIFO via RFST Register	296
25-3.	Access to FIFO Registers	297
25-4.	RAM-Based Registers	298
25-5.	Address Structure for Auto Mode	302



25-6.	Address Structure for Basic Mode	303
25-7.	RAM-Based Registers in RAM Page 5	304
25-8.	Register Settings for Different CRCs	307
25-9.	Register Settings for Some Commonly Used CRCs, Assuming Initialization With All 1s	308
25-10.	Supported Modulation Formats, Data Rates, and Deviations	309
25-11.	Segments for Holding ACK Payload for Each Address Entry	313
25-12.	Commands From MCU to LL Engine via RFST Register	315
25-13.	Timer 2 Capture Settings	316
25-14.	End-of-Task Causes	317
25-15.	Recommended RAM Register Settings for Start Tone	319
25-16.	Interrupt and Counter Operation for Received Messages	320
25-17.	Interrupt and Counter Operation for Received Messages	321
25-18.	End-of-Receive Tasks	322
25-19.	Interrupt and Counter Operation for Received ACK Packets	325
25-20.	End-of-Transmit Tasks	326
25-21.	Additional Reasons for End-of-Transmit on Clear-Channel Tasks	327
25-22.	Packet-Sniffer Modes of Operation	331
25-23.	XREG Register Overview	332
25-24.	Registers That Should Be Updated From Their Default Value, Bit Rates 1 Mbps and Lower	333
25-25.	Registers That Should Be Updated From Their Default Value, Bit Rate 2 Mbps	333



## **Read This First**

#### **About This Manual**

The CC2540/41 is a cost-effective, low-power, and true system-on-chip (SoC) solution for Bluetooth low-energy applications. It enables robust BLE master or slave nodes to be built with very low total bill-of-material costs. The CC2540/41 combines the excellent performance of a leading RF transceiver with an industry-standard enhanced 8051 MCU, in-system programmable flash memory, 8-KB RAM, and many other powerful supporting features and peripherals. The CC2540/41 is suited for systems where very low power consumption is required. Very low-power sleep modes are available. Short transition times between operating modes further enable low power consumption.

The CC2540 comes in two different versions: CC2540F128 and CC2540F256, with 128 KB and 256 KB of flash memory, respectively.

The CC2541 comes in two different versions: CC2541F128 and CC2541F256, with 128 KB and 256 KB of flash memory, respectively.

The CC2541F128/F256 comes in two different versions: CC2541F128/F256, with 128 and 256 KB of flash memory, respectively.

Combined with the *Bluetooth* low-energy protocol stack from Texas Instruments, the CC2540F128/CC2540F256 and CC2541F128/CC2541F256 constitute the market's most comprehensive single-mode Bluetooth low energy solution.

The CC253x System-on-Chip solution for 2.4 GHz is suitable for a wide range of applications. These can easily be built on top of the IEEE 802.15.4 based standard protocols (RemoTI[™] network protocol, TIMAC software, and Z-Stack[™] software for ZigBee® compliant solutions) or on top of the proprietary SimpliciTI[™] network protocol. The usage is, however, not limited to these protocols alone. The CC253x family is, e.g. also suitable for 6LoWPAN and Wireless HART implementations.

Each chapter of this manual describes details of a module or peripheral; however, not all features are present on all devices. To see the differences regarding features, see Table 0-1 in the Devices section.

For detailed technical numbers, such as power consumption and RF performance, see the device-specific data sheet (Appendix C).

#### **Related Documentation and Software From Texas Instruments**

Related documentation (e.g., the CC2530 data sheet <u>http://www-s.ti.com/sc/techlit/swrs081</u> and CC2540 data sheet <u>http://www-s.ti.com/sc/techlit/swrs084</u>) can be found in Appendix C.

For more information regarding software that can be used with the CC253x/CC2540/41 System-on-Chip solution (e.g., SmartRF[™] software for radio performance and functionality evaluation), see Chapter 27, which also contains more information regarding the RemoTI network protocol, the SimpliciTI network protocol, the TIMAC software, the Z-Stack software, and the BLE stack software.

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#### **FCC Warning**

This equipment is intended for use in a laboratory test environment only. It generates, uses, and can radiate radio frequency energy and has not been tested for compliance with the limits of computing devices pursuant to subpart J of part 15 of FCC rules, which are designed to provide reasonable protection against radio frequency interference. Operation of this equipment in other environments may cause interference with radio communications, in which case the user at his own expense will be required to take whatever measures may be required to correct this interference.

#### If You Need Assistance

All technical support is channeled through the TI Product Information Centers (PIC) - <u>www.ti.com/support</u>. To send an E-mail request, please enter your contact information, along with your request at the following link – <u>PIC request form</u>.

Also visit the Low Power RF, ZigBee, and Bluetooth low-energy sections of the TI E2E Community (<u>www.ti.com/lprf-forum</u>), where you can easily get in touch with other CC253x and CC2540/41 users and find FAQs, Design Notes, Application Notes, Videos, etc.

You can also see the TI Knowledgebase for Analog & Mixed-Signal.

#### Glossary

Abbreviations used in this user guide can be found in Appendix A.

#### **Devices**

The CC253x System-on-Chip solution family consists of several devices. The following table provides a device overview and points out the differences regarding memory sizes and peripherals. For a complete feature list of any of the devices, see the corresponding data sheet (Appendix C).

Feature	CC2530F32/F64/ F128/F256	CC2531F128/F256	CC2533F32/F64/F96	CC2540F128/ F256	CC2541F128/ F256
FLASH_SIZE	32 KB/64 KB/128 KB/256 KB	128 KB/256 KB	32 KB/64 KB/96 KB	128 KB/256 KB	128 KB/256 KB
SRAM_SIZE	8 KB/8 KB/8 KB/8 KB	8 KB/8 KB	4 KB/4 KB/6 KB	8 KB	8 kB
USB Not included		Included	Not included	Included	Not included
ADC	Included	Included	Not included	Included	Included
Battery monitor	Not included	Not included	Included	Not included	Not included
l ² C	Not included	Not included	Included	Not included	Included
Operational Included		Included	Not included	Included	Not included
Analog comparator	Included	Included	Not included	Included	Included

Table 0-1. CC253x Family Overview

Legend:

FLASH_SIZE - The size of the flash

SRAM_SIZE - The size of the SRAM

#### **Register Conventions**

Each SFR and XREG register is described in a separate table, where each table title contains the following information in the format indicated:

For SFR registers: REGISTER NAME (SFR address) – register description

For XREG registers: REGISTER NAME (XDATA address) – register description

Each table has five columns to describe the different register fields as described in the following: Column 1 – Bit: Denotes which bits of the register are described/addressed in the specific row Column 2 - Name: Specific name of the register field

Column 3 - Reset: Reset/initial value of the register field

Column 4 – R/W: Key indicating the accessibility of the bits in the field (see Table 0-2 for more details)

Column 5 – Description: More details about the register field, and often a description of the functions of the different values

In the register descriptions, each register field is shown with a symbol (R/W) indicating the access mode of the register field. The register values are always given in binary notation unless prefixed by 0x, which indicates hexadecimal notation.

SYMBOL	ACCESS MODE
R/W	Read/write
R	Read-only
R0	Read as 0
R1	Read as 1
W	Write-only
W0	Write as 0
W1	Write as 1
H0	Hardware clear
H1	Hardware set

Table 0-2. Register Bit Conventions



## Introduction

As mentioned in the preface, the CC253x/CC2540/41 device family provides solutions for a wide range of applications. In order to help the user to develop these applications, this user's guide focuses on the usage of the different building blocks of the CC253x/CC2540/41 device family. For detailed device descriptions, complete feature lists, and performance numbers, see the device-specific data sheet (Appendix C).

In order to provide easy access to relevant information, the following subsections guide the reader to the different chapters in this guide.

Topic

Page

1.1	Overview	20
1.2	Applications	25



#### Overview

#### 1.1 Overview

The block diagrams in Figure 1-1 and Figure 1-2 show the different building blocks of the CC253x and CC2540/41 device family. Not all features and functions of all modules or peripherals are present on all devices of the CC253x/CC2540/41; hence, see the device-specific data sheet for a device-specific block diagram.

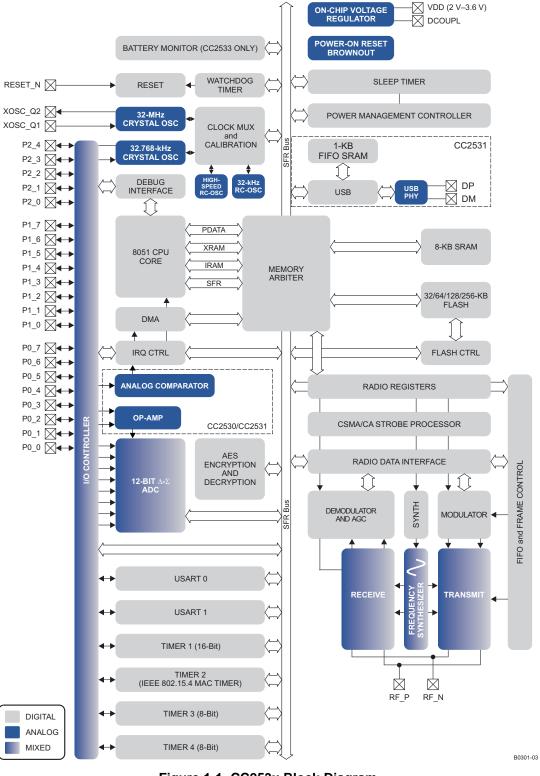


Figure 1-1. CC253x Block Diagram



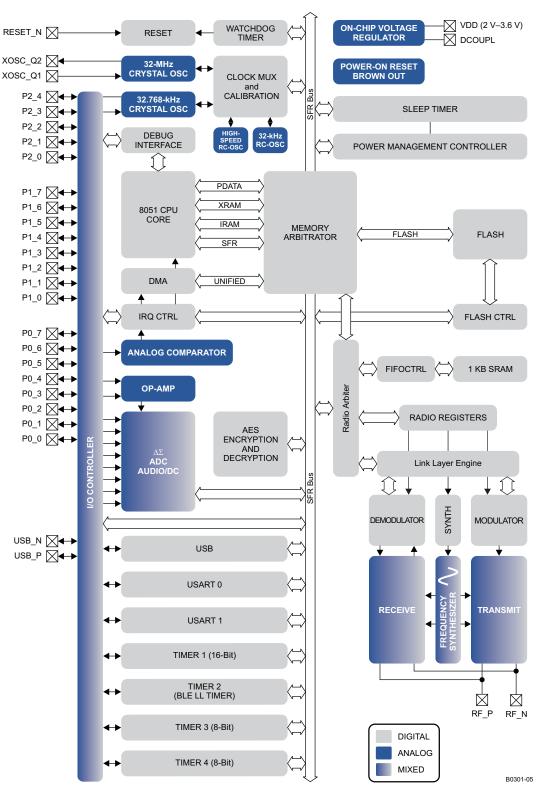
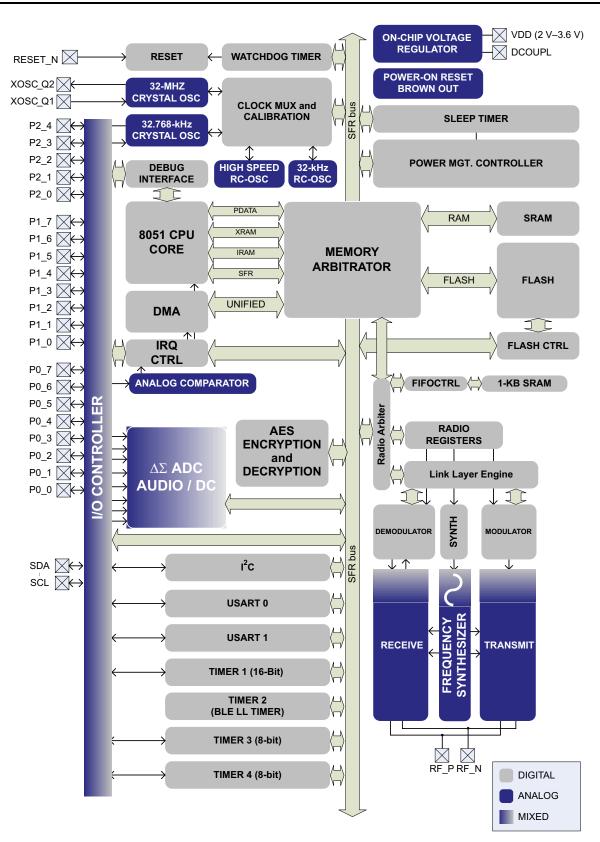


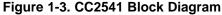
Figure 1-2. CC2540 Block Diagram

The modules can be roughly divided into one of three categories: CPU and memory related modules; modules related to peripherals, clocks, and power management; and radio-related modules.



Overview







#### 1.1.1 CPU and Memory

The **8051 CPU core** is a single-cycle 8051-compatible core . It has three different memory access buses (SFR, DATA, and CODE/XDATA) with single-cycle access to SFR, DATA, and the main SRAM. It also includes a debug interface and an 18-input extended interrupt unit. The detailed functionality of the CPU and the memory is addressed in Chapter 2.

The **interrupt controller** services a total of 18 interrupt sources, divided into six interrupt groups, each of which is associated with one of four interrupt priorities. Any interrupt service request is serviced also when the device is in idle mode by going back to active mode. Some interrupts can also wake up the device from sleep mode (when in sleep mode, the device is in one of the three low-power modes PM1, PM2, or PM3); see Chapter 4 for more details.

The **memory arbiter** is at the heart of the system, as it connects the CPU and DMA controller with the physical memories and all peripherals through the SFR bus. The memory arbiter has four memory access points, access of which can map to one of three physical memories: SRAM, flash memory, and XREG/SFR registers. It is responsible for performing arbitration and sequencing between simultaneous memory accesses to the same physical memory.

The **4/6/8-KB SRAM** maps to the DATA memory space and to parts of the XDATA memory spaces. The SRAM is an ultralow-power SRAM that retains its contents in all power modes. This is an important feature for low-power applications.

The **32/64/96/128/256 KB flash block** provides in-circuit programmable non-volatile program memory for the device, and maps into the CODE and XDATA memory spaces. In addition to holding program code and constants, the non-volatile memory allows the application to save data that must be preserved such that it is available after restarting the device. Using this feature one can, e.g., use saved network-specific data to avoid the need for a full start-up and network find-and-join process.

#### 1.1.2 Clocks and Power Management

The digital core and peripherals are powered by a 1.8-V low-dropout **voltage regulator** (Chapter 26). Additionally, the CC253x/CC2540/41 contains a power management functionality that allows the use of different low-power modes (PM1, PM2, and PM3) for low-power applications with a long battery life (see Chapter 4 for more details). Five different **reset** sources exist to reset the device; see Chapter 5 for more details.

#### 1.1.3 Peripherals

The CC253x/CC2540/41 includes many different peripherals that allow the application designer to develop advanced applications. Not all peripherals are present on all devices. See Table 0-1 for a listing of which peripherals are present on each device.

The **debug interface** (Chapter 3) implements a proprietary two-wire serial interface that is used for in-circuit debugging. Through this debug interface, it is possible to perform an erasure of the entire flash memory, control which oscillators are enabled, stop and start execution of the user program, execute supplied instructions on the 8051 core, set code breakpoints, and single-step through instructions in the code. Using these techniques, it is possible to perform in-circuit debugging and external flash programming elegantly.

The device contains flash memory for storage of program code. The flash memory is programmable from the user software and through the debug interface (as mentioned previously). The **flash controller** (Chapter 6) handles writing and erasing the embedded flash memory. The flash controller allows page-wise erasure and 4-bytewise programming.

The **I/O controller** (Chapter 7) is responsible for all general-purpose I/O pins. The CPU can configure whether peripheral modules control certain pins or whether they are under software control, and if so, whether each pin is configured as an input or output and if a pullup or pulldown resistor in the pad is connected. CPU interrupts can be enabled on each pin individually. Each peripheral that connects to the I/O pins can choose between two different I/O pin locations to ensure flexibility in various applications.

A versatile five-channel **DMA controller** (Chapter 8) is available in the system, accesses memory using the XDATA memory space, and thus has access to all physical memories. Each channel (trigger, priority, transfer mode, addressing mode, source and destination pointers, and transfer count) is configured with DMA descriptors anywhere in memory. Many of the hardware peripherals (AES core, flash controller, USARTs, timers, ADC interface) achieve highly efficient operation by using the DMA controller for data transfers between SFR or XREG addresses and flash/SRAM.

**Timer 1** (Chapter 9) is a 16-bit timer with timer/counter/PWM functionality. It has a programmable prescaler, a 16-bit period value, and five individually programmable counter/capture channels, each with a 16-bit compare value. Each of the counter/capture channels can be used as a PWM output or to capture the timing of edges on input signals. It can also be configured in IR generation mode, where it counts Timer 3 periods and the output is ANDed with the output of Timer 3 to generate modulated consumer IR signals with minimal CPU interaction (see Section 9.9).

**Timer 2 (MAC Timer)** (Chapter 22) is specially designed for supporting an IEEE 802.15.4 MAC or other time-slotted protocol in software. The timer has a configurable timer period and a 24-bit overflow counter that can be used to keep track of the number of periods that have transpired. A 40-bit capture register is also used to record the exact time at which a start-of-frame delimiter is received/transmitted or the exact time at which transmission ends, as well as two 16-bit output compare registers and two 24-bit overflow compare registers that can send various command strobes (start RX, start TX, etc.) at specific times to the radio modules.

**Timer 3 and Timer 4** (Chapter 10) are 8-bit timers with timer/counter/PWM functionality. They have a programmable prescaler, an 8-bit period value, and one programmable counter channel with an 8-bit compare value. Each of the counter channels can be used as a PWM output.

The **Sleep Timer** (Chapter 11) is an ultralow-power timer that counts 32-kHz crystal oscillator or 32-kHz RC oscillator periods. The Sleep Timer runs continuously in all operating modes except power mode 3 (PM3). Typical applications of this timer are as a real-time counter or as a wake-up timer for coming out of power mode 1 (PM1) or power mode 2 (PM2).

The **ADC** (Chapter 12) supports 7 bits (30 kHz bandwidth) to 12 bits (4 kHz bandwidth) of resolution. DC and audio conversions with up to eight input channels (Port 0) are possible. The inputs can be selected as single-ended or differential. The reference voltage can be internal, AVDD, or a single-ended or differential external signal. The ADC also has a temperature-sensor input channel. The ADC can automate the process of periodic sampling or conversion over a sequence of channels.

The **battery monitor** (Chapter 13) (CC2533 only) enables simple voltage monitoring in devices that do not include an ADC. It is designed such that it is accurate in the voltage areas around 2 V, with lower resolution at higher voltages.

The **random-number generator** (Chapter 14) uses a 16-bit LFSR to generate pseudorandom numbers, which can be read by the CPU or used directly by the command strobe processor. It can be seeded with random data from noise in the radio ADC.

The **AES coprocessor** (Chapter 15) allows the user to encrypt and decrypt data using the AES algorithm with 128-bit keys. The core is able to support the security operations required by IEEE 802.15.4 MAC security, the ZigBee network layer, and the application layer.

A built-in **Watchdog Timer** (Chapter 16) allows the device to reset itself in case the firmware hangs. When enabled by software, the Watchdog Timer must be cleared periodically; otherwise, it resets the device when it times out. It can alternatively be configured for use as a general 32-kHz timer.

**USART 0** and **USART 1** (Chapter 18) are each configurable as either a SPI master/slave or a UART. They provide double buffering on both RX and TX and hardware flow control, and are thus well suited to high-throughput full-duplex applications. Each has its own high-precision baud-rate generator, thus leaving the ordinary timers free for other uses.

The **I**²**C** module (Chapter 20) (CC2533 and CC2541) provides a digital peripheral connection with two pins and supports both master and slave operation.

The **USB 2.0 controller** (Chapter 21) (CC2531 and CC2540) operates at Full-Speed, 12 Mbps transfer rate. The controller has five bidirectional endpoints in addition to control endpoint 0. The endpoints support Bulk, Interrupt, and Isochronous operation for implementation of a wide range of applications. The 1024 bytes of dedicated, flexible FIFO memory combined with DMA access ensures that a minimum of CPU involvement is needed for USB communication.

The **operational amplifier** (Chapter 18) (CC2530, CC2531, and CC2540) is intended to provide front-end buffering and gain for the ADC. Both the inputs as well as the output are available on pins, so the feedback network is fully customizable. A chopper-stabilized mode is available for applications that need good accuracy with high gain.

The ultralow-power **analog comparator** (Chapter 19) (CC2530, CC2531, CC2540, and CC2541) enables applications to wake up from PM2 or PM3 based on an analog signal. Both inputs are brought out to pins; the reference voltage must be provided externally. The comparator output is mapped into the digital I/O port and can be treated by the MCU as a regular digital input.

#### 1.1.4 Radio

The CC2540/41 provides a Bluetooth low-energy-compliant radio transceiver. The RF core which controls the analog and digital radio modules is only indirectly accessible through API commands to the BLE stack. More details about the CC2540/41 BLE radio can be found in Chapter 24. The CC2541 can also be run in proprietary modes; more details can be found in Chapter 25.

The CC253x device family provides an **IEEE 802.15.4-compliant radio transceiver**. The RF Core controls the analog radio modules. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events. The radio also includes a packet-filtering and address-recognition module. More details about the CC253x radio can be found in Chapter 23.

#### 1.2 Applications

As shown in the overview (Section 1.1), this user's guide focuses on the functionality of the different modules that are available to build different types of applications based on the CC253x/CC2540/41 device family. When looking at the complete application development process, additional information is useful. However, as this information and help is not device-specific (i.e., not unique for the CC253x/CC2540/41 device family), see the additional information sources in the following paragraphs.

The first step is to set up the development environment (HW, tools, etc.) by purchasing a **development kit** (see the device-specific product Web site to find links to the relevant development kits). The development kits come with an out-of-the-box demo and information on how to set up the development environment; install required drivers (done easily by installing the **SmartRF software**, Section 27.1), set up the compiler tool chain, etc. As soon as one has installed the development environment, one is ready to start the application development.

The easiest way to write the application software is to base the application on one of the available standard protocols (**RemoTI** network protocol, Section 27.2; **TIMAC** software, Section 27.4; **Z-Stack** software for ZigBee-compliant solutions, Section 27.5); BLE stack software for Bluetooth low-energy-compliant solutions Section 27.6; or the proprietary **SimpliciTI** network protocol (Section 27.3). They all come with several sample applications.

For the hardware layout design of the user-specific HW, the designer can find reference designs on the different product pages (Section B.1). By copying these designs, the designer achieves optimal performance. The developed HW can then be tested easily using the SmartRF Studio software (Section 27.1).

In case the final system should not have the expected performance, it is recommended to try out the developed software on the development kit hardware and see how it works there. To check the user-specific HW, it is a good first step to use SmartRF Studio software to compare the development kit performance versus the user-specific HW using the same settings.

The user can also find additional information and help by joining the **Low-Power RF Online Community** (Section B.2) and by subscribing to the **Low-Power RF eNewsletter** (Section B.4).

To contact a third-party to help with development or to use modules, check out the Texas Instruments **Low-Power RF Developer Network** (Section B.3).



## 8051 CPU

Page

The System-on-Chip solution is based on an enhanced 8051 core. More details regarding the core, memory map, instruction set, and interrupts are described in the following subsections.

#### Topic

2.1 8051 CPU Introduction	28
2.2 Memory	
2.3 CPU Registers	
2.4 Instruction Set Summary	
2.5 Interrupts	43



#### 2.1 8051 CPU Introduction

The enhanced 8051 core uses the standard 8051 instruction set. Instructions execute faster than the standard 8051 due to the following:

- One clock per instruction cycle is used as opposed to 12 clocks per instruction cycle in the standard 8051.
- Wasted bus states are eliminated.

Because an instruction cycle is aligned with memory fetch when possible, most of the single-byte instructions are performed in a single clock cycle. In addition to the speed improvement, the enhanced 8051 core also includes architectural enhancements:

- A second data pointer
- An extended 18-source interrupt unit

The 8051 core is object-code-compatible with the industry-standard 8051 microcontroller. That is, object code compiled with an industry-standard 8051 compiler or assembler executes on the 8051 core and is functionally equivalent. However, because the 8051 core uses a different instruction timing than many other 8051 variants, existing code with timing loops may require modification. Also, because the peripheral units such as timers and serial ports differ from those on other 8051 cores, code which includes instructions using the peripheral-unit SFRs does not work correctly.

Flash prefetching is not enabled by default, but improves CPU performance by up to 33%. This is at the expense of slightly increased power consumption, but in most cases improves energy consumption as it is faster. Flash prefetching can be enabled in the FCTL register.

#### 2.2 Memory

The 8051 CPU architecture has four different memory spaces. The 8051 has separate memory spaces for program memory and data memory. The 8051 memory spaces are the following (see Section 2.2.1 and Section 2.2.2 for details):

CODE. A read-only memory space for program memory. This memory space addresses 64 KB.

**DATA**. A read/write data memory space that can be directly or indirectly accessed by a single-cycle CPU instruction. This memory space addresses 256 bytes. The lower 128 bytes of the DATA memory space can be addressed either directly or indirectly, the upper 128 bytes only indirectly.

**XDATA**. A read/write data memory space, access to which usually requires 4–5 CPU instruction cycles. This memory space addresses 64 KB. Access to XDATA memory is also slower than DATA access, as the CODE and XDATA memory spaces share a common bus on the CPU core, and instruction prefetch from CODE can thus not be performed in parallel with XDATA accesses.

**SFR**. A read/write register memory space which can be directly accessed by a single CPU instruction. This memory space consists of 128 bytes. For SFR registers whose address is divisible by eight, each bit is also individually addressable.

The four different memory spaces are distinct in the 8051 architecture, but are partly overlapping in the device to ease DMA transfers and hardware debugger operation.

How the different memory spaces are mapped onto the three physical memories (flash program memory, SRAM, and memory-mapped registers) is described in Section 2.2.1 and Section 2.2.2.

#### 2.2.1 Memory Map

The memory map differs from the standard 8051 memory map in two important aspects, as described in the following paragraphs.

First, in order to allow the DMA controller access to all physical memory and thus allow DMA transfers between the different 8051 memory spaces, parts of SFR and the DATA memory space are mapped into the XDATA memory space.

Second, two alternative schemes for CODE memory space mapping can be used. The first scheme is the standard 8051 mapping where only the program memory (i.e., flash memory) is mapped to CODE memory space. This mapping is the default after a device reset.



The second scheme is used for executing code from SRAM. In this mode, the SRAM is mapped into the region of 0x8000 through ( $0x8000 + SRAM_SIZE - 1$ ). The map is shown in Figure 2-2. Executing code from SRAM improves performance and reduces power consumption.

The upper 32 KB of XDATA is a read-only area called XBANK. Any of the available 32 KB flash banks can be mapped in here. This gives software access to the whole flash memory. This area is typically used to store additional constant data.

Details about mapping of all 8051 memory spaces are given in Section 2.2.2.

The memory map showing how the different physical memories are mapped into the CPU memory spaces is given in Figure 2-1 through Figure 2-3. The number of available flash banks depends on the flash size option.

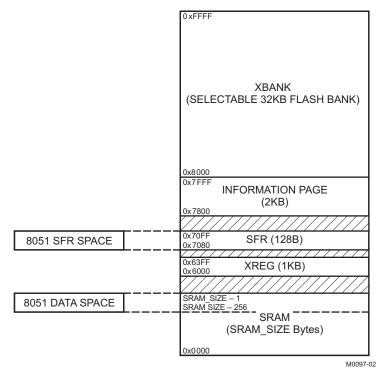
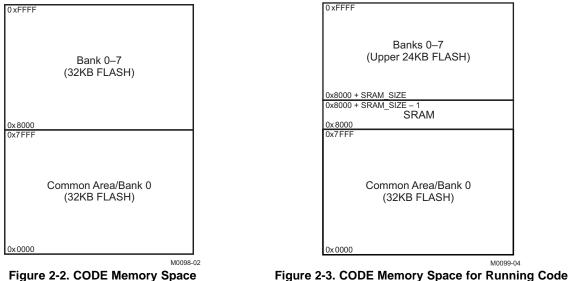


Figure 2-1. XDATA Memory Space (Showing SFR and DATA Mapping)





#### Memory

#### 2.2.2 CPU Memory Space

**XDATA memory space.** The XDATA memory map is given in Figure 2-1.

The SRAM is mapped into address range of 0x0000 through (SRAM_SIZE – 1).

The XREG area is mapped into the 1-KB address range (0x6000–0x63FF). These registers are additional registers, effectively extending the SFR register space. Some peripheral registers and most of the radio control and data registers are mapped in here.

The SFR registers are mapped into address range (0x7080–0x70FF).

The flash information page (2 KB) is mapped into the address range (0x7800–0x7FFF). This is a read-only area and contains various information about the device.

The upper 32 KB of the XDATA memory space (0x8000–0xFFF) is a read-only flash code bank (XBANK) and can be mapped to any of the available flash banks using the MEMCTR.XBANK[2:0] bits.

The mapping of flash memory, SRAM, and registers to XDATA allows the DMA controller and the CPU access to all the physical memories in a single unified address space.

Writing to unimplemented areas in the memory map (shaded in the figure) has no effect. Reading from unimplemented areas returns 0x00. Writes to read-only regions, i.e., flash areas, are ignored.

**CODE memory space.** The CODE memory space is 64 KB and is divided into a common area (0x0000–0x7FFF) and a bank area (0x8000–0xFFFF) as shown in Figure 2-2. The common area is always mapped to the lower 32 KB of the physical flash memory (bank 0). The bank area can be mapped to any of the available 32-KB flash banks (from 0 to 7). The number of available flash banks depends on the flash size option. Use the flash-bank-select register, FMAP, to select the flash bank. On 32-KB devices, no flash memory can be mapped into the bank area. Reads from this region return 0x00 on these devices.

To allow program execution from SRAM, it is possible to map the available SRAM into the lower range of the bank area from 0x8000 through (0x8000 + SRAM_SIZE – 1). The rest of of the currently selected bank is still mapped into the address range from (0x8000 + SRAM_SIZE) through 0xFFFF). Set the MEMCTR.XMAP bit to enable this feature.

**DATA memory space.** The 8-bit address range of DATA memory is mapped into the upper 256 bytes of the SRAM, i.e., the address range from (SRAM_SIZE – 256) through (SRAM_SIZE – 1).

**SFR memory space.** The 128-entry hardware register area is accessed through this memory space. The SFR registers are also accessible through the XDATA address space at the address range (0x7080–0x70FF). Some CPU-specific SFR registers reside inside the CPU core and can only be accessed using the SFR memory space and not through the duplicate mapping into XDATA memory space. These specific SFR registers are listed in *SFR Registers*.

#### 2.2.3 Physical Memory

**RAM**. All devices contain static RAM. At power on, the content of RAM is undefined. RAM content is retained in all power modes.

**Flash Memory.** The on-chip flash memory is primarily intended to hold program code and constant data. The flash memory has the following features:

- Page size: 1 KB or 2 KB (details are given in the data sheet of the device.)
- Flash-page erase time: 20 ms
- Flash-chip (mass) erase time: 20 ms
- Flash write time (4 bytes): 20 µs
- Data retention (at room temperature): 100 years
- Program/erase endurance: 20,000 cycles

The flash memory is organized as a set of 1- or 2-KB pages. The 16 bytes of the upper available page



contain page-lock bits and the debug-lock bit. There is one lock bit for each page, except the lock-bit page which is implicitly locked when not in debug mode. When the lock bit for a page is 0, it is impossible to erase/write that page. When the debug lock bit is 0, most of the commands on the debug interface are ignored. The primary purpose of the debug lock bit is to protect the contents of the flash against read-out. The Flash Controller is used to write and erase the contents of the flash memory.

When the CPU reads instructions and constants from flash memory, it fetches the instructions through a cache. Four bytes of instructions and four bytes of constant data are cached, at 4-byte boundaries. That is, when the CPU reads from address 0x00F1 for example, bytes 0x00F0–0x00F3 are cached. A separate prefetch unit is capable of prefetching 4 additional bytes of instructions. The cache is provided mainly to reduce power consumption by reducing the amount of time the flash memory is accessed. The cache may be disabled with the FCTL.CM[1:0] register bits. Doing so increases power consuption and is not recommended. The execution time from flash is not cycle-accurate when using the default cache mode and the cache mode with prefetch, i.e., one cannot determine exactly the number of clock cycles a set of instructions takes. To obtain cycle-accurate execution, enable the real-time cache mode and ensure all DMA transfers have low priority. The prefetch mode improves performance by up to 33%, at the expense of increased power consumption due to wasted flash reads. Typically, performance improves by 15%–20%. Total energy, however, may decrease (depending on the application) due to fewer wasted clock cycles waiting for the flash to return instructions/data. This is very application-dependent and requires the use of power modes to be effective.

The **Information Page** is a 2-KB read-only region that stores various device information. Among other things, it contains for IEEE 802.15.4 or *Bluetooth* low energy compliant devices a unique IEEE address from the TI range of addresses. For CC253x, this is a 64-bit IEEE address stored with least-significant byte first at XDATA address 0x780C. For CC2540/41, this is a 48-bit IEEE address stored with least-significant byte first at XDATA address 0x780E.

**SFR Registers.** The special function registers (SFRs) control several of the features of the 8051 CPU core and/or peripherals. Many of the 8051 core SFRs are identical to the standard 8051 SFRs. However, there are additional SFRs that control features that are not available in the standard 8051. The additional SFRs are used to interface with the peripheral units and RF transceiver.

Table 2-1 shows the addresses of all SFRs in the device. The 8051 internal SFRs are shown with gray background, whereas the other SFRs are the SFRs specific to the device.

**NOTE:** All internal SFRs (shown with gray background in Table 2-1), can only be accessed through SFR space, as these registers are not mapped into XDATA space. One exception is the port registers (P0, P1, and P2) which are readable from XDATA.

Register Name	SFR Address	Module	Description	
ADCCON1	0xB4	ADC	ADC control 1	
ADCCON2	0xB5	ADC	ADC control 2	
ADCCON3	0xB6	ADC	ADC control 3	
ADCL	0xBA	ADC	ADC data low	
ADCH	0xBB	ADC	ADC data high	
RNDL	0xBC	ADC	Random number generator data low	
RNDH	0xBD	ADC	Random number generator data high	
ENCDI	0xB1	AES	Encryption/decryption input data	
ENCDO	0xB2	AES	Encryption/decryption output data	
ENCCS	0xB3	AES	Encryption/decryption control and status	
P0	0x80	CPU	Port 0. Readable from XDATA (0x7080)	
SP	0x81	CPU	Stack pointer	
DPL0	0x82	CPU	Data pointer 0 low byte	
DPH0	0x83	CPU	Data pointer 0 high byte	
DPL1	0x84	CPU	Data pointer 1 low byte	

Table 2-1. SFR Overview

Table 2-1. SFR Overview (continued)

Register Name	SFR Address	Module	Description	
DPH1	0x85	CPU	Data pointer 0 high byte	
PCON	0x87	CPU	Power mode control	
TCON	0x88	CPU	Interrupt flags	
P1	0x90	CPU	Port 1. Readable from XDATA (0x7090)	
DPS	0x92	CPU	Data pointer select	
SOCON	0x98	CPU	Interrupt flags 2	
IEN2	0x9A	CPU	Interrupt enable 2	
S1CON	0x9B	CPU	Interrupt flags 3	
P2	0xA0	CPU	Port 2. Readable from XDATA (0x70A0)	
IENO	0xA8	CPU	Interrupt enable 0	
IP0	0xA9	CPU	Interrupt priority 0	
IEN1	0xB8	CPU	Interrupt enable 1	
IP1	0xB9	CPU	Interrupt priority 1	
IRCON	0xC0	CPU	Interrupt flags 4	
PSW	0xD0	CPU	Program status Word	
ACC	0xE0	CPU	Accumulator	
IRCON2	0xE8	CPU	Interrupt flags 5	
В	0xF0	CPU	B register	
DMAIRQ	0xD1	DMA	DMA interrupt flag	
DMA1CFGL	0xD2	DMA	DMA channel 1–4 configuration address low	
DMA1CFGH	0xD3	DMA	DMA channel 1–4 configuration address high	
DMA0CFGL	0xD4	DMA	DMA channel 0 configuration address low	
DMA0CFGH	0xD5	DMA	DMA channel 0 configuration address high	
DMAARM	0xD6	DMA	DMA channel armed	
DMAREQ	0xD7	DMA	DMA channel start request and status	
—	0xAA	—	Reserved	
—	0x8E	—	Reserved	
—	0x99	—	Reserved	
—	0xB0	—	Reserved	
—	0xB7	—	Reserved	
—	0xC8	_	Reserved	
P0IFG	0x89	IOC	Port 0 interrupt status flag	
P1IFG	0x8A	IOC	Port 1 interrupt status flag	
P2IFG	0x8B	IOC	Port 2 interrupt status flag	
PICTL	0x8C	IOC	Port pins interrupt mask and edge	
POIEN	0xAB	IOC	Port 0 interrupt mask	
P1IEN	0x8D	IOC	Port 1 interrupt mask	
P2IEN	0xAC	IOC	Port 2 interrupt mask	
POINP	0x8F	IOC	Port 0 input mode	
PERCFG	0xF1	IOC	Peripheral I/O control	
APCFG	0xF2	IOC	Analog peripheral I/O configuration	
POSEL	0xF3	IOC	Port 0 function select	
P1SEL	0xF4	IOC	Port 1 function select	
P2SEL	0xF5	IOC	Port 2 function select	
P1INP	0xF6	IOC	Port 1 input mode	
P2INP	0xF7	IOC	Port 2 input mode	
P0DIR	0xFD	IOC	Port 0 direction	

Table 2-1. SFR Overview (continued)

Register Name	SFR Address	Module	Description	
P1DIR	0xFE	IOC	Port 1 direction	
P2DIR	0xFF	IOC	Port 2 direction	
PMUX	0xAE	IOC	Power-down signal mux	
MPAGE	0x93	MEMORY	Memory page select	
MEMCTR	0xC7	MEMORY	Memory system control	
FMAP	0x9F	MEMORY	Flash-memory bank mapping	
RFIRQF1	0x91	RF	RF interrupt flags MSB	
RFD	0xD9	RF	RF data	
RFST	0xE1	RF	RF command strobe	
RFIRQF0	0xE9	RF	RF interrupt flags LSB	
RFERRF	0xBF	RF	RF error interrupt flags	
ST0	0x95	ST	Sleep Timer 0	
ST1	0x96	ST	Sleep Timer 1	
ST2	0x97	ST	Sleep Timer 2	
STLOAD	0xAD	ST	Sleep-timer load status	
SLEEPCMD	0xBE	PMC	Sleep-mode control command	
SLEEPSTA	0x9D	PMC	Sleep-mode control status	
CLKCONCMD	0xC6	PMC	Clock control command	
CLKCONSTA	0x9E	PMC	Clock control status	
T1CC0L	0xDA	Timer 1	Timer 1 channel 0 capture/compare value low	
T1CC0H	0xDB	Timer 1	Timer 1 channel 0 capture/compare value high	
T1CC1L	0xDC	Timer 1	Timer 1 channel 1 capture/compare value low	
T1CC1H	0xDD	Timer 1	Timer 1 channel 1 capture/compare value high	
T1CC2L	0xDE	Timer 1	Timer 1 channel 2 capture/compare value low	
T1CC2H	0xDF	Timer 1	Timer 1 channel 2 capture/compare value high	
T1CNTL	0xE2	Timer 1	Timer 1 counter low	
T1CNTH	0xE3	Timer 1	Timer 1 counter high	
T1CTL	0xE4	Timer 1	Timer 1 control and status	
T1CCTL0	0xE5	Timer 1	Timer 1 channel 0 capture/compare control	
T1CCTL1	0xE6	Timer 1	Timer 1 channel 1 capture/compare control	
T1CCTL2	0xE7	Timer 1	Timer 1 channel 2 capture/compare control	
T1STAT	0xAF	Timer 1	Timer 1 status	
T2CTRL	0x94	Timer 2	Timer 2 control	
T2EVTCFG	0x9C	Timer 2	Timer 2 event configuration	
T2IRQF	0xA1	Timer 2	Timer 2 interrupt flags	
T2M0	0xA2	Timer 2	Timer 2 multiplexed register 0	
T2M1	0xA3	Timer 2	Timer 2 multiplexed register 1	
T2MOVF0	0xA4	Timer 2	Timer 2 multiplexed overflow register 0	
T2MOVF1	0xA5	Timer 2	Timer 2 multiplexed overflow register 1	
T2MOVF2	0xA6	Timer 2	Timer 2 multiplexed overflow register 2	
T2IRQM	0xA7	Timer 2	Timer 2 interrupt mask	
T2MSEL	0xC3	Timer 2	2 Timer 2 multiplex select	
T3CNT	0xCA	Timer 3		
T3CTL	0xCB	Timer 3	Timer 3 control	
T3CCTL0	0xCC	Timer 3		
T3CC0	0xCD	Timer 3	Timer 3 channel 0 compare value	
T3CCTL1	0xCE	Timer 3	Timer 3 channel 1 compare control	

[						
Register Name	SFR Address	Module	Description			
T3CC1	0xCF	Timer 3	Timer 3 channel 1 compare value			
T4CNT	0xEA	Timer 4	Timer 4 counter			
T4CTL	0xEB	Timer 4	Timer 4 control			
T4CCTL0	0xEC	Timer 4	Timer 4 channel 0 compare control			
T4CC0	0xED	Timer 4	Timer 4 channel 0 compare value			
T4CCTL1	0xEE	Timer 4	Timer 4 channel 1 compare control			
T4CC1	0xEF	Timer 4	Timer 4 channel 1 compare value			
TIMIF	0xD8	TMINT	Timers 1/3/4 joint interrupt mask/flags			
U0CSR	0x86	USART 0	USART 0 control and status			
U0DBUF	0xC1	USART 0	USART 0 receive/transmit data buffer			
U0BAUD	0xC2	USART 0	USART 0 baud-rate control			
U0UCR	0xC4	USART 0	USART 0 UART control			
U0GCR	0xC5	USART 0	USART 0 generic control			
U1CSR	0xF8	USART 1	USART 1 control and status			
U1DBUF	0xF9	USART 1	USART 1 receive/transmit data buffer			
U1BAUD	0xFA	USART 1	USART 1 baud-rate control			
U1UCR	0xFB	USART 1	USART 1 UART control			
U1GCR	0xFC	USART 1	USART 1 generic control			
WDCTL	0xC9	WDT	Watchdog Timer control			

Table 2-1. SFR Overview (continued)



**XREG Registers**. The XREG registers are additional registers in the XDATA memory space. These registers are mainly used for radio configuration and control. For more details regarding each register, see the corresponding module/peripheral chapter. Table 2-2 gives a descriptive overview of the register address space.

XDATA Address	Register Name	Description	
0x6000–0x61FF	_	Radio registers (see CC253x Radio Section 23.15 or CC2540 Radio Section 24.1 or CC2541 Radio Section 25.12 for complete list)	
0x61A6	MONMUX	Battery monitor MUX (CC2533)	
0.0170	OPAMPMC	Operational amplifier mode control (CC2530/CC2531)	
0x61AD	OPAMPMC	Operational amplifier mode control (CC2540)	
0x6200–0x622B	—	USB registers (see Section 21.12 for complete list)	
0x6230	I2CCFG	I ² C control	
0x6231	I2CSTAT	I ² C status	
0x6232	I2CDATA	I ² C data	
0x6233	I2CADDR	I ² C own slave address	
0x6234	I2CWC	Wrapper control	
0x6235	12CIO	GPIO	
0x6243	OBSSEL0	Observation output control register 0	
0x6244	OBSSEL1	Observation output control register 1	
0x6245	OBSSEL2	Observation output control register 2	
0x6246	OBSSEL3	Observation output control register 3	
0x6247	OBSSEL4	Observation output control register 4	
0x6248	OBSSEL5	Observation output control register 5	
0x6249	CHVER	Chip version	
0x624A	CHIPID	Chip identification	
0x624B	TR0	Test register 0	
0x6260	DBGDATA	Debug interface write data	
0x6262	SRCRC	Sleep reset CRC	
0x6264	BATTMON	Battery monitor	
0x6265	IVCTRL	Analog control register	
0x6270	FCTL	Flash control	
0x6271	FADDRL	Flash address low	
0x6272	FADDRH	Flash address high	
0x6273	FWDATA	Flash write data	
0x6276	CHIPINFO0	Chip information byte 0	
0x6277	CHIPINF01	Chip information byte 1	
0x6281	IRCTL	Timer 1 IR generation control	
0x6290	CLD	Clock-loss detection	
0x6230	T1CCTL0	Timer 1 channel 0 capture/compare control (additional XREG mapping of SFR register)	
0x62A1	T1CCTL1	Timer 1 channel 1 capture/compare control (additional XREG mapping of SFR register)	
0x62A2	T1CCTL2	Timer 1 channel 2 capture/compare control (additional XREG mapping of SFR register)	
0x62A3	T1CCTL3	Timer 1 channel 3 capture/compare control	
0x62A4	T1CCTL4	Timer 1 channel 4 capture/compare control	
0x62A6	T1CC0L	Timer 1 channel 0 capture/compare value low (additional XREG mapping of SFR register)	
0x62A7	T1CC0H	Timer 1 channel 0 capture/compare value high (additional XREG mapping of SFR register)	

Table 2-2	Overview of	XREG	Registers
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XDATA Address	Register Name	Description
0x62A8	T1CC1L	Timer 1 channel 1 capture/compare value low (additional XREG mapping of SFR register)
0x62A9	T1CC1H	Timer 1 channel 1 capture/compare value high (additional XREG mapping of SFR register)
0x62AA	T1CC2L	Timer 1 channel 2 capture/compare value low (additional XREG mapping of SFR register)
0x62AB	T1CC2H	Timer 1 channel 2 capture/compare value high (additional XREG mapping of SFR register)
0x62AC	T1CC3L	Timer 1 channel 3 capture/compare value low
0x62AD	T1CC3H	Timer 1 channel 3 capture/compare value high
0x62AE	T1CC4L	Timer 1 channel 4 capture/compare value low
0x62AF	T1CC4H	Timer 1 channel 4 capture/compare value high
0x62B0	STCC	Sleep Timer capture control
0x62B1	STCS	Sleep Timer capture status
0x62B2	STCV0	Sleep Timer capture value byte 0
0x62B3	STCV1	Sleep Timer capture value byte 1
0x62B4	STCV2	Sleep Timer capture value byte 2
0x62C0	OPAMPC	Operational amplifier control
0x62C1	OPAMPS	Operational amplifier status
0x62D0	CMPCTL	Analog comparator control and status

Table 2-2. Overview of XREG Registers (continued)

#### 2.2.4 XDATA Memory Access

The **MPAGE** register is used during instructions MOVX A, @Ri and MOVX @Ri, A. MPAGE gives the 8 most-significant address bits, whereas the register Ri gives the 8 least-significant bits.

In some 8051 implementations, this type of XDATA access is performed using P2 to give the most-significant address bits. Existing software may therefore have to be adapted to make use of MPAGE instead of P2.

#### MPAGE (0x93) – Memory Page Select

Bit	Name	Reset	R/W	Description
7:0	MPAGE[7:0]	0x00	R/W	Memory page, high-order bits of address in MOVX instruction

#### 2.2.5 Memory Arbiter

The memory arbiter handles CPU and DMA access to all physical memory except the CPU internal registers. When an access conflict between the CPU and DMA occurs, the memory arbiter stalls one of the bus masters so that the conflict is resolved.

The control registers MEMCTR and FMAP are used to control various aspects of the memory subsystem. The MEMCTR and FMAP registers are described as follows.

MEMCTR.XMAP must be set to enable program execution from RAM.

The flash-bank map register, FMAP, controls mapping of physical 32-KB code banks to the program address region 0x8000–0xFFFF in CODE memory space.



MEMO	CTR (0xC7) – Memory	Arbiter Control		
Bit	Name	Reset	R/W	Description
7:4	—	0000	R0	Reserved
3	ХМАР	0	R/W	XDATA map to code. When this bit is set, the SRAM XDATA region, from 0x0000 through (SRAM_SIZE – 1) is mapped into the CODE region from 0x8000 through (0x8000 + SRAM_SIZE – 1). This enables execution of program code from RAM.
				0: SRAM map into CODE feature disabled
				1: SRAM map into CODE feature enabled
2:0	XBANK [ 2:0 ]	000	R/W	XDATA bank select. Controls which code bank of the physical flash memory is mapped into the XDATA region (0x8000–0xFFF). When set to 0, the root bank is mapped in. Valid settings depend on the flash size for the device. Writing an invalid setting is ignored, i.e., no update to XBANK[2:0] is performed. 32-KB version: 0 only (i.e, the root bank is always mapped in.) 64-KB version: 0–1 96-KB version: 0–2 128-KB version: 0–3 256-KB version: 0–7

#### FMAP (0x9F) – Flash Bank Map

Bit	Name	Reset	R/W	Description
7:3	—	0000 0	R0	Reserved
2:0	MAP[2:0]	001	R/W	Flash bank map. Controls which bank is mapped into the bank area of the CODE memory space (0x8000–0xFFFF). When set to 0, the root bank is mapped in. Valid settings depend on the flash size for the device. Writing an invalid setting is ignored, i.e., no update to MAP[2:0] is performed. 32-KB version: No value can be written. Bank area is only used for running program code from SRAM. See MEMCTR.XMAP. 64-KB version: 0–1 96-KB version: 0–2 128-KB version: 0–3 256-KB version: 0–7

### 2.3 CPU Registers

This section describes the internal registers found in the CPU.

### 2.3.1 Data Pointers

Two data pointers, DPTR0 and DPTR1, exist to accelerate the movement of data blocks to/from memory. The data pointers are generally used to access CODE or XDATA space. For example:

MOVC A,@A+DPTR

MOV A,@DPTR.

The data pointer select bit, bit 0 in the data pointer select register DPS, chooses which data pointer is the active one during execution of an instruction that uses the data pointer, e.g., in one of the preceding instructions.

The data pointers are two bytes wide, consisting of the following SFRs:

- DPTR0–DPH0:DPL0
- DPTR1-DPH1:DPL1

#### DPH0 (0x83) - Data Pointer-0 High Byte

Bit	Name	Reset	R/W	Description
7:0	DPH0[7:0]	0x00	R/W	Data pointer-0, high byte

#### DPL0 (0x82) – Data Pointer-0 Low Byte

Bit	Name	Reset	R/W	Description
7:0	DPL0[7:0]	0x00	R/W	Data pointer-0, low byte



#### CPU Registers

DPH1	(0x85) – Data Pointer-	1 High Byte		
Bit	Name	Reset	R/W	Description
7:0	DPH1[7:0]	0x00	R/W	Data pointer-1, high byte

#### DPL1 (0x84) - Data Pointer-1 Low Byte

Bit	Name	Reset	R/W	Description
7:0	DPL1[7:0]	0x00	R/W	Data pointer-1, low byte

#### DPS (0x92) - Data-Pointer Select

Bit	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved
0	DPS	0	R/W	Data pointer select. Selects active data pointer. 0: DPTR0 1: DPTR1

### 2.3.2 Registers R0-R7

There are four register banks (not to be confused with CODE memory space banks that only apply to flash memory organization) of eight registers each. These register banks are mapped in the DATA memory space at addresses 0x00-0x07, 0x08-0x0F, 0x10-0x17, and 0x18-0x1F. Each register bank contains the eight 8-bit registers R0-R7. The register bank to be used is selected through the program status word PSW.RS[1:0]. Register bank 0 uses flip-flops internally for storing the values (SRAM is bypassed/unused), whereas banks 1–3 use SRAM for storage. This is done to save power. Typically, the current consumption goes down by approximately 200  $\mu$ A by using register bank 0 instead of register banks 1–3.

### 2.3.3 Program Status Word

The program status word (PSW) contains several bits that show the current state of the CPU. The PSW is accessible as an SFR, and it is bit-addressable. The PSW is shown as follows and contains the carry flag, auxiliary carry flag for BCD operations, register-select bits, overflow flag, and parity flag. Two bits in the PSW are uncommitted and can be used as user-defined status flags.

Bit	Name	Reset	R/W	Description	
7	СХ	0	R/W	Carry flag. Set to 1 when the last arithmetic operation resulted in a carry (during addition) or borrow (during subtraction); otherwise, cleared to 0 by all arithmetic operations.	
6	AC	0	R/W	Auxiliary carry flag for BCD operations. Set to 1 when the last arithmetic operation resulted in a carry into (during addition) or borrow from (during subtraction) the high-order nibble, otherwise cleared to 0 by all arithmetic operations.	
5	FO	0	R/W	Jser-defined, bit-addressable	
4:3	RS[1:0]	00	R/W	Register bank select bits. Selects which set of R7–R0 registers to use from four possible banks in DATA space.	
				00: Register bank 0, 0x00–0x07	
				01: Register bank 1, 0x08–0x0F	
				10: Register bank 2, 0x10–0x17	
				11: Register bank 3, 0x18–0x1F	
2	ov	0	R/W	Overflow flag, set by arithmetic operations. Set to 1 when the last arithmetic operation is a carry (addition), borrow (subtraction), or overflow (multiply or divide). Otherwise, the bit is cleared to 0 by all arithmetic operations.	
1	F1	0	R/W	User-defined, bit-addressable	
0	P	0	R/W	Parity flag, parity of accumulator set by hardware to 1 if it contains an odd number of 1s; otherwise it is cleared to 0.	

PSW (0xD0) - Program Status Word



#### Instruction Set Summary

### 2.3.4 Accumulator

ACC is the accumulator. This is the source and destination of most arithmetic instructions, data transfers, and other instructions. The mnemonic for the accumulator (in instructions involving the accumulator) is A instead of ACC.

#### ACC (0xE0) - Accumulator

Bit	Name	Reset	R/W	Description
7:0	ACC[7:0]	0x00	R/W	Accumulator

### 2.3.5 B Register

The B register is used as the second 8-bit argument during execution of multiply and divide instructions. When not used for these purposes, it may be used as a scratchpad register to hold temporary data.

#### B (0xF0) - B Register

Bit	Name	Reset	R/W	Description
7:0	B[7:0]	0x00	R/W	B register. Used in MUL/DIV instructions

### 2.3.6 Stack Pointer

The stack resides in DATA memory space and grows upwards. The PUSH instruction first increments the stack pointer (SP) and then copies the byte into the stack. The SP is initialized to 0x07 after a reset, and it is incremented once to start from location 0x08, which is the first register (R0) of the second register bank. Thus, in order to use more than one register bank, the SP should be initialized to a different location not used for data storage.

#### SP (0x81) - Stack Pointer

Bit	Name	Reset	R/W	Description
7:0	SP[7:0]	0x07	R/W	Stack pointer

### 2.4 Instruction Set Summary

The 8051 instruction set is summarized in Table 2-3. All mnemonics copyrighted © Intel Corporation, 1980.

The following conventions are used in the instruction set summary:

- Rn Register R7-R0 of the currently selected register bank
- Direct 8-bit internal data-location address. This can be DATA area (0x00–0x7F) or SFR area (0x80–0xFF).
- @Ri 8-bit internal data location, DATA area (0x00–0xFF) addressed indirectly through register R1 or R0
- #data 8-bit constant included in instruction
- #data16 16-bit constant included in instruction
- addr16 16-bit destination address. Used by LCALL and LJMP. A branch can be anywhere within the 64-KB CODE memory space.
- addr11 11-bit destination address. Used by ACALL and AJMP. The branch is within the same 2-KB page of program memory as the first byte of the following instruction.
- rel Signed (2s-complement) 8-bit offset byte. Used by SJMP and all conditional jumps. Range is –128 to 127 bytes relative to first byte of the following instruction.
- bit Direct addressed bit in DATA area or SFR



The instructions that affect CPU flag settings located in PSW are listed in Table 2-4. Note that operations on the PSW register or bits in PSW also affect the flag settings. Also note that the cycle count for many instructions assumes single-cycle access to the memory element being accessed, i.e., the best-case situation. This is not always the case. Reads from flash may take 1–3 cycles, for example.

Mnemonic	Description	Hex Opcode	Bytes	Cycles
	ARITHMETIC OPERATIO	NS		
ADD A,Rn	Add register to accumulator	28–2F	1	1
ADD A,direct	Add direct byte to accumulator	25	2	2
ADD A,@Ri	Add indirect RAM to accumulator	26–27	1	2
ADD A,#data	Add immediate data to accumulator	24	2	2
ADDC A,Rn	Add register to accumulator with carry flag	38–3F	1	1
ADDC A, direct	Add direct byte to A with carry flag	35	2	2
ADDC A,@Ri	Add indirect RAM to A with carry flag	36–37	1	2
ADDC A,#data	Add immediate data to A with carry flag	34	2	2
SUBB A,Rn	Subtract register from A with borrow	98–9F	1	1
SUBB A,direct	Subtract direct byte from A with borrow	95	2	2
SUBB A,@Ri	Subtract indirect RAM from A with borrow	96–97	1	2
SUBB A,#data	Subtract immediate data from A with borrow	94	2	2
INC A	Increment accumulator	04	1	1
INC Rn	Increment register	08–0F	1	2
INC direct	Increment direct byte	05	2	3
INC @Ri	Increment indirect RAM	06–07	1	3
INC DPTR	Increment data pointer	A3	1	1
DEC A	Decrement accumulator	14	1	1
DEC Rn	Decrement register	18–1F	1	2
DEC direct	Decrement direct byte	15	2	3
DEC @Ri	Decrement indirect RAM	16–17	1	3
MUL AB	Multiply A and B	A4	1	5
DIV A	Divide A by B	84	1	5
DA A	Decimal adjust accumulator	D4	1	1
	LOGICAL OPERATION	S	1	4
ANL A,Rn	AND register to accumulator	58–5F	1	1
ANL A, direct	AND direct byte to accumulator	55	2	2
ANL A,@Ri	AND indirect RAM to accumulator	56–57	1	2
ANL A,#data	AND immediate data to accumulator	54	2	2
ANL direct,A	AND accumulator to direct byte	52	2	3
ANL direct,#data	AND immediate data to direct byte	53	3	4
ORL A,Rn	OR register to accumulator	48–4F	1	1
ORL A, direct	OR direct byte to accumulator	45	2	2
ORL A,@Ri	OR indirect RAM to accumulator	46–47	1	2
ORL A,#data	OR immediate data to accumulator	44	2	2
ORL direct,A	OR accumulator to direct byte	42	2	3
ORL direct,#data	OR immediate data to direct byte	43	3	4
XRL A,Rn	Exclusive OR register to accumulator	68–6F	1	1
XRL A, direct	Exclusive OR direct byte to accumulator	65	2	2
XRL A,@Ri	Exclusive OR indirect RAM to accumulator	66–67	1	2
XRL A,#data	Exclusive OR immediate data to accumulator	64	2	2
XRL direct,A	Exclusive OR accumulator to direct byte	62	2	3

### Table 2-3. Instruction Set Summary

Mnemonic	Description	Hex Opcode	Bytes	Cycles
XRL direct,#data	Exclusive OR immediate data to direct byte	63	3	4
CLR A	Clear accumulator	E4	1	1
CPL A	Complement accumulator	F4	1	1
RL A	Rotate accumulator left	23	1	1
RLC A	Rotate accumulator left through carry	33	1	1
RR A	Rotate accumulator right	03	1	1
RRC A	Rotate accumulator right through carry	13	1	1
SWAP A	Swap nibbles within the accumulator	C4	1	1
	DATA TRANSFERS			
MOV A,Rn	Move register to accumulator	E8–EF	1	1
MOV A, direct	Move direct byte to accumulator	E5	2	2
MOV A,@Ri	Move indirect RAM to accumulator	E6–E7	1	2
MOV A,#data	Move immediate data to accumulator	74	2	2
MOV Rn,A	Move accumulator to register	F8–FF	1	2
MOV Rn,direct	Move direct byte to register	A8–AF	2	4
MOV Rn,#data	Move immediate data to register	78–7F	2	2
MOV direct,A	Move accumulator to direct byte	F5	2	3
MOV direct,Rn	Move register to direct byte	88–8F	2	3
MOV direct1, direct2	Move direct byte to direct byte	85	3	4
MOV direct,@Ri	Move indirect RAM to direct byte	86–87	2	4
MOV direct,#data	Move immediate data to direct byte	75	3	3
MOV @Ri,A	Move accumulator to indirect RAM	F6–F7	1	3
MOV @Ri,direct	Move direct byte to indirect RAM	A6–A7	2	5
MOV @Ri,#data	Move immediate data to indirect RAM	76–77	2	3
MOV DPTR,#data16	Load data pointer with a 16-bit constant	90	3	3
MOVC A,@A+DPTR	Move code byte relative to DPTR to accumulator	93	1	3
MOVC A,@A+PC	Move code byte relative to PC to accumulator	83	1	3
MOVX A,@Ri	Move external RAM (8-bit address) to A	E2–E3	1	3
MOVX A,@DPTR	Move external RAM (16-bit address) to A	E0	1	3
MOVX @Ri,A	Move A to external RAM (8-bit address)	F2–F3	1	4
MOVX @DPTR,A	Move A to external RAM (16-bit address)	F0	1	4
PUSH direct	Push direct byte onto stack	C0	2	4
POP direct	Pop direct byte from stack	D0	2	3
XCH A,Rn	Exchange register with accumulator	C8–CF	1	2
XCH A, direct	Exchange direct byte with accumulator	C5	2	3
XCH A,@Ri	Exchange indirect RAM with accumulator	C6–C7	1	3
XCHD A,@Ri	Exchange low-order nibble indirect. RAM with A	D6–D7	1	3
	PROGRAM BRANCHING			
ACALL addr11	Absolute subroutine call	xxx11	2	6
LCALL addr16	Long subroutine call	12	3	6
RET	Return from subroutine	22	1	4
RETI	Return from interrupt	32	1	4
AJMP addr11	Absolute jump	xxx01	2	3
LJMP addr16	Long jump	02	3	4
SJMP rel	Short jump (relative address)	80	2	3
JMP @A+DPTR	Jump indirect relative to the DPTR	73	1	2
JZ rel	Jump if accumulator is zero	60	2	3

# Table 2-3. Instruction Set Summary (continued)

Mnemonic	Description	Hex Opcode	Bytes	Cycles	
JNZ rel	Jump if accumulator is not zero	70	2	3	
JC rel	Jump if carry flag is set	40	2	3	
JNC	Jump if carry flag is not set	50	2	3	
JB bit,rel	Jump if direct bit is set	20	3	4	
JNB bit,rel	Jump if direct bit is not set	30	3	4	
JBC bit,direct rel	Jump if direct bit is set and clear bit	10	3	4	
CJNE A, direct rel	Compare direct byte to A and jump if not equal	B5	3	4	
CJNE A,#data rel	Compare immediate to A and jump if not equal	B4	3	4	
CJNE Rn,#data rel	Compare immediate to reg. and jump if not equal	B8–BF	3	4	
CJNE @Ri,#data rel	Compare immediate to indirect and jump if not equal	B6–B7	3	4	
DJNZ Rn,rel	Decrement register and jump if not zero	D8–DF	1	3	
DJNZ direct,rel	Decrement direct byte and jump if not zero	3	4		
NOP	00	1	1		
	Boolean VARIABLE OPERATIO	NS	1		
CLR C	Clear carry flag	C3	1	1	
CLR bit	Clear direct bit	C2	2	3	
SETB C	Set carry flag	D3	1	1	
SETB bit	Set direct bit	D2	2	3	
CPL C	Complement carry flag	B3	1	1	
CPL bit	Complement direct bit	B2	2	3	
ANL C,bit	AND direct bit to carry flag	82	2	2	
ANL C,/bit	AND complement of direct bit to carry	B0	2	2	
ORL C,bit	OR direct bit to carry flag	72	2	2	
ORL C,/bit	OR complement of direct bit to carry	A0	2	2	
MOV C,bit	Move direct bit to carry flag	A2	2	2	
MOV bit,C	Move carry flag to direct bit	92	2	3	

# Table 2-3. Instruction Set Summary (continued)

Instruction	CY	ov	AC
ADD	x	х	х
ADDC	x	х	х
SUBB	x	х	х
MUL	0	х	_
DIV	0	х	_
DA	x	-	-
RRC	x	_	_
RLC	x	-	_
SETB C	1	-	_
CLR C	x	-	-
CPLC	x	-	-
ANL C,bit	x	_	_
ANL C,/bit	x	-	_
ORL C,bit	x	-	-
ORL C,/bit	x	-	_
MOV C,bit	x	-	_
CJNE	х	-	_

Table 2-4. Instructions That Affect Flag Settings⁽¹⁾

⁽¹⁾ 0 = set to 0, 1 = set to 1, x = set to 0/1, - = not affected

### 2.5 Interrupts

The CPU has 18 interrupt sources. Each source has its own request flag located in a set of interrupt flag SFR registers. Each interrupt requested by the corresponding flag can be individually enabled or disabled. The definitions of the interrupt sources and the interrupt vectors are given in Table 2-5.

The interrupts are grouped into a set of priority-level groups with selectable priority levels.

The interrupt-enable registers are described in Section 2.5.1 and the interrupt priority settings are described in Section 2.5.3.

### 2.5.1 Interrupt Masking

Each interrupt can be individually enabled or disabled by the interrupt-enable bits in the interrupt-enable SFRs IEN0, IEN1, and IEN2. The CPU interrupt-enable SFRs are described as follows and summarized in Table 2-5.

Note that some peripherals have several events that can generate the interrupt request associated with that peripheral. This applies to Port 0, Port 1, Port 2, Timer 1, Timer 2, Timer 3, Timer 4, DMA controller, and Radio. These peripherals have interrupt mask bits for each internal interrupt source in the corresponding SFR or XREG registers.

In order to enable any of the interrupts, the following steps must be taken:

- 1. Clear interrupt flags.
- 2. Set individual interrupt-enable bit in the peripherals SFR register, if any.
- 3. Set the corresponding individual interrupt-enable bit in the IEN0, IEN1, or IEN2 register to 1.
- 4. Enable global interrupt by setting the EA bit in IEN0 to 1.
- 5. Begin the interrupt service routine at the corresponding vector address of that interrupt. See Table 2-5 for addresses.

Figure 2-4 gives a complete overview of all interrupt sources and associated control and state registers. Shaded boxes in Figure 2-4 are interrupt flags that are automatically cleared by hardware when the interrupt service routine is called. Indicates a one-shot, either due to the level source or due to edge shaping. Interrupts missing this are to be treated as level-triggered (apply to ports P0, P1, and P2). The



switch boxes are shown in the default state, and  $\Box$  or  $\Box$  indicates rising- or falling-edge detection, i.e., at what time instance the interrupt is generated. As a general rule for pulsed or edge-shaped interrupt sources, one should clear CPU interrupt flag registers prior to clearing the source flag bit, if available, for flags that are not automatically cleared. For level sources, one must clear the source prior to clearing the CPU flag.

Note that when clearing source interrupt flags in a register that contains several flags, interrupts may be lost if a read-modify-write operation is done (even in a single assembly instruction), as it will also clear interrupt flags that became active between the read and write operation. The source interrupt flags (with the exception of the USB controller interrupt flags) have the access mode R/W0. This means that writing 1 to a bit has no effect, so 1 should be written to an interrupt flag that is not to be cleared. For instance, to clear the TIMER2_OVF_PERF bit (bit 3) of T2IRQF in C code, one should do:

T2IRQF = ~(1 << 3); and not: T2IRQF &= ~(1 << 3); // wrong!

Interrupt Number	Description	Interrupt Name	Interrupt Vector	Interrupt Mask, CPU	Interrupt Flag, CPU
0	RF core error situation	RFERR	0x03	IEN0.RFERRIE	TCON.RFERRIF ⁽¹⁾
1	ADC end of conversion	ADC	0x0B	IEN0.ADCIE	TCON.ADCIF ⁽¹⁾
2	USART 0 RX complete	URX0	0x13	IEN0.URX0IE	TCON.URX0IF ⁽¹⁾
3	USART 1 RX complete	URX1	0x1B	IEN0.URX1IE	TCON.URX11F ⁽¹⁾
4	AES encryption/decryption complete	ENC	0x23	IEN0.ENCIE	SOCON.ENCIF
5	Sleep Timer compare	ST	0x2B	IEN0.STIE	IRCON.STIF
6	Port-2 inputs/USB/I ² C	P2INT	0x33	IEN2.P2IE	IRCON2.P2IF ⁽²⁾
7	USART 0 TX complete	UTX0	0x3B	IEN2.UTX0IE	IRCON2.UTX0IF
8	DMA transfer complete	DMA	0x43	IEN1.DMAIE	IRCON.DMAIF
9	Timer 1 (16-bit) capture/compare/overflow	T1	0x4B	IEN1.T1IE	IRCON.T1IF ⁽¹⁾ ⁽²⁾
10	Timer 2	T2	0x53	IEN1.T2IE	IRCON.T2IF ⁽¹⁾ ⁽²⁾
11	Timer 3 (8-bit) capture/compare/overflow	Т3	0x5B	IEN1.T3IE	IRCON.T3IF ^{(1) (2)}
12	Timer 4 (8-bit) capture/compare/overflow	T4	0x63	IEN1.T4IE	IRCON.T4IF ^{(1) (2)}
13	Port 0 inputs	POINT	0x6B	IEN1.P0IE	IRCON.P0IF ⁽²⁾
14	USART 1 TX complete	UTX1	0x73	IEN2.UTX1IE	IRCON2.UTX11F
15	Port 1 inputs	P1INT	0x7B	IEN2.P1IE	IRCON2.P1IF ⁽²⁾
16	RF general interrupts	RF	0x83	IEN2.RFIE	S1CON.RFIF ⁽²⁾
17	Watchdog overflow in timer mode	WDT	0x8B	IEN2.WDTIE	IRCON2.WDTIF

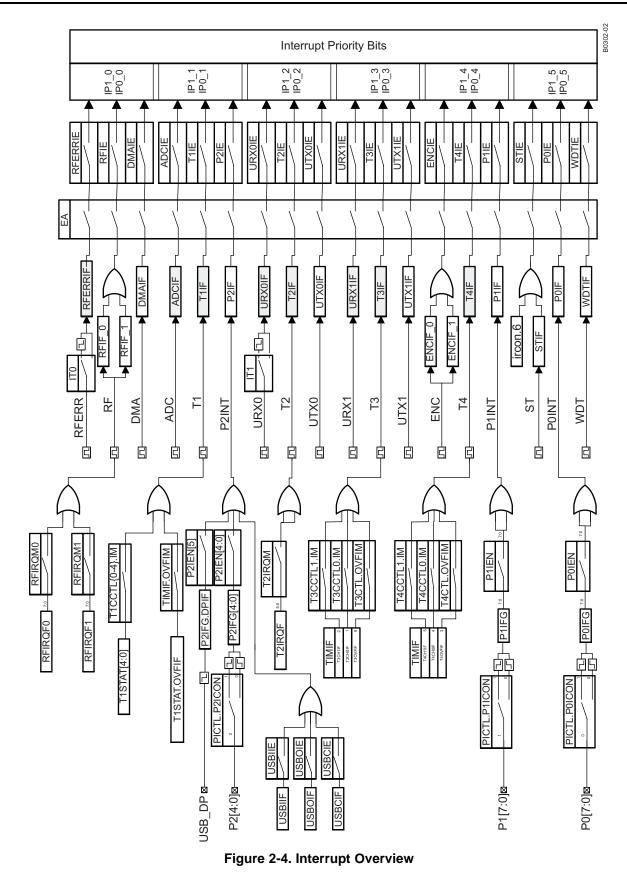
#### Table 2-5. Interrupts Overview

⁽¹⁾ Hardware-cleared when interrupt service routine is called

⁽²⁾ Additional IRQ mask and IRQ flag bits exist.



Interrupts



#### Interrupts

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Bit	Name	Reset	R/W	Description
7	EA	0	R/W	Disables all interrupts.
				0: No interrupt is acknowledged.
				1: Each interrupt source is individually enabled or disabled by setting its corresponding enable bit.
6	-	0	R0	Reserved. Read as 0
5	STIE	0	R/W	Sleep Timer interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
4	ENCIE	0	R/W	AES encryption/decryption interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
3	URX1IE	0	R/W	USART 1 RX interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
2	URXOIE	0	R/W	USART0 RX interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
1	ADCIE	0	R/W	ADC interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
0	RFERRIE	0	R/W	RF core error interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled

### IEN1 (0xB8) – Interrupt Enable 1

Bit	Name	Reset	R/W	Description
			-	•
7:6	-	00	R0	Reserved. Read as 0
5	POIE	0	R/W	Port 0 interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
4	T4IE	0	R/W	Timer 4 interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
3	T3IE	0	R/W	Timer 3 interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
2	T2IE	0	R/W	Timer 2 interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
1	T1IE	0	R/W	Timer 1 interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled
0	DMAIE	0	R/W	DMA transfer interrupt enable
				0: Interrupt disabled
				1: Interrupt enabled

IEN2	(0x9A) – Interrupt En	EN2 (0x9A) – Interrupt Enable 2					
Bit	Name	Reset	R/W	Description			
7:6	-	00	R0	Reserved. Read as 0			
5	WDTIE	0	R/W	Watchdog Timer interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			
4	P1IE	0	R/W	Port 1 interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			
3	UTX1IE	0	R/W	USART 1 TX interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			
2	UTXOIE	0	R/W	USART 0 TX interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			
1	P2IE	0	R/W	Port 2 and USB interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			
0	RFIE	0	R/W	RF general interrupt enable			
				0: Interrupt disabled			
				1: Interrupt enabled			

# 2.5.2 Interrupt Processing

When an interrupt occurs, the CPU vectors to the interrupt-vector address as shown in Table 2-5. Once an interrupt service has begun, it can be interrupted only by a higher-priority interrupt. The interrupt service is terminated by an RETI (return-from-interrupt instruction). When an RETI is performed, the CPU returns to the instruction that would have been next when the interrupt occurred.

When the interrupt condition occurs, the CPU also indicates this by setting an interrupt flag bit in the interrupt flag registers. This bit is set regardless of whether the interrupt is enabled or disabled. If the interrupt is enabled when an interrupt flag is set, then on the next instruction cycle, the interrupt is acknowledged by hardware, forcing an LCALL to the appropriate vector address.

Interrupt response requires a varying amount of time, depending on the state of the CPU when the interrupt occurs. If the CPU is performing an interrupt service with equal or greater priority, the new interrupt is pending until it becomes the interrupt with highest priority. In other cases, the response time depends on current instruction. The fastest possible response to an interrupt is seven machine cycles. This includes one machine cycle for detecting the interrupt and six cycles to perform the LCALL.

- **NOTE:** If an interrupt is disabled and the interrupt flag is polled, the 8051 assembly instruction JBC must not be used to poll the interrupt flag and clear it when set. If the JBC instruction is used, the interrupt flag may be re-asserted immediately.
- **NOTE:** If the assembly instruction XCH A, IEN0 is used to clear the global interrupt enable flag EA, the CPU may enter the interrupt routine on the cycle following this instruction. If that happens, the interrupt routine is executed with EA set to 0, which may delay the service of higher-priority interrupts.



### Interrupts

TCON (0x8	8) – Interrupt	Flags
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Bit	Name	Reset	R/W	Description
7	URX1IF	0	R/W H0	USART 1 RX interrupt flag. Set to 1 when USART 1 RX interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
6	-	0	R/W	Reserved
5	5 ADCIF	0	R/W H0	ADC interrupt flag. Set to 1 when ADC interrupt occurs and cleared when CPU vectors to the interrupt service routine.
		0: Interrupt not pending 1: Interrupt pending	0: Interrupt not pending	
				1: Interrupt pending
4	-	0	R/W	Reserved
3	URX0IF	0	R/W H0	USART 0 RX interrupt flag. Set to 1 when USART 0 interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1:Interrupt pending
2	IT1	1	R/W	Reserved. Must always be set to 1. Setting a zero enables low-level interrupt detection, which is almost always the case (one-shot when interrupt request is initiated).
1	RFERRIF	0	R/W H0	RF core error interrupt flag. Set to 1 when RFERR interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
0	ITO	1	R/W	Reserved. Must always be set to 1. Setting a zero enables low-level interrupt detection, which is almost always the case (one-shot when interrupt request is initiated).

#### S0CON (0x98) – Interrupt Flags 2

Bit	Name	Reset	R/W	Description
7:2	-	0000 00	R/W	Reserved
1	ENCIF_1	0	R/W	AES interrupt. ENC has two interrupt flags, ENCIF_1 and ENCIF_0. Setting one of these flags requests interrupt service. Both flags are set when the AES coprocessor requests the interrupt.
				0: Interrupt not pending
				1: Interrupt pending
0	ENCIF_0	0	R/W	AES interrupt. ENC has two interrupt flags, ENCIF_1 and ENCIF_0. Setting one of these flags requests interrupt service. Both flags are set when the AES coprocessor requests the interrupt.
				0: Interrupt not pending
				1: Interrupt pending

### S1CON (0x9B) – Interrupt Flags 3

Bit	Name	Reset	R/W	Description
7:2	-	0000 00	R/W	Reserved
1	RFIF_1	0	R/W	RF general interrupt. RF has two interrupt flags, RFIF_1 and RFIF_0. Setting one of these flags requests interrupt service. Both flags are set when the radio requests the interrupt.
				0: Interrupt not pending
				1: Interrupt pending
0	RFIF_0	0	R/W	RF general interrupt. RF has two interrupt flags, RFIF_1 and RFIF_0. Setting one of these flags requests interrupt service. Both flags are set when the radio requests the interrupt.
				0: Interrupt not pending
				1: Interrupt pending



IRCO	N (0xC0) – Interrupt	Flags 4		
Bit	Name	Reset	R/W	Description
7	STIF	0	R/W	Sleep Timer interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
6	-	0	R/W	Must be written 0. Writing a 1 always enables the interrupt source.
5	POIF	0	R/W	Port 0 interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
4	T4IF	0	R/W H0	Timer 4 interrupt flag. Set to 1 when Timer 4 interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
3	3 T3IF	0	R/W H0	Timer 3 interrupt flag. Set to 1 when Timer 3 interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
2	T2IF	0	R/W H0	Timer 2 interrupt flag. Set to 1 when Timer 2 interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
1	T1IF	0	R/W H0	Timer 1 interrupt flag. Set to 1 when Timer 1 interrupt occurs and cleared when CPU vectors to the interrupt service routine.
				0: Interrupt not pending
				1: Interrupt pending
0	DMAIF	0	R/W	DMA-complete interrupt flag
				0: Interrupt not pending
				1: Interrupt pending

#### IRCON2 (0xE8) – Interrupt Flags 5

Bit	Name	Reset	R/W	Description
7:5	-	000	R/W	Reserved
4	WDTIF	0	R/W	Watchdog Timer interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
3	P1IF	0	R/W	Port 1 interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
2	UTX1IF	0	R/W	USART 1 TX interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
1	UTX0IF	0	R/W	USART 0 TX interrupt flag
				0: Interrupt not pending
				1: Interrupt pending
0	P2IF	0	R/W	Port 2 interrupt flag
				0: Interrupt not pending
				1: Interrupt pending

# 2.5.3 Interrupt Priority

The interrupts are grouped into six interrupt priority groups, and the priority for each group is set by registers IP0 and IP1. In order to assign a higher priority to an interrupt, i.e., to its interrupt group, the corresponding bits in IP0 and IP1 must be set as shown in Table 2-6.

The interrupt priority groups with assigned interrupt sources are shown in Table 2-7. Each group is assigned one of four priority levels. While an interrupt service request is in progress, it cannot be interrupted by a lower- or same-level interrupt.



### Interrupts

In the case when interrupt requests of the same priority level are received simultaneously, the polling sequence shown in Table 2-8 is used to resolve the priority of each request. Note that the polling sequence in Figure 2-4 is the algorithm found in Table 2-8, not that polling is among the IP bits as listed in the figure.

#### IP1 (0xB9) – Interrupt Priority 1

Bit	Name	Reset	R/W	Description
7:6	-	00	R/W	Reserved
5	IP1_IPG5	0	R/W	Interrupt group 5, priority control bit 1, see Table 2-7: Interrupt Priority Groups
4	IP1_IPG4	0	R/W	Interrupt group 4, priority control bit 1, see Table 2-7: Interrupt Priority Groups
3	IP1_IPG3	0	R/W	Interrupt group 3, priority control bit 1, see Table 2-7: Interrupt Priority Groups
2	IP1_IPG2	0	R/W	Interrupt group 2, priority control bit 1, see Table 2-7: Interrupt Priority Groups
1	IP1_IPG1	0	R/W	Interrupt group 1, priority control bit 1, see Table 2-7: Interrupt Priority Groups
0	IP1_IPG0	0	R/W	Interrupt group 0, priority control bit 1, see Table 2-7: Interrupt Priority Groups

### IP0 (0xA9) – Interrupt Priority 0

Bit	Name	Reset	R/W	Description
7:6	-	00	R/W	Reserved
5	IP0_IPG5	0	R/W	Interrupt group 5, priority control bit 0, see Table 2-7: Interrupt Priority Groups
4	IP0_IPG4	0	R/W	Interrupt group 4, priority control bit 0, see Table 2-7: Interrupt Priority Groups
3	IP0_IPG3	0	R/W	Interrupt group 3, priority control bit 0, see Table 2-7: Interrupt Priority Groups
2	IP0_IPG2	0	R/W	Interrupt group 2, priority control bit 0, see Table 2-7: Interrupt Priority Groups
1	IP0_IPG1	0	R/W	Interrupt group 1, priority control bit 0, see Table 2-7: Interrupt Priority Groups
0	IP0_IPG0	0	R/W	Interrupt group 0, priority control bit 0, see Table 2-7: Interrupt Priority Groups

### Table 2-6. Priority Level Setting

IP1_x	IP0_x	Priority Level
0	0	0 – lowest
0	1	1
1	0	2
1	1	3 – highest

### Table 2-7. Interrupt Priority Groups

Group	Interrupts					
IPG0	RFERR	RF	DMA			
IPG1	ADC	T1	P2INT			
IPG2	URX0	T2	UTX0			
IPG3	URX1	Т3	UTX1			
IPG4	ENC	T4	P1INT			
IPG5	ST	POINT	WDT			



### Table 2-8. Interrupt Polling Sequence

Interrupt Number	Interrupt Name	
0	RFERR	
16	RF	
8	DMA	
1	ADC	
9	T1	
2	URX0	
10	T2	
3	URX1	
11	Т3	Polling sequence
4	ENC	↓
12	T4	
5	ST	
13	POINT	
6	P2INT	
7	UTX0	
14	UTX1	
15	P1INT	
17	WDT	



# **Debug Interface**

The two-wire debug interface allows programming of the on-chip flash, and it provides access to memory and register contents and debug features such as breakpoints, single-stepping, and register modification.

The debug interface uses I/O pins P2.1 and P2.2 as debug data and debug clock, respectively, during debug mode. These I/O pins can be used as general-purpose I/O only while the device is not in debug mode. Thus, the debug interface does not interfere with any peripheral I/O pins.

Горіс		Page
3.1	Debug Mode	<b>54</b>
3.2	Debug Communication	54
3.3	Debug Commands	56
3.4	Flash Programming	60
3.5	Debug Interface and Power Modes	60
3.6	Registers	62
	0	



Debug Mode

### 3.1 Debug Mode

Debug mode is entered by forcing two falling-edge transitions on pin P2.2 (debug clock) while the RESET_N input is held low. When RESET_N is set high, the device is in debug mode.

On entering debug mode, the CPU is in the halted state with the program counter reset to address 0x0000.

While in debug mode, pin P2.1 is the debug-data bidirectional pin, and P2.2 is the debug-clock input pin.

**NOTE:** Note that the debugger cannot be used with a divided system clock. When running the debugger, the value of CLKCONCMD.CLKSPD should be set to 000 when CLKCONCMD.OSC = 0 or to 001 when CLKCONCMD.OSC = 1.

### 3.2 Debug Communication

The debug interface uses a SPI-like two-wire interface consisting of the P2.1 (debug data) and P2.2 (debug clock) pins. Data is driven on the bidirectional debug-data pin at the positive edge of the debug clock, and data is sampled on the negative edge of this clock.

The direction of the debug-data pin depends on the command being issued. Data is driven on the positive edge of the debug clock and sampled on the negative edge. Figure 3-1 shows how data is sampled.

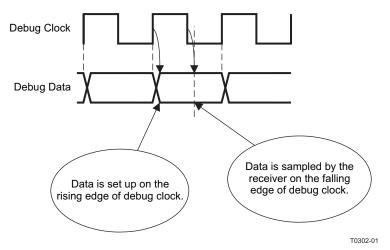
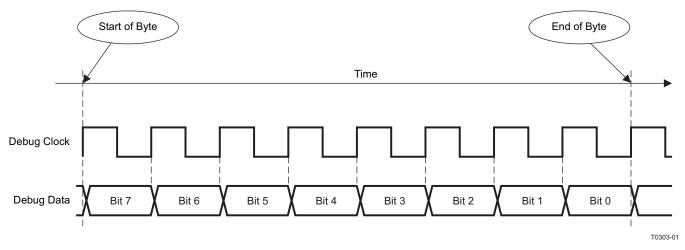


Figure 3-1. External Debug Interface Timing

The data is byte-oriented and is transmitted MSB-first. A sequence of one byte is shown in Figure 3-2.







A debug command sequence always starts with the host transmitting a command through the serial interface. This command encodes the number of bytes containing further parameters to follow, and whether a response is required. Based on this command, the debug module controls the direction of the debug data pad. A typical command sequence is shown in Figure 3-3. Note that the debug-data signal is simplified for the clarity of the figure, not showing each individual bit change. The direction is not explicitly indicated to the outside world, but must be derived by the host from the command protocol.

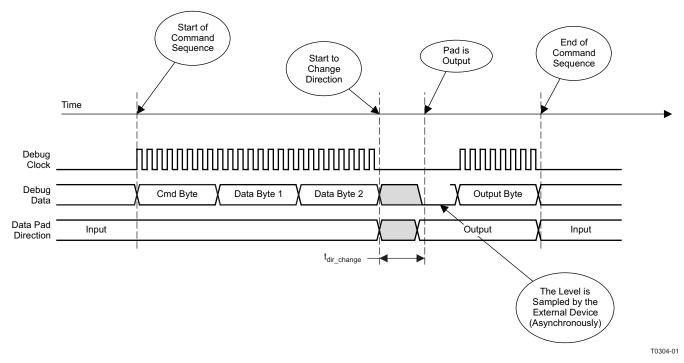


Figure 3-3. Typical Command Sequence—No Extra Wait for Response

For commands that require a response, there must be a small idle period between the command and the response to allow the pad to change direction. After the minimum waiting time  $(t_{dir_change})$ , the chip indicates whether it is ready to deliver the response data by pulling the data pad low. The external debugger, which is sampling the data pad, detects this and begins to clock out the response data. If the data pad is high after the waiting time, it is an indication to the debugger that the chip is not ready yet. Figure 3-4 shows how the wait works.



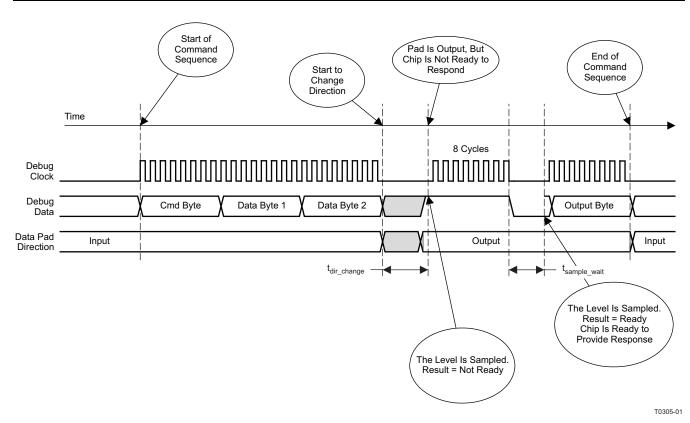


Figure 3-4. Typical Command Sequence. Wait for Response

If the debug interface indicates by pulling the data line high that it is not ready to return data, the external device must issue exactly eight clock pulses before it samples the ready level again. This must be repeated until the level is low. The wait cycle is equivalent to reading a byte from the debug interface, but ignoring the result. Note that the pad starts to change direction on the falling edge of the debug clock. Thus, the pad driver drives against the driver in the programmer until the programmer changes pad direction. This duration should be minimized in a programmer implementation.

### 3.3 Debug Commands

The debug commands are shown in Table 3-1. Some of the debug commands are described in further detail in the following subsections.

The 3 least-significant bits (the Xs) are don't care values.

Command	Instruction Byte	Additi onal Input Bytes	Output Bytes	Description
CHIP_ERASE	00010XXX	0	1	Perform flash chip erase (mass erase) and clear lock bits. If any other command except READ_STATUS is issued, then the use of CHIP_ERASE is disabled. Input byte: none Output byte: Debug status byte. See Table 3-3.
WR_CONFIG	00011XXX	1	1	Write debug configuration data. Input byte: See Table 3-2 for details. Output byte: Debug status byte. See Table 3-3.
RD_CONFIG	00100XXX	0	1	Read debug configuration data. Input byte: none. Output byte: Returns value set by WR_CONFIG command. See Table 3-2.

Table 3-1. Debug Commands

Command	Instruction Byte	Additi onal Input Bytes	Output Bytes	Description
GET_PC	00101XXX	0	2	Return value of 16-bit program counter. Input byte: none Output bytes: Returns 2 bytes.
READ_STATUS	00110XXX	0	1	Read status byte. Input byte: none Output byte: Debug status byte. See Table 3-3.
SET_HW_BRKPNT	00111XXX	3	1	Set hardware breakpoint. Input bytes: See Section 3.3.3 for details. Output byte: Debug status byte. See Table 3-3.
HALT	01000XXX	0	1	Halt CPU operation Input byte: none Output byte: Debug status byte. See Table 3-3. If the CPU was already halted, the output is undefined.
RESUME	01001XXX	0	1	Resume CPU operation. The CPU must be in the halted state for this command to be run. Input byte: none Output byte: Debug status byte. See Table 3-3.
DEBUG_INSTR	01010Хуу	1–3	1	Run debug instruction. The supplied instruction is executed by the CPU without incrementing the program counter. The CPU must be in halted state for this command to be run. Note that yy is number of bytes following the command byte, i.e., how many bytes the CPU instruction has (see Table 2-3). Input byte(s): CPU instruction Output byte: The resulting accumulator register value after the instruction has been executed
STEP_INSTR	01011XXX	0	1	Step CPU instruction. The CPU executes the next instruction from program memory and increments the program counter after execution. The CPU must be in the halted state for this command to be run. Input byte: none Output byte: The resulting accumulator register value after the instruction has been executed
GET_BM	01100XXX	0	1	This command does the same thing as GET_PC, except that it returns the memory bank. It returns one byte, where the 3 least-significant bits are the currently used memory bank. Input byte: none Output byte: Memory bank (current value of FMAP.MAP)
GET_CHIP_ID	01101XXX	0	2	Return value of 16-bit chip ID and version number. Input byte: none. Output bytes: The CHIPID and CHVER register values
BURST_WRITE	10000kkk	2–2049	1	This command writes a sequence of 1–2048 bytes to the DBGDATA register. Each time the register is updated, a DBG_BW DMA trigger is generated. The number of parameters to the BURST_WRITE command is variable. The number of data bytes in the burst is indicated using the 3 last bits of the command byte (kkk), and the whole next byte. The command sequence is shown in Figure 3-5. The burst length is indicated by an 11-bit value (b10–b0). After these two bytes, the given number of data bytes must be appended. The value 0 means 2048 data bytes; thus, the smallest number of bytes to transfer is 1. Input bytes: Command sequence Output byte: Debug status byte. See Table 3-3.



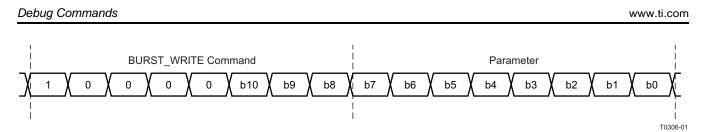


Figure 3-5. Burst Write Command (First 2 Bytes)

# 3.3.1 Debug Configuration

The commands WR_CONFIG and RD_CONFIG are used to access the debug-configuration data byte. The format and description of this configuration data are shown in Table 3-2.

Bit	Name	Reset	Description	
7:6	-	00	Reserved	
5	SOFT_POWER_MODE	1	When set, the digital regulator is not turned off during PM2 and PM3. If this bit is cleared, the debug interface is reset during PM2 and PM3.	
4	-	0	Reserved	
3	TIMERS_OFF	0	Disable timers. Disable timer operation. This overrides the <code>TIMER_SUSPEND</code> bit and its function.	
			0: Do not disable timers	
			1: Disable timers	
2	DMA_PAUSE	1	DMA pause. The DMA registers must not be accessed while this bit is set.	
			0: Enable DMA transfers	
			1: Pause all DMA transfers	
1	TIMER_SUSPEND	1	Suspend timers. Suspend timers when the chip is halted. The timers are also suspended during debug instructions. When executing a STEP, the timers receive exactly (or as close as possible) as many ticks as they would if the program were free-running.	
			0: Do not suspend timers	
			1: Suspend timers	
0	-	0	Reserved. Always write 0.	

### 3.3.2 Debug Status

A debug-status byte is read using the READ_STATUS command. The format and description of this debug status is shown in Table 3-3.

The READ_STATUS command is, for example, used for:

- Polling the status of the chip erase (CHIP_ERASE_BUSY) after a CHIP_ERASE command.
- Checking whether the oscillator is stable (OSCILLATOR_STABLE); required for debug commands HALT, RESUME, DEBUG_INSTR, STEP_REPLACE, and STEP_INSTR.

Bit	Name	Reset	Description
7	CHIP_ERASE_BUSY	0	Flash chip erase busy The signal is only high when a chip erase is in progress. It goes high immediately after a CHIP_ERASE command is received and returns to low when the flash is fully erased. 0: -
			1: Chip erase in progress

#### Table 3-3. Debug Status

Bit	Name	Reset	Description			
6	PCON_IDLE	0	PCON idle. See also Table 3-4.			
			0: CPU is running. Chip in operational mode controlled by debugger.			
			1: CPU is not running. Chip is in power mode defined by SLEEPCMD.MODE register setting. See Section 4.1 - Section 4.3 for details.			
5	CPU_HALTED	0	CPU was halted			
			0: CPU is running.			
			1: CPU was halted from a breakpoint or from a HALT debug command.			
4	PM_ACTIVE	0	Chip is active. Note that PM0 and PM1 are not supported in debug mode. See also Table 3-4.			
			0: Chip is in normal operation with CPU running (if not halted).			
			<ol> <li>Chip is out of normal operation (active mode) and either in transition up or down from power mode or stable in the power mode defined by SLEEPCMD.MODE register setting. See Section 4.1 - Section 4.3 for details.</li> </ol>			
3	HALT_STATUS	0	Halt status. Returns cause of last CPU halt			
			0: CPU was halted by HALT debug command			
			1: CPU was halted by hardware breakpoint			
2	DEBUG_LOCKED	0	Debug interface is locked. Returns value of DBGLOCK bit. See Section 3.4.1.			
			0: Debug interface is not locked.			
			1: Debug interface is locked.			
1	OSCILLATOR_STABLE	0	System clock oscillator stable.			
			0: Oscillators not stable			
			1: Oscillators stable			
0	STACK_OVERFLOW	0	Stack overflow. This bit indicates when the CPU writes to DATA memory space at address 0xFF, which is possibly a stack overflow.			
			0: No stack overflow			
			1: Stack overflow			

#### Table 3-4. Relation Between PCON_IDLE and PM_ACTIVE

PCON_IDLE	PM_ACTIVE	Description		
0	0	Chip in normal operation with CPU running (if not halted)		
0	1	Chip in transition to start-up from power mode		
1	0	Chip in transition to enter power mode		
1	1	Chip stable in power mode		

### 3.3.3 Hardware Breakpoints

The debug command SET_HW_BRKPNT is used to set one of the four available hardware breakpoints. When a hardware breakpoint is enabled, it compares the CPU address bus with the breakpoint. When a match occurs, the CPU is halted.

When issuing the SET_HW_BRKPNT, the external host must supply three data bytes that define the hardware breakpoint. The hardware breakpoint itself consists of 19 bits, whereas three bits are used for control purposes. The format of the three data bytes for the SET_HW_BRKPNT command is as follows.

The first data byte consists of the following:

- Bits 7-6: Unused
- Bits 5-4: Breakpoint number, 0-3
- Bit 3: 1 = enable, 0 = disable
- Bits 2–0: Memory bank bits. Bits 18–16 of hardware breakpoint.

The second data byte consists of bits 15-8 of the hardware breakpoint.



#### Flash Programming

The third data byte consists of bits 7–0 of the hardware breakpoint. Thus, the second and third data bytes set the CPU CODE address at which to stop execution.

### 3.4 Flash Programming

Programming of the on-chip flash is performed via the debug interface. The external host must initially send instructions using the DEBUG_INSTR debug command to perform the flash programming with the flash controller.

### 3.4.1 Lock Bits

For software and/or access protection, a set of lock bits can be written to the upper available flash page—the lock-bit page. The lock-bit structure consists of FLASH_PAGES-1 lock bits followed by one debug lock bit (see Table 3-5). The structure starts at address (flash-page size in bytes - 16) in the lock-bit page and occupies up to 16 bytes. The rest of the lock-bit page (addresses 0–(flash-page size in bytes - 17)) can be used to store code/constants, but cannot be changed without entering debug mode.

The PAGELOCK[FLASH_PAGES-2:0] lock-protect bits are used to enable erase/write protection for individual flash memory pages (2 KB). There is one bit for each available page.

When the debug-lock bit, DBGLOCK, is set to 0 (see Table 3-5), all debug commands except CHIP_ERASE, READ_STATUS, and GET_CHIP_ID are disabled. The status of the debug-lock bit can be read using the READ_STATUS command (see Section 3.3.2).

Note that after the debug-lock bit has changed due to a write to the lock-bit page or a CHIP_ERASE command, the device must be reset to lock/unlock the debug interface.

Issuing a CHIP_ERASE command is the only way to clear the debug-lock bit, thereby unlocking the debug interface.

Table 3-5 defines the 16-byte structure containing the flash lock-protection bits. Bit 0 of the first byte contains the lock bit for page 0, bit 1 of the first byte contains the lock bit for page 1, and so on. Bit 7 of the last byte in the flash is the DBGLOCK bit (bit 127 in the structure).

Bit	Name	Description		
127	DBGLOCK	Debug-lock bit 0: Disable debug commands 1: Enable debug commands		
FLASH_PAGES-2:0	PAGELOCK[FLASH_PAGES-2:0]	Page-lock bits. There is one bit for each of the up to 128 pages. Page-lock bits for unavailable pages are not used. 0: Page locked 1: Page not locked		

 Table 3-5. Flash Lock-Protection Bit Structure Definition

### 3.5 Debug Interface and Power Modes

Power modes PM2 and PM3 may be handled in two different ways when the chip is in debug mode. The default behavior is never to turn off the digital voltage regulator. This emulates power modes while maintaining debug mode operation. The clock sources are turned off as in ordinary power modes. The other option is to turn off the 1.8-V internal digital power. This leads to a complete shutdown of the digital part, which disables debug mode. When the chip is in debug mode, the two options are controlled by configuration bit 5 (SOFT_POWER_MODE).

The debug interface still responds to a reduced set of commands while in one of the power modes. The chip can be woken up from sleep mode by issuing a HALT command to the debug interface. The HALT command brings the chip up from sleep mode in the halted state. The RESUME command must be issued to resume software execution.

The debug status may be read when in power modes. The status must be checked when leaving a power mode by issuing a HALT command. The time needed to power up depends on which power mode the chip is in, and must be checked in the debug status. The debug interface only accepts commands that are available in sleep mode before the chip is operational.



**NOTE:** Debugging in Idle mode and PM1 is not supported. It is recommended to use active mode or another power mode when debugging.

# 3.6 Registers

#### DBGDATA (0x6260) – Debug Data

Bit	Name	Reset	R/W	Description
7:0	BYTE[7:0]	0	R	Debug data from BURST_WRITE command This register is updated each time a new byte has been transferred to the debug interface using the BURST_WRITE command. A DBG_BW DMA trigger is generated when this byte is updated. This allows the DMA controller to fetch the data.

### CHVER (0x6249) – Chip Version

Bit	Name	Reset	R/W	Description
7:0	VERSION[7:0]	Chip dependent	R	Chip revision number

#### CHIPID (0x624A) - Chip ID

Bit	Name	Reset	R/W	Description
7:0	CHIPID[7:0]	Chip dependent	R	Chip identification number. CC2530: 0xA5 CC2531: 0xB5 CC2533: 0x95 CC2540: 0x8D CC2541: 0x41

### CHIPINFO0 (0x6276) - Chip Information Byte 0

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Always 0.
6:4	FLASHSIZE[2:0]	Chip dependent	R	Flash Size. 001 – 32 KB, 010 – 64 KB, 011 – 128 KB (for CC2533: 011 – 96 KB), 100 – 256 KB
3	USB	Chip dependent	R	1 if chip has USB, 0 otherwise
2	-	1	R1	Reserved. Always 1
1:0	-	00	R0	Reserved. Always 00

#### CHIPINFO1 (0x6277) - Chip Information Byte 1

Bit	Name	Reset	R/W	Description	
7:3	_	Chip dependent	R	Reserved.	
2:0	SRAMSIZE[2:0]	Chip dependent	R	SRAM size in KB minus 1. For example, a 4-KB device has this field set to 011. Add 1 to the number to get the number of KB available.	



# **Power Management and Clocks**

Low-power operation is enabled through different operating modes (power modes). The various operating modes are referred to as active mode, idle mode, and power modes 1, 2, and 3 (PM1–PM3). Ultralow-power operation is obtained by turning off the power supply to modules to avoid static (leakage) power consumption and also by using clock gating and turning off oscillators to reduce dynamic power consumption.

#### Topic

### Page

4.1	Power Management Introduction	64
4.2	Power-Management Control	65
4.3	Power-Management Registers	66
4.4	Oscillators and Clocks	69
4.5	Timer Tick Generation	72
4.6	Data Retention	72

### 4.1 **Power Management Introduction**

Different operating modes, or power modes, are used to allow low-power operation. Ultralow-power operation is obtained by turning off the power supply to modules to avoid static (leakage) power consumption and also by using clock gating and turning off oscillators to reduce dynamic power consumption.

The five various operating modes (power modes) are called active mode, idle mode, PM1, PM2, and PM3 (PM1/PM2/PM3 are also referred to as sleep modes). Active mode is the normal operating mode, whereas PM3 has the lowest power consumption. The impact of the different power modes on system operation is shown in Table 4-1, together with voltage regulator and oscillator options.

Power Mode	High-I	Frequency Oscillator	Low-Fre	Low-Frequency Oscillator		
Configuration	А	32-MHz XOSC	С	32-kHz XOSC		
Configuration	В	16-MHz RCOSC	D	32-kHz RCOSC		
Active / idle mode	A or B		C or D		ON	
PM1	None		C or D		ON	
PM2	None		C or D		OFF	
PM3	None		None		OFF	

### Table 4-1. Power Modes

Active mode: The fully functional mode. The voltage regulator to the digital core is on, and either the 16-MHz RC oscillator or the 32-MHz crystal oscillator or both are running. Either the 32-kHz RCOSC or the 32-kHz XOSC is running.

Idle mode: Identical to active mode, except that the CPU core stops operating (is idle).

**PM1:** The voltage regulator to the digital part is on. Neither the 32-MHz XOSC nor the 16-MHz RCOSC is running. Either the 32-kHz RCOSC or the 32-kHz XOSC is running. The system goes to active mode on reset, an external interrupt, or when the Sleep Timer expires.

**PM2:** The voltage regulator to the digital core is turned off. Neither the 32-MHz XOSC nor the 16-MHz RCOSC is running. Either the 32-kHz RCOSC or the 32-kHz XOSC is running. The system goes to active mode on reset, an external interrupt, or when the Sleep Timer expires.

**PM3**: The voltage regulator to the digital core is turned off. None of the oscillators is running. The system goes to active mode on reset or an external interrupt.

The POR is active in PM2/PM3, but the BOD is powered down, which gives a limited voltage supervision. If the supply voltage is lowered to below 1.4 V during PM2/PM3, at temperatures of 70°C or higher, and then brought back up to good operating voltage before active mode is re-entered, registers and RAM contents that are saved in PM2/PM3 may become altered. Hence, care should be taken in the design of the system power supply to ensure that this does not occur. The voltage can be periodically supervised accurately by entering active mode, as a BOD reset is triggered if the supply voltage is below approximately 1.7 V.

The CC2533 and CC2541 have functionality to perform automatically a CRC check of the retained configuration register values in PM2/PM3 to check that the device state was not altered during sleep. The bits in SRCRC.CRC_RESULT indicate whether there were any changes, and by enabling SRCRC.CRC_RESET_EN, the device immediately resets itself with a watchdog reset if SRCRC.CRC_RESULT is not 00 (= CRC of retained registers passed) after wakeup from PM2/PM3. The SRCRC register also contains the SRCRC.FORCE_RESET bit that can be used by software to immediately trigger a watchdog reset to reboot the device.

For CC2533 and CC2541, additional analog reset architecture adds another brownout detector (the 3VBOD) that senses on the unregulated voltage. The purpose of this 3VBOD is to reduce the current consumption of the device when supplied with voltages well below the operating voltage.



### 4.1.1 Active and Idle Mode

Active mode is the fully functional mode of operation where the CPU, peripherals, and RF transceiver are active. The digital voltage regulator is turned on.

Active mode is used for normal operation. By enabling the PCON. IDLE bit while in active mode (SLEEPCMD.MODE = 0x00), the CPU core stops operating and the idle mode is entered. All other peripherals function normally, and any enabled interrupt wakes up the CPU core (to transition back from idle mode to active mode).

### 4.1.2 PM1

In PM1, the high-frequency oscillators are powered down (32-MHz XOSC and 16-MHz RCOSC). The voltage regulator and the enabled 32-kHz oscillator are on. When PM1 is entered, a power-down sequence is run.

PM1 is used when the expected time until a wakeup event is relatively short (less than 3 ms), because PM1 uses a fast power-down/up sequence.

### 4.1.3 PM2

PM2 has the second-lowest power consumption. In PM2, the power-on reset, external interrupts, selected 32-kHz oscillator, and Sleep Timer peripherals are active. I/O pins retain the I/O mode and output value set before entering PM2. All other internal circuits are powered down. The voltage regulator is also turned off. When PM2 is entered, a power-down sequence is run.

PM2 is typically entered when using the Sleep Timer as the wakeup event, and also combined with external interrupts. PM2 should typically be choosen, compared to PM1, when expected sleep time exceeds 3 ms. Using less sleep time does not reduce system power consumption compared to using PM1.

### 4.1.4 PM3

PM3 is used to achieve the operating mode with the lowest power consumption. In PM3, all internal circuits that are powered from the voltage regulator are turned off (basically all digital modules; the only exceptions are interrupt detection and POR level sensing). The internal voltage regulator and all oscillators are also turned off.

Reset (POR or external) and external I/O port interrupts are the only functions that operate in this mode. I/O pins retain the I/O mode and output value set before entering PM3. A reset condition or an enabled external I/O interrupt event wakes the device up and places it into active mode (an external interrupt starts from where it entered PM3, whereas a reset returns to start-of-program execution). The content of RAM and registers is partially preserved in this mode (see Section 4.6). PM3 uses the same power-down/up sequence as PM2.

PM3 is used to achieve ultralow power consumption when waiting for an external event. It should be used when expected sleep time exceeds 3 ms.

### 4.2 Power-Management Control

The required power mode is selected by the MODE bits in the SLEEPCMD control register and the PCON.IDLE bit. Setting the SFR register PCON.IDLE bit enters the mode selected by SLEEPCMD.MODE.

An enabled interrupt from port pins or Sleep Timer or a power-on reset wakes the device from other power modes and brings it into active mode.

When PM1, PM2, or PM3 is entered, a power-down sequence is run. When the device is taken out of PM1, PM2, or PM3, it starts at 16 MHz and automatically changes to 32 MHz if CLKCONCMD.OSC was 0 when entering the power mode (setting PCON.IDLE). If CLKCONCMD.OSC was 1 when PCON.IDLE was set, when entering the power mode, it continues to run at 16 MHz.



#### Power-Management Registers

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The instruction that sets the PCON. IDLE bit must be aligned in a certain way for correct operation. The first byte of the assembly instruction immediately following this instruction must not be placed on a 4-byte boundary. Furthermore, cache must not be disabled (see CM in the FCTL register description in Chapter 6). Failure to comply with this requirement may cause higher current consumption. Provided this requirement is fulfilled, the first assembly instruction after the instruction that sets the PCON. IDLE bit is performed before the ISR of the interrupt that caused the system to wake up, but after the system woke up. If this instruction is a global interrupt disable, it is possible to have it followed by code for execution after wakeup, but before the ISR is serviced.

An example of how this can be done in the IAR compiler is shown as follows. The command for setting PCON to 1 is placed in a function written in assembly code. In a C file calling this function, a declaration such as extern void EnterSleepModeDisableInterruptsOnWakeup(void); is used. The RSEG NEAR_CODE:CODE:NOROOT(2) statement ensures that the MOV PCON, #1 instruction is placed on a 2-byte boundary. It is a 3-byte instruction, so the following instruction is not placed on a 4-byte boundary, as required. In the following example, this instruction is CLR EA, which disables all interrupts. That means that the ISR of the interrupt that woke up the system is not executed until after the IENO.EA bit has been set again later in the code. If this functionality is not wanted, the CLR EA instruction can be replaced by a NOP.

PUBLIC EnterSleepModeDisableInterruptsOnWakeup FUNCTION EnterSleepModeDisableInterruptsOnWakeup,0201H RSEG NEAR_CODE:CODE:NOROOT(2) EnterSleepModeDisableInterruptsOnWakeup: MOV PCON,#1 CLR EA RET

#### 4.3 **Power-Management Registers**

This section describes the power-management registers. All register bits retain their previous values when entering PM2 or PM3.

Bit	Name	Reset	R/W	Description
7	XOSC_AMP_DET_EN	0	R/W	0: Disable
				1: Enable the amplitude detector for the 32-MHz XOSC, CC2533 only
6	-	0	R0	Reserved. Always read 0.
5	FORCE_RESET	0	R/W	0: No action
				1: Force watchdog reset.
4	-	0	R	Reserved
3:2	CRC_RESULT	00	R/W0	00: CRC of retained registers passed
				01: Low CRC value failed
				10: High CRC value failed
				11: Both CRC values failed
1	-	0	R	Reserved
0	CRC_RESET_EN	0	R/W	0: Disable reset of chip due to CRC.
				1: Enable reset of chip if CRC_RESULT != 00 after wakeup from PM2/PM3.

#### SRCRC (0x6262) – Sleep Reset CRC (CC2533 and CC2541 only)

#### PCON (0x87) – Power Mode Control

Bit	Name	Reset	R/W	Description
7:1	-	000 000	R/W	Reserved, always write as 0000 000.
0	IDLE	0	R0/W H0	Power mode control. Writing 1 to this bit forces the device to enter the power mode set by SLEEPCMD.MODE (note that MODE = 0x00 AND IDLE = 1 stops the CPU core activity). This bit is always read as 0.
				All enabled interrupts clear this bit when active, and the device re-enters active mode.



Bit	Name	Reset	R/W	Description
7	OSC32K_CALDIS	0	R/W	Disable 32-kHz RC oscillator calibration
				0: 32-kHz RC oscillator calibration is enabled.
				1: 32-kHz RC oscillator calibration is disabled.
				This setting can be written at any time, but does not take effect before the chip has been running on the 16-MHz high-frequency RC oscillator.
6:3	-	000 0	R0	Reserved
2	-	1	R/W	Reserved. Always write as 1
1:0	MODE[1:0]	00	R/W	Power-mode setting
				00: Active / Idle mode
				01: Power mode 1 (PM1)
				10: Power mode 2 (PM2)
				11: Power mode 3 (PM3)

#### SLEEPSTA (0x9D) – Sleep-Mode Control Status

Bit	Name	Reset	R/W	Description
7	OSC32K_CALDIS	0	R	32-kHz RC oscillator calibration status SLEEPSTA.OSC32K_CALDIS shows the current status of disabling of the 32-kHz RC calibration. The bit is not set to the same value as SLEEPCMD.OSC32K_CALDIS before the chip has been run on the 32-kHz RC oscillator.
6:5	-	00	R	Reserved
4:3	RST[1:0]	XX	R	Status bit indicating the cause of the last reset. If there are multiple resets, the register only contains the last event.
				00: Power-on reset and brownout detection
				01: External reset
				10: Watchdog Timer reset
				11: Clock loss reset
2:1	-	00	R	Reserved
0	CLK32K	0	R	The 32-kHz clock signal (synchronized to the system clock)



Power-Management Registers

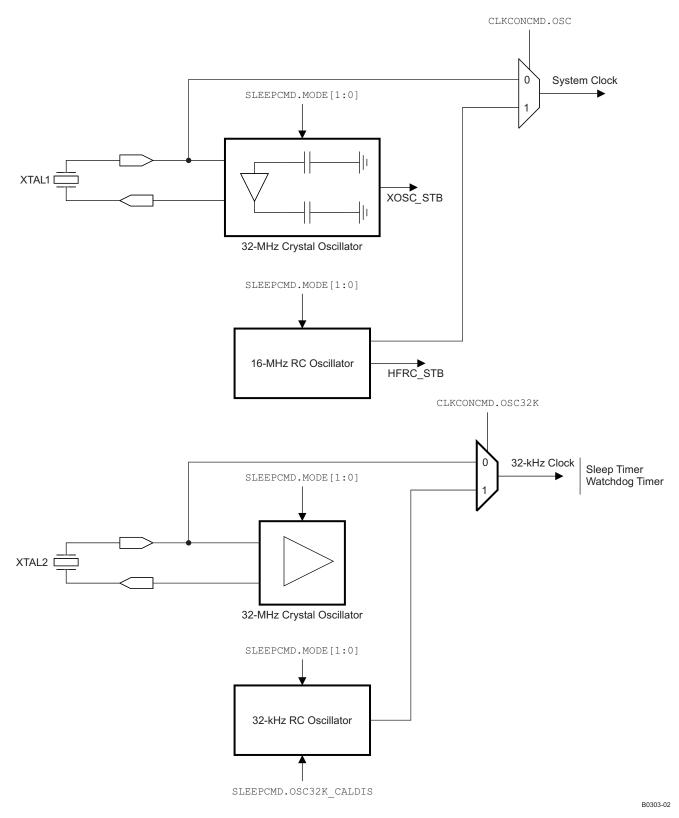


Figure 4-1. Clock System Overview



### 4.4 Oscillators and Clocks

The device has one internal system clock, or main clock. The source for the system clock can be either the 16-MHz RC oscillator or the 32-MHz crystal oscillator. Clock control is performed using the CLKCONCMD SFR register.

There is also one 32-kHz clock source that can either be an RC oscillator or a crystal oscillator, also controlled by the CLKCONCMD register.

The CLKCONSTA register is a read-only register used for getting the current clock status.

The choice of oscillator allows a trade-off between high accuracy in the case of the crystal oscillator and low power consumption when the RC oscillator is used. Note that operation of the RF transceiver requires that the 32-MHz crystal oscillator is used.

In the CC2533, CC2540 and CC2541, an additional module for detection of 32-MHz XOSC stability is available. This amplitude detector can be useful in environments with significant noise on the power supply, to ensure that the clock source is not used until the clock signal is stable. In the CC2533, this module can be enabled by setting the SRCRC.XOSC_AMP_DET_EN bit, and this adds around 20 µs to the 32-MHz XOSC startup time. In the CC2540/41, the module is always enabled.

### 4.4.1 Oscillators

Figure 4-1 gives an overview of the clock system with available clock sources.

Two high-frequency oscillators are present in the device:

- 32-MHz crystal oscillator
- 16-MHz RC oscillator

The 32-MHz crystal-oscillator start-up time may be too long for some applications; therefore, the device can run on the 16-MHz RC oscillator until the crystal oscillator is stable. The 16-MHz RC oscillator consumes less power than the crystal oscillator, but because it is not as accurate as the crystal oscillator it cannot be used for RF transceiver operation.

Two low-frequency oscillators are present in the device:

- 32-kHz crystal oscillator
- 32-kHz RC oscillator.

The 32-kHz XOSC is designed to operate at 32.768 kHz and provide a stable clock signal for systems requiring time accuracy. The 32-kHz RCOSC runs at 32.753 kHz when calibrated. The calibration can only take place when the 32-MHz XOSC is enabled, and this calibration can be disabled by enabling the SLEEPCMD.OSC32K_CALDIS bit. The 32-kHz RCOSC should be used to reduce cost and power consumption compared to the 32-kHz XOSC solution. The two 32-kHz oscillators cannot be operated simultaneously.

### 4.4.2 System Clock

The system clock is derived from the selected system clock source, which is the 32-MHz XOSC or the 16-MHz RCOSC. The CLKCONCMD.OSC bit selects the source of the system clock. Note that to use the RF transceiver, the 32-MHz crystal oscillator must be selected and stable.

Note that changing the CLKCONCMD.OSC bit does not cause the system clock to change instantly. The clock source change first takes effect when CLKCONSTA.OSC = CLKCONCMD.OSC. This is due to the requirement to have stable clocks prior to actually changing the clock source. Also note that the CLKCONCMD.CLKSPD bit reflects the frequency of the system clock and thus is a mirror of the CLKCONCMD.OSC bit.

The 16 MHz RC oscillator is calibrated once after the 32-MHz XOSC has been selected and is stable, i.e,. when the CLKCONSTA.OSC bit switches from 1 to 0.

- **NOTE:** The change from the 16-MHz clock source to the 32-MHz clock source (and vice versa) aligns with the CLKCONCMD.TICKSPD setting. A slow CLKCONCMD.TICKSPD setting when CLKCONCMD.OSC is changed results in a longer time before the actual source change takes effect. The fastest switching is obtained when CLKCONCMD.TICKSPD equals 000.
- **NOTE:** After coming up from PM1, PM2, or PM3, the CPU must wait for CLKCONSTA.OSC to be 0 before operations requiring the system to run on the 32-MHz XOSC (such as the radio) are started.

# 4.4.3 32 kHz Oscillators

Two 32 kHz oscillators are present in the device as clock sources for the 32-kHz clock:

- 32-kHz XOSC
- 32-kHz RCOSC

By default, after a reset, the 32-kHz RCOSC is enabled and selected as the 32-kHz clock source. The RCOSC consumes less power, but is less accurate compared to the 32-kHz XOSC. The chosen 32-kHz clock source drives the Sleep Timer, generates the tick for the Watchdog Timer, and is used as a strobe in Timer 2 to calculate the Sleep Timer sleep time. The CLKCONCMD.OSC32K register bit selects the oscillator to be used as the 32-kHz clock source.

The CLKCONCMD.OSC32K register bit can be written at any time, but does not take effect before the 16-MHz RCOSC is the active system clock source. When system clock is changed from the 16-MHz RCOSC to the 32-MHz XOSC (CLKCONCMD.OSC from 1 to 0), calibration of the 32-kHz RCOSC starts up and is performed once if the 32-kHz RCOSC is selected. During calibration, a divided version of the 32-MHz XOSC is used. The result of the calibration is that the 32-kHz RSOSC is running at 32.753 kHz. The 32-kHz RCOSC calibration may take up to 2 ms to complete. Calibration can be disabled by setting SLEEPCMD.OSC32K_CALDIS to 1. At the end of the calibration, an extra pulse may occur on the 32-kHz clock source, which causes the sleep timer to be incremented by 1.

Note that after having switched to the 32-kHz XOSC and when coming up from PM3 with the 32-kHz XOSC enabled, the oscillator requires up to 500 ms to stabilize on the correct frequency. The Sleep Timer, Watchdog Timer and clock-loss detector should not be used before the 32-kHz XOSC is stable.

### 4.4.4 Oscillator and Clock Registers

This section describes the oscillator and clock registers. All register bits retain their previous values when entering PM2 or PM3.



Bit	Name	Reset	R/W	Description
7	OSC32K	1	R/W	32 kHz clock-source select. Setting this bit initiates a clock-source change only. CLKCONSTA.OSC32K reflects the current setting. The 16 MHz RCOSC must be selected as system clock when this bit is to be changed.
				0: 32 kHz XOSC
				1: 32 kHz RCOSC
6	OSC	1	R/W	System clock-source select. Setting this bit initiates a clock-source change only. CLKCONSTA.OSC reflects the current setting.
				0: 32 MHz XOSC
				1: 16 MHz RCOSC
5:3	TICKSPD[2:0]	001	R/W	Timer ticks output setting. Cannot be higher than system clock setting given by OSC bit setting.
				000: 32 MHz
				001: 16 MHz
				010: 8 MHz
				011: 4 MHz
				100: 2 MHz
				101: 1 MHz
				110: 500 kHz
				111: 250 kHz
				Note that CLKCONCMD.TICKSPD can be set to any value, but the effect is limited by the CLKCONCMD.OSC setting; i.e., if CLKCONCMD.OSC = 1 and CLKCONCMD.TICKSPD = 000, CLKCONSTA.TICKSPD reads 001, and the
				real TICKSPD is 16 MHz.
2:0	CLKSPD	001	R/W	Clock speed. Cannot be higher than system clock setting given by the OSC bit setting. Indicates current system-clock frequency
				000: 32 MHz
				001: 16 MHz
				010: 8 MHz
				011: 4 MHz
				100: 2 MHz
				101: 1 MHz
				110: 500 kHz
				111: 250 kHz
				Note that CLKCONCMD. CLKSPD can be set to any value, but the effect is         limited by the       CLKCONCMD.OSC setting; i.e., if         CLKCONCMD.OSC = 1 and CLKCONCMD.CLKSPD = 000,         CLKCONSTA.CLKSPD reads 001, and the real CLKSPD is 16 MHz.
				Note also that the debugger cannot be used with a divided system clock. When running the debugger, the value of CLKCONCMD.CLKSPD should be set to 000 when CLKCONCMD.OSC = 0 or to 001 when CLKCONCMD.OSC = 1.

CLKC	CLKCONSTA (0x9E) – Clock Control Status					
Bit	Name	Reset	R/W	Description		
7	OSC32K	1	R	Current 32-kHz clock source selected:		
				0: 32-kHz XOSC		
				1: 32-kHz RCOSC		
6	OSC	1	R	Current system clock selected:		
				0: 32-MHz XOSC		
				1: 16-MHz RCOSC		
5:3	TICKSPD[2:0]	001	R	Current timer ticks output setting		
				000: 32 MHz		
				001: 16 MHz		
				010: 8 MHz		
				011: 4 MHz		
				100: 2 MHz		
				101: 1 MHz		
				110: 500 kHz		
				111: 250 kHz		
2:0	CLKSPD	001	R	Current clock speed		
				000: 32 MHz		
				001: 16 MHz		
				010: 8 MHz		
				011: 4 MHz		
				100: 2 MHz		
				101: 1 MHz		
				110: 500 kHz		
				111: 250 kHz		

### 4.5 Timer Tick Generation

The value of the CLKCONCMD.TICKSPD register controls a global prescaler for Timer 1, Timer 3, and Timer 4. The prescaler value can be set to a value from 0.25 MHz to 32 MHz. It should be noted that if CLKCONCMD.TICKSPD indicates a higher frequency than the system clock, the actual prescaler value indicated in CLKCONSTA.TICKSPD is the same as the system clock.

# 4.6 Data Retention

In power modes PM2 and PM3, power is removed from most of the internal circuitry. However, SRAM retains its contents, and the content of internal registers is also retained in PM2 and PM3.

All CPU, RF, and peripheral registers retain their contents in PM2 and PM3, except the AES, I²C, and USB registers, OBSSEL0–OBSSEL5, TR0, and in the CC2541, LLECTRL.

Switching to the PM2 or PM3 low-power modes appears transparent to software. Note that the value of the Sleep Timer is not preserved in PM3.

All registers retain their values in PM1.



Chapter 5 SWRU191C–April 2009–Revised January 2012

# Reset

The device has five reset sources. The following events generate a reset:

- Forcing the RESET_N input pin low
- A power-on reset condition
- A brownout reset condition
- Watchdog Timer reset condition
- Clock-loss reset condition

The initial conditions after a reset are as follows:

- I/O pins are configured as inputs with pullups (P1.0 and P1.1 are inputs, but do not have pullup/pulldown)
- · CPU program counter is loaded with 0x0000 and program execution starts at this address
- · All peripheral registers are initialized to their reset values (see register descriptions)
- Watchdog Timer is disabled
- · Clock-loss detetector is disabled

During reset, the I/O pins are configured as inputs with pullups (P1.0 and P1.1 are inputs, but do not have pullup/pulldown).

In the CC2533 and CC2541, a watchdog reset can be generated immediately in software by writing the SRCRC.FORCE_RESET bit to 1 (see Section 4.3 for the register description). In the other devices in the family, a watchdog reset can be triggered from software by enabling the watchdog timer with the shortest time-out and waiting for it to trigger.

#### Topic

#### Page

5.1	Power-On Reset and Brownout Detector	74
5.2	Clock-Loss Detector	74



#### 5.1 Power-On Reset and Brownout Detector

The device includes a power-on reset (POR), providing correct initialization during device power on. It also includes a brownout detector (BOD) operating on the regulated 1.8-V digital power supply only. The BOD protects the memory contents during supply voltage variations which cause the regulated 1.8-V power to drop below the minimum level required by digital logic, flash memory, and SRAM.

When power is initially applied, the POR and BOD hold the device in the reset state until the supply voltage rises above the power-on-reset and brownout voltages.

The cause of the last reset can be read from the register bits SLEEPSTA.RST. It should be noted that a BOD reset is read as a POR reset.

#### 5.2 Clock-Loss Detector

The clock-loss detector can be used in safety-critical systems to detect that one of the XOSC clock sources (32-MHz XOSC or 32-kHz XOSC) has stopped. This can typically happen due to damage to the external crystal or supporting components. When the clock-loss detector is enabled, the two clocks monitor each other continously. If one of the clocks stops toggling, a clock-loss detector reset is generated within a certain maximum time-out period. The time-out depends on which clock stops. If the 32-kHz clock stops, the time-out period is 0.5 ms. If the 32-MHz clock stops, the time-out period is 0.25 ms. When the system comes up again from reset, software can detect the cause of the reset by reading SLEEPSTA.RST[1:0]. After a reset, the internal RC oscillators are used. Thus, the system is able to start up again and can then be powered down gracefully. The clock-loss detector is enabled/disabled with the CLD. EN bit. It is assumed that the 32-MHz XOSC is selected as system clock source when using the clock-loss detector. The 32-kHz clock can be 32-kHz RCOSC (should be calibrated for accurate reset timeout) or 32-kHz XOSC.

In power modes 1 and 2, the clock-loss detector is automatically stopped and restarted when the clocks start up again.

Before entering power mode 3, switch to the 16-MHz RCOSC and disable the clock-loss detector. When entering active mode again, turn on the clock-loss detector and then switch back to the 32-MHz XOSC.

Bit	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved
0	EN	0	R/W	Clock-loss detector enable

#### CLD (0x6290) - Clock-Loss Detection



# Flash Controller

The device contains flash memory for storage of program code. The flash memory is programmable from the user software and through the debug interface.

The flash controller handles writing and erasing the embedded flash memory. The embedded flash memory consists of up to 128 pages of 2048 bytes (CC2530/CC2531/CC2540/CC2541) or 1024 bytes (CC2533) each.

The flash controller has the following features:

- 32-bit word programmable
- Page erase
- · Lock bits for write protection and code security
- Flash-page erase timing 20 ms
- · Flash-chip erase timing 20 ms
- Flash-write timing (4 bytes) 20 µs

#### Topic

#### Page

6.1	Flash Memory Organization	76
6.2	Flash Write	76
6.3	Flash Page Erase	78
6.4	Flash DMA Trigger	79
6.5	Flash Controller Registers	79



#### Flash Memory Organization

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#### 6.1 Flash Memory Organization

The flash memory is divided into 2048-byte or 1024-byte flash pages. A flash page is the smallest erasable unit in the memory, whereas a 32-bit word is the smallest writable unit that can be written to the flash.

When performing write operations, the flash memory is word-addressable using a 16-bit address written to the address registers FADDRH:FADDRL.

When performing page-erase operations, the flash memory page to be erased is addressed through the register bits FADDRH[7:1] (CC2530/CC2531/CC2540/CC2541) or FADDRH[6:0] (CC2533).

Note the difference in addressing the flash memory; when accessed by the CPU to read code or data, the flash memory is byte-addressable. When accessed by the flash controller, the flash memory is word-addressable, where a word consists of 32 bits.

The following sections describe the procedures for flash write and flash page-erase in detail.

### 6.2 Flash Write

The flash is programmed serially with a sequence of one or more 32-bit words (4 bytes), starting at the start address (set by FADDRH: FADDRL). In general, a page must be erased before writing can begin. The page-erase operation sets all bits in the page to 1. The chip-erase command (through the debug interface) erases all pages in the flash. This is the only way to set bits in the flash to 1. When writing a word to the flash, the 0-bits are programmed to 0 and the 1-bits are ignored (leaves the bit in the flash unchanged). Thus, bits are erased to 1 and can be written to 0. It is possible to write multiple times to a word. This is described in Section 6.2.2.

### 6.2.1 Flash-Write Procedure

The flash-write sequence algorithm is as follows:

- 1. Set FADDRH: FADDRL to the start address. (This is the 16 MSBs of the 18-bit byte address).
- 2. Set FCTL.WRITE to 1. This starts the write-sequence state machine.
- 3. Write four times to FWDATA within 20 µs (since the last time FCTL.FULL became 0, if not first iteration). LSB is written first. (FCTL.FULL goes high after the last byte.)
- 4. Wait until FCTL.FULL goes low. (The flash controller has started programming the 4 bytes written in step 3 and is ready to buffer the next 4 bytes).
- 5. Optional status check step:
  - If the 4 bytes were not written fast enough in step 3, the operation has timed out and FCTL.BUSY (and FCTL.WRITE) are 0 at this stage.
  - If the 4 bytes could not be written to the flash due to the page being locked, FCTL.BUSY (and FCTL.WRITE) are 0 and FCTL.ABORT is 1.
- 6. If this was the last 4 bytes then quit, otherwise go to step 3.

The write operation is performed using one of two methods:

- Using DMA transfer (preferred method)
- Using CPU, running code from SRAM

The CPU cannot access the flash, e.g., to read program code, while a flash-write operation is in progress. Therefore, the program code executing the flash write must be executed from RAM. See Section 2.2.1 for a description of how to run code from RAM.

When a flash-write operation is executed from RAM, the CPU continues to execute code from the next instruction after initiation of the flash-write operation (FCTL.WRITE = 1).

Power mode 1, 2, or 3 must not be entered while writing to the flash. Also, the system clock source (XOSC/RCOSC) must not be changed while writing. Note that setting CLKCONSTA.CLKSPD to a high value makes it impossible to meet the timing requirement of 20-µs write timing. With CLKCONSTA.CLKSPD = 111, the clock period is only 4 µs. It is therefore recommended to keep CLKCONSTA.CLKSPD at 000 or 001 while writing to the flash.

### 6.2.2 Writing Multiple Times to a Word

The following rules apply when writing multiple times to a 32-bit word between erase:

- Writing 0 to a bit within a 32-bit flash word, which has been set to 1 by the last erase operation, changes the state of the bit to 0, subject to the last bullet below.
- It is possible to write 0 to a bit within a 32-bit word repeatedly (subject to the last bullet below) once the bit has been written with 0. This does not change the state of the bit.
- Writing 1 to a bit does not change the state of the bit, subject to the last bullet below.
- The following limitations apply to writes subsequent to the last page erase:
  - A 0 must not be written more than two times to a single bit.
  - A 32-bit word shall not be written more than 8 times.
  - A page must not be written more than 1024 times.

The state of any bit of a 32-bit flash word is nondeterministic if these limitations are violated.

This makes it possible to write up to 4 new bits to a 32-bit word 8 times. One example write sequence to a word is shown in Table 6-1. Here  $b_n$  represents the 4 new bits written to the word for each update. This technique is useful to maximize the lifetime of the flash for data-logging applications.

Step	Value Written	FLASH Contents After Writing	Comment						
1	(page erase)	0xFFFFFFF	The erase sets all bits to 1.						
2	0xFFFFFFb ₀	0xFFFFFFb ₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
3	0xFFFFFb₁F	0xFFFFFb ₁ b ₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
4	0xFFFFFb ₂ FF	0xFFFFb₂b₁b₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
5	0xFFFFb ₃ FFF	0xFFFFb ₃ b ₂ b ₁ b ₀	Only the bits written 0 are set to 0, whereas all bit written 1 are ignored.						
6	0xFFFb ₄ FFFF	0xFFFb ₄ b ₃ b ₂ b ₁ b ₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
7	0xFFb₅FFFFF	0xFFb ₅ b ₄ b ₃ b ₂ b ₁ b ₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
8	0xFb ₆ FFFFFF	$0xFb_6b_5b_4b_3b_2b_1b_0$	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						
9	0xb ₇ FFFFFFF	0xb ₇ b ₆ b ₅ b ₄ b ₃ b ₂ b ₁ b ₀	Only the bits written 0 are set to 0, whereas all bits written 1 are ignored.						

### Table 6-1. Example Write Sequence

### 6.2.3 DMA Flash Write

When using DMA write operations, the data to be written into flash is stored in the XDATA memory space (RAM or registers). A DMA channel is configured to read the data to be written from the memory source address and write this data to the flash write-data register (FWDATA) fixed destination address, with the DMA trigger event FLASH (TRIG[4:0] = 1 0010 in DMA configuration) enabled. Thus, the flash controller triggers a DMA transfer when the flash write-data register, FWDATA, is ready to receive new data. The DMA channel should be configured to perform single-mode, byte-size transfers with the source address set to start-of-data block and destination address to fixed FWDATA (note that the block size, LEN in configuration data, must be divisible by 4; otherwise, the last word is not written to the flash). High priority should also be ensured for the DMA channel, so it is not interrupted in the write process. If interrupted for more than 20 µs, the write operation may time out, and the write bit, FCTL.WRITE, is set to 0.

When the DMA channel is armed, starting a flash write by setting FCTL.WRITE to 1 triggers the first DMA transfer (DMA and flash controller handle the reset of the transfer).

Figure 6-1 shows an example of how a DMA channel is configured and how a DMA transfer is initiated to write a block of data from a location in XDATA to flash memory.



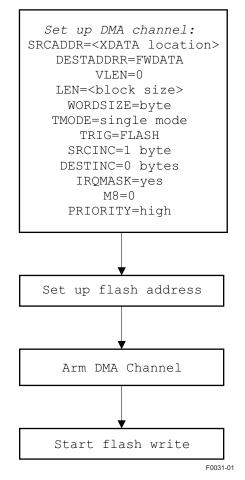


Figure 6-1. Flash Write Using DMA

### 6.2.4 CPU Flash Write

To write to the flash using the CPU, a program executing from SRAM must implement the steps outlined in the procedure described in Section 6.2.1. Disable interrupts to ensure the operation does not time out.

### 6.3 Flash Page Erase

The flash page-erase operation sets all bits in the page to 1.

A page erase is initiated by setting FCTL.ERASE to 1. The page addressed by FADDRH[7:1] (CC2530/CC2531/CC2540/CC2541) or FADDRH[6:0] (CC2533) is erased when a page erase is initiated simultaneously with a page write, i.e., FCTL.WRITE is set to 1, the page erase is performed before the page-write operation starts. The FCTL.BUSY bit can be polled to see when the page erase has completed.

Power mode 1, 2, or 3 must not be entered while erasing a page. Also, the system clock source (XOSC/RCOSC) must not be changed while erasing.

**NOTE:** If a flash page-erase operation is performed from within flash memory and the Watchdog Timer is enabled, a Watchdog Timer interval must be selected that is longer than 20 ms, the duration of the flash page-erase operation, so that the CPU can clear the Watchdog Timer.



## 6.3.1 Performing Flash Erase From Flash Memory

Note that while executing program code from within flash memory, when a flash erase or write operation is initiated the CPU stalls, and program execution resumes from the next instruction when the flash controller has completed the operation.

The following code example of how to erase one flash page in the CC2530 is given for use with the IAR compiler:

#include <ioCC2530.h>

```
unsigned char erase_page_num = 3; /* page number to erase, here: flash page #3 */
/* Erase one flash page */
EA = 0; /* disable interrupts */
while (FCTL & 0x80); /* poll FCTL.BUSY and wait until flash controller is ready */
FADDRH = erase_page_num << 1; /* select the flash page via FADDRH[7:1] bits */
FCTL |= 0x01; /* set FCTL.ERASE bit to start page erase */
while (FCTL & 0x80); /* optional: wait until flash write has completed (~20 ms) */
EA = 1; /* enable interrupts */</pre>
```

## 6.3.2 Different Flash Page Size on CC2533

The flash page size has been reduced from 2 KB (2048 bytes) on CC2530/CC2531/CC2540/CC2541 to 1 KB (1024 bytes) on CC2533. When performing page-erase operations on the flash memory, the page to be erased is addressed with the register bits FADDRH[6:0] on CC2533 as opposed to FADDRH[7:1] on CC2530/CC2531/CC2540. The page lock bits are still placed in the upper 16 bytes of the last accessible flash page.

### 6.4 Flash DMA Trigger

The flash DMA trigger is activated when flash data written to the FWDATA register has been written to the specified location in the flash memory, thus indicating that the flash controller is ready to accept new data to be written to FWDATA. Four trigger pulses are generated. In order to start the first transfer, one must set the FCTL.WRITE bit to 1. The DMA and the flash controller then handle all transfers automatically for the defined block of data (LEN in DMA configuration). It is further important that the DMA is armed prior to setting the FCTL.WRITE bit, that the trigger source is set to FLASH (TRIG[4:0] = 10010), and that the DMA has high priority so the transfer is not interrupted. If interrupted for more than 20  $\mu$ s, the write operation times out and FCTL.WRITE bit is cleared.

## 6.5 Flash Controller Registers

The flash controller registers are described in this section.



#### Flash Controller Registers

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Bit	Name	Reset	R/W	Description
7	BUSY	0	R	Indicates that write or erase is in operation. This flag is set when the WRITE or ERASE bit is set.
				0: No write or erase operation active
				1: Write or erase operation activated
6	FULL		R/H0	Write buffer-full status. This flag is set when 4 bytes have been written to FWDATA during flash write. The write buffer is then full and does not accept more data; i.e, writes to FWDATA are ignored when the FULL flag is set. The FULL flag is cleared when the write buffer again is ready to receive 4 more bytes. This flag is only needed when the CPU is used to write to the flash.
				0: Write buffer can accept more data.
				1: Write buffer full
5	ABORT	0	R/H0	Abort status. This bit is set when a write operation or page erase is aborted. An operation is aborted when the page accessed is locked. The abort bit is cleared when a write or page erase is started.
4	-	0	R	Reserved
3:2	CM[1:0]	01	R/W	Cache mode
				00: Cache disabled
				01: Cache enabled
				10: Cache enabled, prefetch mode
				11: Cache enabled, real-time mode
				Cache mode. Disabling the cache increases the power consumption and reduces performance. Prefetching, for most applications, improves performance by up to 33% at the expense of potentially increased power consumption. Real-time mode provides predictable flash-read access time; the execution time is equal to that in cache-disabled mode, but the power consumption is lower.
				Note: The value read always represents the current cache mode. Writing a new cache mode starts a cache mode-change request that may take several clock cycles to complete. Writing to this register is ignored if there is a current cache-change request in progress.
1	WRITE	0	R/W1/ H0	Write. Start writing word at location given by FADDRH: FADDRL. The WRITE bit stays at 1 until the write completes. The clearing of this bit indicates that the erase has completed, i.e., it has timed out or aborted.
				If ERASE is also set to 1, a page erase of the whole page addressed by FADDRH[7:1] is performed before the write. Setting WRITE to 1 when ERASE is 1 has no effect.
0	ERASE	0	R/W1/ H0	Page erase. Erase the page that is given by FADDRH[7:1] (CC2530/CC2531/CC2540/CC2541) or FADDRH[6:0] (CC2533). The ERASE bit stays at 1 until the erase completes. The clearing of this bit indicates that the erase has completed successfully or aborted. Setting ERASE to 1 when WRITE is 1 has no effect.

#### FWDATA (0x6273) – Flash Write Data

Bit	Name	Reset	R/W	Description
7:0	FWDATA[7:0]	0x00	R0/W	Flash write data. This register can only be written to when FCTL.WRITE is 1.

### FADDRH (0x6272) - Flash-Address High Byte

Bit	Name	Reset	R/W	Description
7:0	FADDRH[7:0]	0x00		Page address/high byte of flash word address Bits [7:1] (CC2530/CC2531/CC2540/CC2541) or bits [6:0] (CC2533) select which page to access.

#### FADDRL (0x6271) - Flash-Address Low Byte

Bit	Name	Reset	R/W	Description
7:0	FADDRL[7:0]	0x00	R/W	Low byte of flash word address



Page

There are 21 digital input/output pins that can be configured as general-purpose digital I/O or as peripheral I/O signals connected to the ADC, timers, or USART peripherals. The use of the I/O ports is fully configurable from user software through a set of configuration registers.

The I/O ports have the following key features:

- · 21 digital input/output pins
- General-purpose I/O or peripheral I/O
- Pullup or pulldown capability on inputs
- External interrupt capability

The external interrupt capability is available on all 21 I/O pins. Thus, external devices may generate interrupts if required. The external interrupt feature can also be used to wake the device up from sleep mode (power modes PM1/PM2/PM3).

#### Topic

#### 7.1 7.2 Low I/O Supply Voltage ......82 7.3 7.4 General-Purpose I/O Interrupts ..... 82 7.5 General-Purpose I/O DMA ......83 7.6 Peripheral I/O ......83 7.7 7.8 7.9 Radio Test Output Signals ......86 7.10 7.11

#### 7.1 Unused I/O Pins

Unused I/O pins should have a defined level and not be left floating. One way to do this is to leave the pin unconnected and configure the pin as a general-purpose I/O input with pullup resistor. This is also the state of all pins during and after reset (except P1.0 and P1.1, which do not have pullup/pulldown capability). Alternatively, the pin can be configured as a general-purpose I/O output. In either case, the pin should not be connected directly to VDD or GND, in order to avoid excessive power consumption.

### 7.2 Low I/O Supply Voltage

In applications where the digital I/O power supply voltage pins, DVDD1 and DVDD2, are below 2.6 V, the register bit PICTL.PADSC should be set to 1 in order to obtain the output dc characteristics specified in the *DC Characteristics* table in the device data sheet (Appendix C).

### 7.3 General-Purpose I/O

When used as general-purpose I/O, the pins are organized as three 8-bit ports, Port 0, Port 1, and Port 2; denoted P0, P1, and P2. P0 and P1 are complete 8-bit-wide ports, whereas P2 has only five usable bits. All ports are both bit- and byte-addressable through the SFR registers P0, P1, and P2. Each port pin can individually be set to operate as a general-purpose I/O or as a peripheral I/O.

The output drive strength is 4 mA on all outputs, except for the two high-drive outputs, P1.0 and P1.1, which each have 20-mA output drive strength.

The registers PxSEL, where x is the port number 0–2, are used to configure each pin in a port as either a general-purpose I/O pin or as a peripheral I/O signal. By default, after a reset, all digital input/output pins are configured as general-purpose input pins.

To change the direction of a port pin, the registers PxDIR are used to set each port pin to be either an input or an output. Thus, by setting the appropriate bit within PxDIR to 1, the corresponding pin becomes an output.

When reading the port registers P0, P1, and P2, the logic values on the input pins are returned regardless of the pin configuration. This does not apply during the execution of read-modify-write instructions. The read-modify-write instructions are: ANL, ORL, XRL, JBC, CPL, INC, DEC, DJNZ, MOV, CLR, and SETB. Operating on a port register, the following is true: When the destination is an individual bit in port register P0, P1, or P2, the value of the register, not the value on the pin, is read, modified, and written back to the port register.

When used as an input, the general-purpose I/O port pins can be configured to have a pullup, pulldown or three-state mode of operation. By default, after a reset, inputs are configured as inputs with pullup. To deselect the pullup or pulldown function on an input, the appropriate bit within the PxINP must be set to 1. The I/O port pins P1.0 and P1.1 do not have pullup/pulldown capability. Note that pins configured as peripheral I/O signals do not have pullup/pulldown capability, even if the peripheral function is an input.

In power modes PM1, PM2, and PM3, the I/O pins retain the I/O mode and output value (if applicable) that was set when PM1/PM2/PM3 was entered.

### 7.4 General-Purpose I/O Interrupts

General-purpose I/O pins configured as inputs can be used to generate interrupts. The interrupts can be configured to trigger on either a rising or falling edge of the external signal. Each of the P0, P1, and P2 ports has port interrupt-enable bits common for all bits within the port located in the IEN1-IEN2 registers as follows:

- IEN1.POIE: P0 interrupt enable
- IEN2.P1IE: P1 interrupt enable
- IEN2.P2IE: P2 interrupt enable

In addition to these common interrupt enables, the bits within each port have individual interrupt enables located in SFR registers POIEN, PIIEN, and P2IEN. Even I/O pins configured as peripheral I/O or general-purpose outputs have interrupts generated when enabled.

When an interrupt condition occurs on one of the I/O pins, the interrupt status flag in the corresponding P0–P2 interrupt flag register, P0IFG, P1IFG, or P2IFG, is set to 1. The interrupt status flag is set regardless of whether the pin has its interrupt enable set. When an interrupt is serviced, the interrupt status flag is cleared by writing a 0 to that flag. This flag must be cleared prior to clearing the CPU port interrupt flag (PxIF).

The SFR registers used for interrupts are described later in this section. The registers are summarized as follows:

- POIEN: P0 interrupt enables
- PIIEN: P1 interrupt enables
- P2IEN: P2 interrupt enables
- PICTL: P0, P1, and P2 edge configuration
- POIFG: P0 interrupt flags
- P1IFG: P1 interrupt flags
- P2IFG: P2 interrupt flags

### 7.5 General-Purpose I/O DMA

When used as general-purpose I/O pins, the P0 and P1 ports are each associated with one DMA trigger. These DMA triggers are IOC_0 for P0 and IOC_1 for P1, as shown in Table 8-1.

The IOC_0 trigger is activated when an interrupt occurs on the P0 pins. The IOC_1 trigger is activated when an interrupt occurs on the P1 pins.

#### 7.6 Peripheral I/O

This section describes how the digital I/O pins are configured as peripheral I/Os. For each peripheral unit that can interface with an external system through the digital input/output pins, a description of how peripheral I/Os are configured is given in the following subsections.

For USART and timer I/O, setting the appropriate PxSEL bits to 1 is required for the output signals on a digital I/O pin to be controlled by the peripheral. For peripheral inputs from digital I/O pins, this is optional. PxSEL = 1 overrides the pullup/pulldown settings of a pin, so to be able to control pullup/pulldown with the PxINP bits, the PxSEL bit should be set to 0 for that pin.

Note that peripheral units have two alternative locations for their I/O pins; see Table 7-1. Priority can be set between peripherals if conflicting settings regarding I/O mapping are present (using the P2SEL.PRIxP1 and P2DIR.PRIP0 bits). All combinations not causing conflicts can be used.

Note that a peripheral normally is present at the selected location even if it is not used, and another peripheral that is to use the pins must be given higher priority. The exception is the RTS and CTS pins of a USART in UART mode with flow control disabled and the SSN pin of a USART configured in SPI master mode.

Note also that peripheral units that have input pins receive an input from the pin regardless of the PxINP setting, and this may influence on the state of the peripheral unit. For instance, a UART should be flushed before use if there may have been activity on the RX pin prior to taking it in use as a UART pin.

Periphery/				P	0				P1									P2				
Function	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	4	3	2	1	0	
ADC	A7	A6	A5	A4	A3	A2	A1	A0													Т	
Operational amplifier						0	-	+														
Analog comparator			+	_																		
USART 0 SPI			С	SS	MO	MI																
Alt. 2											MO	MI	С	SS								
USART 0 UART			RT	СТ	ΤХ	RX																

Table 7-1. Peripheral I/O Pin Mapping

				Ius	10 1	••••	on p	iora		• •••	map	ping	(0)	0		,					
Periphery/				F	0							F	י1						P2		
Function	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	4	3	2	1	0
Alt. 2											ТΧ	RX	RT	СТ							
USART 1 SPI			MI	MO	С	SS															
Alt. 2									MI	MO	С	SS									
USART 1 UART			RX	тх	RT	СТ															
Alt. 2									RX	ΤХ	RT	СТ									
TIMER 1		4	3	2	1	0															
Alt. 2	3	4												0	1	2					
TIMER 3												1	0								
Alt. 2									1	0											
TIMER 4															1	0					
Alt. 2																		1			0
32-kHz XOSC																	Q1	Q2			
DEBUG																			DC	DD	
OBSSEL											5	4	3	2	1	0					

### Table 7-1. Peripheral I/O Pin Mapping (continued)

## 7.6.1 Timer 1

PERCFG.T1CFG selects whether to use alternative 1 or alternative 2 locations.

In Table 7-1, the Timer 1 signals are shown as the following:

- 0: Channel 0 capture/compare pin
- 1: Channel 1 capture/compare pin
- 2: Channel 2 capture/compare pin
- 3: Channel 3 capture/compare pin
- 4: Channel 4 capture/compare pin

P2DIR.PRIP0 selects the order of precedence when assigning several peripherals to Port 0. When set to 10, Timer 1 channels 0–1 have precedence, and when set to 11, Timer 1 channels 2–3 have precedence. To have all Timer 1 channels visible in the alternative 1 location, move both USART 0 and USART 1 to the alternative 2 location.

P2SEL.PRI1P1 and P2SEL.PRI0P1 select the order of precedence when assigning several peripherals to Port 1. The Timer 1 channels have precedence when the former is set low and the latter is set high.

## 7.6.2 Timer 3

PERCFG.T3CFG selects whether to use alternative 1 or alternative 2 locations.

In Table 7-1, the Timer 3 signals are shown as the following:

- 0: Channel 0 capture/compare pin
- 1: Channel 1 capture/compare pin

P2SEL.PRI2P1 and P2SEL.PRI3P1 select the order of precedence when assigning several peripherals to Port 1. The Timer 3 channels have precedence when both bits are set high. If P2SEL.PRI2P1 is set high and P2SEL.PRI3P1 is set low, the Timer 3 channels have precedence over USART 1, but USART 0 has precedence over the Timer 3 channels as well as over USART 1.

### 7.6.3 Timer 4

PERCFG. T4CFG selects whether to use alternative 1 or alternative 2 locations.

In Table 7-1, the Timer 4 signals are shown as the following:

• 0: Channel 0 capture/compare pin

• 1: Channel 1 capture/compare pin

P2SEL.PRI1P1 selects the order of precedence when assigning several peripherals to Port 1. The Timer 4 channels have precedence when the bit is set.

## 7.6.4 USART 0

The SFR register bit PERCFG.U0CFG selects whether to use alternative 1 or alternative 2 locations.

In Table 7-1, the USART 0 signals are shown as follows:

UART:

- RX: RXDATA
- TX: TXDATA
- RT: RTS
- CT: CTS

SPI:

- MI: MISO
- MO: MOSI
- C: SCK
- SS: SSN

P2DIR.PRIP0 selects the order of precedence when assigning several peripherals to Port 0. When set to 00, USART 0 has precedence. Note that if UART mode is selected and hardware flow control is disabled, USART 1 or Timer 1 has precedence to use ports P0.4 and P0.5.

P2SEL.PRI3P1 and P2SEL.PRI0P1 select the order of precedence when assigning several peripherals to Port 1. USART 0 has precedence when both are set to 0. Note that if UART mode is selected and hardware flow control is disabled, Timer 1 or Timer 3 has precedence to use ports P1.2 and P1.3.

### 7.6.5 USART 1

The SFR register bit PERCFG.U1CFG selects whether to use alternative 1 or alternative 2 locations.

In Table 7-1, the USART 1 signals are shown as follows:

UART:

- RX: RXDATA
- TX: TXDATA
- RT: RTS
- CT: CTS

SPI:

- MI: MISO
- MO: MOSI
- C: SCK
- SS: SSN

P2DIR.PRIP0 selects the order of precedence when assigning several peripherals to Port 0. When set to 01, USART 1 has precedence. Note that if UART mode is selected and hardware flow control is disabled, USART 0 or Timer 1 has precedence to use ports P0.2 and P0.3.

P2SEL.PRI3P1 and P2SEL.PRI2P1 select the order of precedence when assigning several peripherals to Port 1. USART 1 has precedence when the former is set to 1 and the latter is set to 0. Note that if UART mode is selected and hardware flow control is disabled, USART 0 or Timer 3 has precedence to use ports P1.4 and P1.5.

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Debug Interface

### 7.6.6 ADC

In Table 7-1, the ADC signals are shown as follows:

- A0: ADC input 0
- A1: ADC input 1
- A2: ADC input 2
- A3: ADC input 3
- A4: ADC input 4
- A5: ADC input 5
- A6: ADC input 6
- A7: ADC input 7
- T: ADC external trigger pin

When using the ADC, Port 0 pins must be configured as ADC inputs. Up to eight ADC inputs can be used. To configure a Port 0 pin to be used as an ADC input, the corresponding bit in the APCFG register must be set to 1. The default values in this register select the Port 0 pins as non-ADC input, i.e., digital input/outputs.

The settings in the APCFG register override the settings in POSEL.

The ADC can be configured to use the general-purpose I/O pin P2.0 as an external trigger to start conversions. P2.0 must be configured as a general-purpose I/O in input mode when being used for ADC external trigger.

## 7.6.7 Operational Amplifier and Analog Comparator

When using the operational amplifier and analog comparator, the corresponding Port 0 pins must be configured as ADC inputs (see Table 7-1). To configure a Port 0 pin to be used as an ADC input, the corresponding bit in the APCFG register must be set to 1. The default values in this register select the Port 0 pins as non-ADC input, i.e., digital input/outputs.

The settings in the APCFG register override the settings in POSEL.

## 7.7 Debug Interface

Ports P2.1 and P2.2 are used for debug data and clock signals, respectively. These are shown as DD (debug data) and DC (debug clock) in Table 7-1. When in debug mode, the debug interface controls the direction of these pins. Pullup/pulldown is disabled on these pins while in debug mode.

## 7.8 32-kHz XOSC Input

Ports P2.3 and P2.4 can be used to connect an external 32-kHz crystal. These port pins are used by the 32-kHz XOSC when CLKCONCMD.OSC32K is low, regardless of register settings. The port pins are set in analog mode when CLKCONCMD.OSC32K is low.

## 7.9 Radio Test Output Signals

By using the OBSSELx registers (OBSSEL0-OBSSEL5) the user can output different signals from the RF Core to GPIO pins. These signals can be useful for debugging of low-level protocols or control of external PA, LNA, or switches. The control registers OBSSEL0-OBSSEL5 can be used to override the standard GPIO behavior and output RF Core signals (rfc_obs_sig0, rfc_obs_sig1, and rfc_obs_sig2) on the pins P1[0:5]. For a list of available signals, see the respective RFC_OBS_CTRLx registers in Section 23.15.3 for CC253x or Section 24.1 for CC2540 or Chapter 25 for CC2541.

## 7.10 Power-Down Signal MUX (PMUX)

The PMUX register can be used to output the 32-kHz clock and/or the digital regulator status.

The selected 32-kHz clock source can be output on one of the P0 pins. The enable bit CKOEN enables the output on P0, and the pin of P0 is selected using the CKOPIN (see the PMUX register description for details). When CKOEN is set, all other configurations for the selected pin are overridden. The clock is output in all power modes; however, in PM3 the clock stops (see PM3 in Chapter 4).

Furthermore, the digital regulator status can be output on one of the P1 pins. When the DREGSTA bit is set, the status of the digital regulator is output. DREGSTAPIN selects the P1 pin (see the PMUX register description for details). When DREGSTA is set, all other configurations for the selected pin are overridden. The selected pin outputs 1 when the 1.8-V on-chip digital regulator is powered up (chip has regulated power). The selected pin outputs 0 when the 1.8-V on-chip digital regulator is powered down, i.e., in PM2 and PM3.

## 7.11 I/O Registers

The registers for the I/O ports are described in this section. The registers are:

- P0: Port 0
- P1: Port 1
- P2: Port 2
- PERCFG: Peripheral-control register
- APCFG: Analog peripheral I/O configuration
- POSEL: Port 0 function-select register
- PISEL: Port 1 function-select register
- P2SEL: Port 2 function-select register
- PODIR: Port 0 direction register
- PIDIR: Port 1 direction register
- P2DIR: Port 2 direction register
- POINP: Port 0 input-mode register
- Plinp: Port 1 input-mode register
- P2INP: Port 2 input-mode register
- POIFG: Port 0 interrupt-status flag register
- PIIFG: Port 1 interrupt-status flag register
- P2IFG: Port 2 interrupt-status flag register
- PICTL: Interrupt edge register
- POIEN: Port 0 interrupt-mask register
- PIIEN: Port 1 interrupt-mask register
- P2IEN: Port 2 interrupt-mask register
- PMUX: Power-down signal-mux register
- OBSSEL0: Observation output control register 0
- OBSSEL1: Observation output control register 1
- OBSSEL2: Observation output control register 2
- OBSSEL3: Observation output control register 3
- OBSSEL4: Observation output control register 4
- OBSSEL5: Observation output control register 5

#### P0 (0x80) - Port 0

Bit	Name	Reset	R/W	Description
7:0	P0[7:0]	0xFF	R/W	Port 0. General-purpose I/O port. Bit-addressable from SFR. This CPU-internal register is readable, but not writable, from XDATA (0x7080).



I/O Registers

P1 (0	P1 (0x90) – Port 1											
Bit	Name	Reset	R/W	Description								
7:0	P1[7:0]	0xFF	R/W	Port 1. General-purpose I/O port. Bit-addressable from SFR. This CPU-internal register is readable, but not writable, from XDATA (0x7090).								

#### P2 (0xA0) - Port 2

Bit	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved
4:0	P2[4:0]	1 1111	R/W	Port 2. General-purpose I/O port. Bit-addressable from SFR. This CPU-internal register is readable, but not writable, from XDATA (0x70A0).

### PERCFG (0xF1) – Peripheral Control

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	T1CFG	0	R/W	Timer 1 I/O location
				0: Alternative 1 location
				1: Alternative 2 location
5	T3CFG	0	R/W	Timer 3 I/O location
				0: Alternative 1 location
				1: Alternative 2 location
4	T4CFG	0	R/W	Timer 4 I/O location
				0: Alternative 1 location
				1: Alternative 2 location
3:2	-	00	R/W	Reserved
1	U1CFG	0	R/W	USART 1 I/O location
				0: Alternative 1 location
				1: Alternative 2 location
0	U0CFG	0	R/W	USART 0 I/O location
				0: Alternative 1 location
				1: Alternative 2 location

#### APCFG (0xF2) – Analog Peripheral I/O Configuration

Bit	Name	Reset	R/W	Description
7:0	APCFG[7:0]	0x00	R/W	Analog Perpheral I/O configuration . APCFG[7:0] select P0.7–P0.0 as analog I/O.
				0: Analog I/O disabled
				1: Analog I/O enabled

#### P0SEL (0xF3) – Port 0 Function Select

Bit	Name	Reset	R/W	Description
7:0	SELP0_[7:0]	0x00	R/W	P0.7 to P0.0 function select
				0: General-purpose I/O
				1: Peripheral function

#### P1SEL (0xF4) – Port 1-Function Select

Bit	Name	Reset	R/W	Description
7:0	SELP1_[7:0]	0x00	R/W	P1.7 to P1.0 function select
				0: General-purpose I/O
				1: Peripheral function



Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	PRI3P1 0 F		R/W	Port 1 peripheral priority control. This bit determines which module has priority in the case when modules are assigned to the same pins.
				0: USART 0 has priority.
				1: USART 1 has priority.
5	PRI2P1 0 R/W		R/W	Port 1 peripheral priority control. This bit determines the order of priority in the case when PERCFG assigns USART 1 and Timer 3 to the same pins.
				0: USART 1 has priority.
				1: Timer 3 has priority.
4	PRI1P1	0	R/W	Port 1 peripheral priority control. This bit determines the order of priority in the case when PERCFG assigns Timer 1 and Timer 4 to the same pins.
				0: Timer 1 has priority.
				1: Timer 4 has priority.
3	PRIOP1	0	R/W	Port 1 peripheral priority control. This bit determines the order of priority in the case when PERCFG assigns USART 0 and Timer 1 to the same pins.
				0: USART 0 has priority.
				1: Timer 1 has priority.
2	SELP2_4	0	R/W	P2.4 function select
				0: General-purpose I/O
				1: Peripheral function
1	SELP2_3	0	R/W	P2.3 function select
				0: General-purpose I/O
				1: Peripheral function
0	SELP2_0	0	R/W	P2.0 function select
				0: General-purpose I/O
				1: Peripheral function

#### P0DIR (0xFD) - Port 0 Direction

Bit	Name	Reset	R/W	Description
7:0	DIRP0_[7:0]	0x00	R/W	P0.7 to P0.0 I/O direction
				0: Input
				1: Output

### P1DIR (0xFE) – Port 1 Direction

Bit	Name	Reset	R/W	Description
7:0	DIRP1_[7:0]	0x00	R/W	P1.7 to P1.0 I/O direction
				0: Input
				1: Output



#### I/O Registers

Bit	Name	Reset	R/W	Description
7:6	PRIP0[1:0]	00	R/W	Port 0 peripheral priority control. These bits determine the order of priority in the case when PERCFG assigns several peripherals to the same pins.
				Detailed priority list:
				00:
				1st priority: USART 0
				2nd priority: USART 1
				3rd priority: Timer 1
				01:
				1st priority: USART 1
				2nd priority: USART 0
				3rd priority: Timer 1
				10:
				1st priority: Timer 1 channels 0-1
				2nd priority: USART 1
				3rd priority: USART 0
				4th priority: Timer 1 channels 2-3
				11:
				1st priority: Timer 1 channels 2-3
				2nd priority: USART 0
				3rd priority: USART 1
				4th priority: Timer 1 channels 0-1
5	_	0	R0	Reserved
4:0	DIRP2_[4:0]	0 0000	R/W	P2.4 to P2.0 I/O direction
				0: Input
				1: Output

#### P0INP (0x8F) - Port 0 Input Mode

Bit	Name	Reset	R/W	Description
7:0	MDP0_[7:0]	0x00	R/W	P0.7 to P0.0 I/O input mode
				0: Pullup/pulldown [see P2INP (0xF7) – Port 2 input mode]
				1: 3-state

#### P1INP (0xF6) - Port 1 Input Mode

Bit	Name	Reset	R/W	Description
7:2	MDP1_[7:2]	0000 00	R/W	P1.7 to P1.2 I/O input mode
				0: Pullup/pulldown [see P2INP (0xF7) - Port 2 input mode]
				1: 3-state
1:0	-	00	R0	Reserved



Bit	Name	Reset	R/W	Description
7	PDUP2	0	R/W	Port 2 pullup/pulldown select. Selects function for all Port 2 pins configured as pullup/pulldown inputs.
				0: Pullup
				1: Pulldown
6	PDUP1	0	R/W	Port 1 pullup/down select. Selects function for all Port 1 pins configured as pullup/pulldown inputs.
				0: Pullup
				1: Pulldown
5	PDUP0	0	R/W	Port 0 pullup/pulldown select. Selects function for all Port 0 pins configured as pullup/pulldown inputs.
				0: Pullup
				1: Pulldown
4:0	MDP2_[4:0]	0 0000	R/W	P2.4 to P2.0 I/O input mode
				0: Pullup/pulldown
				1: 3-state

#### P0IFG (0x89) - Port 0 Interrupt Status Flag

Bit	Name	Reset	R/W	Description
7:0	P0IF[7:0]	0x00	R/W0	Port 0, inputs 7 to 0 interrupt status flags. When an input port pin has an interrupt request pending, the corresponding flag bit is set.

#### P1IFG (0x8A) – Port 1 Interrupt Status Flag

Bit	Name	Reset	R/W	Description
7:0	P1IF[7:0]	0x00	R/W0	Port 1, inputs 7 to 0 interrupt status flags. When an input port pin has an interrupt request pending, the corresponding flag bit is set.

#### P2IFG (0x8B) – Port 2 Interrupt Status Flag

Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5	DPIF	0	R/W0	USB D+ interrupt-status flag. This flag is set when the D+ line has an interrupt request pending and is used to detect USB resume events in USB suspend state. This flag is not set when the USB controller is not suspended.
4:0	P2IF[4:0]	0 0000	R/W0	Port 2, inputs 4 to 0 interrupt status flags. When an input port pin has an interrupt request pending, the corresponding flag bit is set.



#### I/O Registers

Bit	Name	Reset	R/W	Description
7	PADSC	0	R/W	Drive strength control for I/O pins in output mode. Selects output drive strength enhancement to account for low I/O supply voltage on pin DVDD (this to ensure the same drive strength at lower voltages as at higher).
				0: Minimum drive strength enhancement. DVDD1/2 equal to or greater than 2.6 V
				1: Maximum drive strength enhancement. DVDD1/2 less than 2.6 V
6:4	-	000	R0	Reserved
3	P2ICON	0	R/W	Port 2, inputs 4 to 0 interrupt configuration. This bit selects the interrupt request condition for Port 2 inputs 4 to 0.
				0: Rising edge on input gives interrupt.
				1: Falling edge on input gives interrupt.
2	PIICONH	0	R/W	Port 1, inputs 7 to 4 interrupt configuration. This bit selects the interrupt request condition for the high nibble of Port 1 inputs.
				0: Rising edge on input gives interrupt.
				1: Falling edge on input gives interrupt
1	PIICONL	0	R/W	Port 1, inputs 3 to 0 interrupt configuration. This bit selects the interrupt request condition for the low nibble of Port 1 inputs.
				0: Rising edge on input gives interrupt.
				1: Falling edge on input gives interrupt.
0	POICON	0	R/W	Port 0, inputs 7 to 0 interrupt configuration. This bit selects the interrupt request condition for all Port 0 inputs.
				0: Rising edge on input gives interrupt.
	1			1: Falling edge on input gives interrupt.

#### P0IEN (0xAB) – Port 0 Interrupt Mask

Bit	Name	Reset	R/W	Description
7:0	P0_[7:0]IEN	0x00	R/W	Port P0.7 to P0.0 interrupt enable
				0: Interrupts are disabled.
				1: Interrupts are enabled.

#### P1IEN (0x8D) – Port 1 Interrupt Mask

Bit	Name	Reset	R/W	Description
7:0	P1_[7:0]IEN	0x00	R/W	Port P1.7 to P1.0 interrupt enable
				0: Interrupts are disabled.
				1: Interrupts are enabled.

#### P2IEN (0xAC) – Port 2 Interrupt Mask

Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5	DPIEN	0	R/W	USB D+ interrupt enable
				0: USB D+ interrupt disabled.
				1: USB D+ interrupt enabled.
4:0	P2_[4:0]IEN	0 0000	R/W	Port P2.4 to P2.0 interrupt enable
				0: Interrupts are disabled.
				1: Interrupts are enabled.



PMUX (0xAE) - Power-Down Signal Mux Bit Name Reset R/W Description 7 CKOEN 0 R/W Clock Out Enable. When this bit is set, the selected 32-kHz clock is output on one of the P0 pins. CKOPIN selects the pin to use. This overrides all other configuration for the selected pin. The clock is output in all power modes; however, in PM3 the clock stops (see PM3 in Chapter 4). CKOPIN[2:0] 000 R/W Clock Out Pin. Selects which P0 pin is to be used to output the selected 32-kHz clock. 6:4 DREGSTA Digital Regulator Status. When this bit is set, the status of the digital regulator is output on one 3 0 R/W of the P1 pins. DREGSTAPIN selects the pin. When DREGSTA is set, all other configurations for the selected pin are overridden. The selected pin outputs 1 when the 1.8-V on-chip digital regulator is powered up (chip has regulated power). The selected pin outputs 0 when the 1.8-V on-chip digital regulator is powered down. Digital Regulator Status Pin. Selects which P1 pin is to be used to output DREGSTA signal. DREGSTAPIN[2:0] 000 R/W 2:0

Note that registers OBSSEL0 through OBSSEL5 do not retain data in states PM2 and PM3.

#### OBSSEL0 (0x6243) – Observation Output Control Register 0

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling the observation output 0 on P1[0]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.0.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 0 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 111 1101 (125): rfc_obs_sig2 Others: Reserved

#### OBSSEL1 (0x6244) – Observation Output Control Register 1

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling observation output 1 on P1[1]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.1.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 1 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 111 1101 (125): rfc_obs_sig2 Others: Reserved

#### OBSSEL2 (0x6245) – Observation Output Control Register 2

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling observation output 2 on P1[2]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.2.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 2 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 111 1101 (125): rfc_obs_sig2 Others: Reserved

#### OBSSEL3 (0x6246) – Observation Output Control Register 3

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling observation output 3 on P1[3]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.3.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 3 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 111 1101 (125): rfc_obs_sig2 Others: Reserved



I/O Registers

### OBSSEL4 (0x6247) – Observation Output Control Register 4

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling observation output 4 on P1[4]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.4.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 4 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 11 11101 (125): rfc_obs_sig2 Others: Reserved

#### OBSSEL5 (0x6248) – Observation Output Control Register 5

Bit	Name	Reset	R/W	Description
7	EN	0	R/W	Bit controlling the observation output 5 on P1[5]. 0 – Observation output disabled 1 – Observation output enabled Note: If enabled, this overwrites the standard GPIO behavior of P1.5.
6:0	SEL[6:0]	000 0000	R/W	Select output signal on observation output 5 111 1011 (123): rfc_obs_sig0 111 1100 (124): rfc_obs_sig1 111 1101 (125): rfc_obs_sig2 Others: Reserved



The Direct Memory Access (DMA) Controller can be used to relieve the 8051 CPU core of handling data movement operations, thus achieving high overall performance with good power efficiency. The DMA controller can move data from a peripheral unit such as ADC or RF transceiver to memory with minimum CPU intervention.

The DMA controller coordinates all DMA transfers, ensuring that DMA requests are prioritized appropriately relative to each other and to CPU memory access. The DMA controller contains a number of programmable DMA channels for memory-memory data movement.

The DMA controller controls data transfers over the entire address range in XDATA memory space. Because most of the SFR registers are mapped into the DMA memory space, these flexible DMA channels can be used to unburden the CPU in innovative ways, e.g., to feed a USART with data from memory or periodically to transfer samples between ADC and memory, etc. Use of the DMA can also reduce system power consumption by keeping the CPU in a low-power mode without having to wake up to move data to or from a peripheral unit (see Section 4.1.1 for CPU low-power mode). Note that Section 2.2.3 describes the SFR registers that are not mapped into XDATA memory space.

The main features of the DMA controller are as follows:

- Five independent DMA channels
- Three configurable levels of DMA channel priority
- 32 configurable transfer trigger events
- · Independent control of source and destination address
- Single, block and repeated transfer modes
- Supports length field in transfer data, setting variable transfer length
- · Can operate in either word-size or byte-size mode

#### Topic

#### Page

8.1	DMA Operation	. 96
8.2	DMA Configuration Parameters	. 98
8.3	DMA Configuration Setup	100
8.4	Stopping DMA Transfers	101
8.5	DMA Interrupts	101
8.6	DMA Configuration Data Structure	101
8.7	DMA Memory Access	101
8.8	DMA Registers	104



#### 8.1 DMA Operation

There are five DMA channels available in the DMA controller, numbered channel 0 through channel 4. Each DMA channel can move data from one place within the DMA memory space to another, i.e., between XDATA locations.

In order to use a DMA channel, it must first be configured as described in Section 8.2 and Section 8.3. Figure 8-1 shows the DMA state diagram.

Once a DMA channel has been configured, it must be armed before any transfers are allowed to be initiated. A DMA channel is armed by setting the appropriate bit in the DMA channel-arm register DMAARM.

When a DMA channel is armed, a transfer begins when the configured DMA trigger event occurs. Note that the time to arm one channel (i.e., get configuration data) takes nine system clocks; thus, if the corresponding DMAARM bit is set and a trigger appears within the time it takes to configure the channel, the wanted trigger is lost. If two or more DMA channels are armed simultaneously, the time for all channels to be configured is longer (sequential read from memory). If all five are armed, it takes 45 system clocks, and channel 1 is ready first, then channel 2, and lastly channel 0 (all within the last eight system clocks). There are 32 possible DMA trigger events (see Table 8-1), e.g., UART transfer, timer overflow. The trigger event to be used by a DMA channel is set by the DMA channel configuration; thus, no knowledge of this is available until after the configuration has been read. The DMA trigger events are listed in Table 8-1.

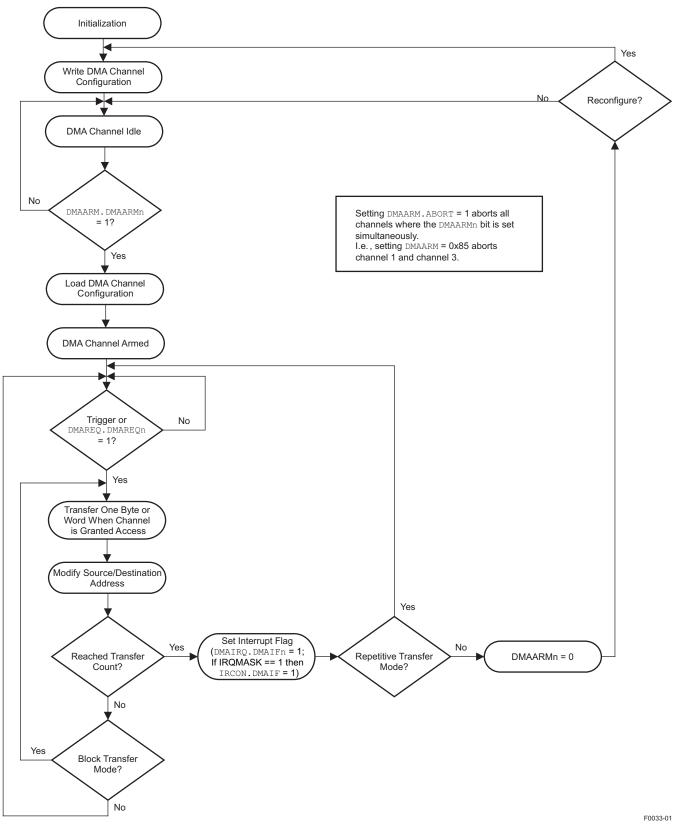
In addition to starting a DMA transfer through the DMA trigger events, the user software may force a DMA transfer to begin by setting the corresponding DMAREQ bit.

It should be noted that if the previously configured trigger source generates trigger events while DMA is being configured, these are counted as missed events, and as soon as the DMA channel is ready, the transfer is started. This occurs even though the new trigger source is not the same as the previous one. In some situations, this leads to errors in the transfer. In order to account for this, trigger source 0 should be the source between reconfigurations. This is achieved by setting up dummy source and destination addresses, using fixed length of one byte, block transfer, and trigger source 0. Enabling a software trigger (DMAREQ) clears missed-trigger counting, and no new triggers are generated while a new configuration is fetched from memory (unless software writes to DMAREQ for this channel).

A DMAREQ bit is cleared only when the corresponding DMA transfer occurs. The DMAREQ bit is not cleared when the channel is disarmed.



DMA Operation







#### 8.2 DMA Configuration Parameters

Setup and control of the DMA operation is performed by the user software. This section describes the parameters which must be configured before a DMA channel can be used. Section 8.3 describes how the parameters are set up in software and passed to the DMA controller.

The behavior of each of the five DMA channels is configured with the following parameters:

**Source address:** The first address from which the DMA channel should read data.

**Destination address:** The first address to which the DMA channel should write the data read from the source address. The user must ensure that the destination is writable.

**Transfer count:** The number of transfers to perform before rearming or disarming the DMA channel and alerting the CPU with an interrupt request. The length can be defined in the configuration or it can be defined as described next for the VLEN setting.

**VLEN setting:** The DMA channel is capable of variable-length transfers, using the first byte or word to set the transfer length. When doing this, various options are available regarding how to count the number of bytes to transfer.

**Priority:** The priority of the DMA transfers for the DMA channel with respect to the CPU and other DMA channels and access ports.

**Trigger event:** All DMA transfers are initiated by so-called DMA trigger events. This trigger either starts a DMA block transfer or a single DMA transfer. In addition to the configured trigger, a DMA channel can always be triggered by setting its designated DMAREQ.DMAREQx flag. The DMA trigger sources are described in Table 8-1.

**Source and destination increment:** The source and destination addresses can be controlled to increment or decrement or not change.

**Transfer mode:** The transfer mode determines whether the transfer should be a single transfer or a block transfer, or repeated versions of these.

**Byte or word transfers:** Determines whether each DMA transfer should be 8-bit (byte) or 16-bit (word).

**Interrupt mask:** An interrupt request is generated on completion of the DMA transfer. The interrupt mask bit controls whether the interrupt generation is enabled or disabled.

**M8:** Decide whether to use seven or eight bits per byte byte for transfer length. This is only applicable when doing byte transfers.

A detailed description of all configuration parameters is given in Section 8.2.1 through Section 8.2.11.

#### 8.2.1 Source Address

The address in XDATA memory where the DMA channel starts to read data. This can be any XDATA address – in RAM, in the mapped flash bank (see MEMCTR.XBANK), XREG, or XDATA addressed SFR.

#### 8.2.2 Destination Address

The first address to which the DMA channel should write the data read from the source address. The user must ensure that the destination is writable. This can be any XDATA address – in RAM, XREG, or XDATA addressed SFR.

### 8.2.3 Transfer Count

The number of bytes/words that must be transferred for the DMA transfer to be complete. When the transfer count is reached, the DMA controller rearms or disarms the DMA channel and alerts the CPU with an interrupt request. The transfer count can be defined in the configuration or it can be defined as variable-length, as described in Section 8.2.4.

### 8.2.4 VLEN Setting

The DMA channel is capable of using the first byte or word (for word, bits 12:0 are used) in source data as



the transfer length. This allows variable-length transfers. When using variable-length transfer, various options regarding how to count number of bytes to transfer is given. In any case, the transfer-count (LEN) setting is used as a maximum transfer count. If the transfer length specified by the first byte or word is greater than LEN, then LEN bytes/words are transferred. When using variable-length transfers, then LEN should be set to the largest allowed transfer length plus one.

Note that the M8 bit (Section 8.2.11) is only used when byte-size transfers are chosen.

Options which can be set with VLEN are the following:

- 1. Transfer number of bytes/words commanded by first byte/word + 1 (transfers the length byte/word, and then as many bytes/words as dictated by the length byte/word)
- 2. Transfer number of bytes/words commanded by first byte/word
- 3. Transfer number of bytes/words commanded by first byte/word + 2 (transfers the length byte/word, and then as many bytes/words as dictated by the length byte/word + 1)
- 4. Transfer number of bytes/words commanded by first byte/word + 3 (transfers the length byte/word, and then as many bytes/words as dictated by the length byte/word + 2)

Byte/Word n + 2 Byte/Word n + 1 Byte/Word n + 1 Byte/Word n Byte/Word n Byte/Word n Byte/Word n - 1 Byte/Word n - 1 Byte/Word n - 1 Byte/Word n - 1 Time Byte/Word 3 Byte/Word 3 Byte/Word 3 Byte/Word 3 Byte/Word 2 Byte/Word 2 Byte/Word 2 Byte/Word 2 Byte/Word 1 Byte/Word 1 Byte/Word 1 Byte/Word 1 LENGTH = n LENGTH = n LENGTH = n LENGTH = n **VLEN = 001 VLEN = 010 VLEN = 011 VLEN = 100** 

Figure 8-2 shows the VLEN options.

Figure 8-2. Variable Length (VLEN) Transfer Options

## 8.2.5 Trigger Event

Each DMA channel can be set up to sense on a single trigger. This field determines which trigger the DMA channel senses.

## 8.2.6 Source and Destination Increment

When the DMA channel is armed or rearmed, the source and destination addresses are transferred to internal address pointers. The possibilities for address increment are:

- Increment by zero. The address pointer remains fixed after each transfer.
- Increment by one. The address pointer increments one count after each transfer.
- Increment by two. The address pointer increments two counts after each transfer.
- Decrement by one. The address pointer decrements one count after each transfer.

where a count equals 1 byte in byte mode and 2 bytes in word mode.

## 8.2.7 DMA Transfer Mode

The transfer mode determines how the DMA channel behaves when it starts transferring data. There are four transfer modes described as follows:

M0103-02



**Single:** On a trigger, a single DMA transfer occurs, and the DMA channel awaits the next trigger. After the number of transfers specified by the transfer count is completed, the CPU is notified, and the DMA channel is disarmed.

**Block:** On a trigger, the number of DMA transfers specified by the transfer count is performed as quickly as possible, after which the CPU is notified and the DMA channel is disarmed.

**Repeated single:** On a trigger, a single DMA transfer occurs, and the DMA channel awaits the next trigger. After the number of transfers specified by the transfer count is completed, the CPU is notified, and the DMA channel is rearmed.

**Repeated block:** On a trigger, the number of DMA transfers specified by the transfer count is performed as quickly as possible, after which the CPU is notified and the DMA channel is rearmed.

#### 8.2.8 DMA Priority

A DMA priority is configurable for each DMA channel. The DMA priority is used to determine the winner in the case of multiple simultaneous internal memory requests, and whether the DMA memory access should have priority or not over a simultaneous CPU memory access. In case of an internal tie, a round-robin scheme is used to ensure access for all. There are three levels of DMA priority:

High: Highest internal priority. DMA access always prevails over CPU access.

**Normal:** Second-highest internal priority. DMA access prevails over the CPU on at least every second try.

Low: Lowest internal priority. DMA access always defers to a CPU access.

#### 8.2.9 Byte or Word Transfers

Determines whether 8-bit (byte) or 16-bit (word) transfers are done.

#### 8.2.10 Interrupt Mask

On completing a DMA transfer, the channel can generate an interrupt to the processor. This bit masks the interrupt.

#### 8.2.11 Mode 8 Setting

This field determines whether to use 7 or 8 bits per byte for transfer length. Only applicable when doing byte transfers.

#### 8.3 DMA Configuration Setup

The DMA channel parameters such as address mode, transfer mode, and priority, described in the previous section, must be configured before a DMA channel can be armed and activated. The parameters are not configured directly through SFR registers, but instead they are written in a special DMA configuration data structure in memory. Each DMA channel in use requires its own DMA configuration-data structure. The DMA configuration data structure consists of eight bytes and is described in Section 8.6. A DMA configuration-data structure may reside at any location decided on by the user software, and the address location is passed to the DMA controller through a set of SFRs, DMAxCFGH: DMAxCFGL. Once a channel has been armed, the DMA controller reads the configuration data structure for that channel, given by the address in DMAxCFGH: DMAxCFGL.

It is important to note that the method for specifying the start address for the DMA configuration data structure differs between DMA channel 0 and DMA channels 1–4 as follows:

DMA0CFGH: DMA0CFGL gives the start address for the DMA channel 0 configuration data structure.

DMA1CFGH: DMA1CFGL gives the start address for the DMA channel 1 configuration data structure, followed by the channel 2–4 configuration data structures.

Thus, the DMA controller expects the DMA configuration data structures for DMA channels 1–4 to lie in a contiguous area in memory starting at the address held in DMA1CFGH:DMA1CFGL and consisting of 32 bytes.



#### 8.4 Stopping DMA Transfers

Ongoing DMA transfers or armed DMA channels are aborted using the DMAARM register to disarm the DMA channel.

One or more DMA channels are aborted by writing a 1 to the DMAARM.ABORT register bit, and at the same time selecting which DMA channels to abort by setting the corresponding DMAARM.DMAARMx bits to 1. When setting DMAARM.ABORT to 1, the DMAARM.DMAARMx bits for nonaborted channels must be written as 0.

No DMA interrupt is generated when aborting an ongoing DMA transfer (disarming a DMA channel).

### 8.5 DMA Interrupts

Each DMA channel can be configured to generate an interrupt to the CPU on completing a DMA transfer. This is accomplished with the IRQMASK bit in the channel configuration. The corresponding interrupt flag in the DMAIRQ SFR register is set when the interrupt is generated.

Regardless of the IRQMASK bit in the channel configuration, the corresponding interrupt flag in the DMAIRQ register is set on DMA channel completion. Thus, software should always check (and clear) this register when rearming a channel with a changed IRQMASK setting. Failure to do so could generate an interrupt based on the stored interrupt flag.

If a DMA transfer is aborted prior to its completion, the corresponding bit in the DMAIRQ register is not set, and an interrupt is not generated.

### 8.6 DMA Configuration Data Structure

For each DMA channel, the DMA configuration data structure consists of eight bytes. The configuration data structure is described in Table 8-2.

### 8.7 DMA Memory Access

The DMA data transfer is affected by endian convention. Note that the DMA descriptors follow big-endian convention while the other registers follow little-endian convention. This must be accounted for in compilers.

DMA Trigger		E-metional Unit	Description			
Number	Name	Functional Unit	Description			
0	NONE	DMA	No trigger, setting the DMAREQ.DMAREQx bit starts transfer.			
1	PREV	DMA	DMA channel is triggered by completion of previous channel.			
2	T1_CH0	Timer 1	Timer 1, compare, channel 0			
3	T1_CH1	Timer 1	Timer 1, compare, channel 1			
4	T1_CH2	Timer 1	Timer 1, compare, channel 2			
5	T2_EVENT1	Timer 2	Timer 2, event pulse 1			
6	T2_EVENT2	Timer 2	Timer 2, event pulse 2			
7	T3_CH0	Timer 3	Timer 3, compare, channel 0			
8	T3_CH1	Timer 3	Timer 3, compare, channel 1			
9	T4_CH0	Timer 4	Timer 4, compare, channel 0			
10	T4_CH1	Timer 4	Timer 4, compare, channel 1			
11	ST	Sleep Timer (not in CC2540/41)	Sleep Timer compare			
	RADIO1	Radio (CC2541)	Radio DMA trigger 1 (see Section 25.3.2)			
12	IOC_0	I/O controller	Port 0 I/O pin input transition ⁽¹⁾			
13	IOC_1	I/O controller	Port 1 I/O pin input transition ⁽¹⁾			

 Table 8-1. DMA Trigger Sources

⁽¹⁾ Using this trigger source must be aligned with port interrupt-enable bits. Note that all interrupt-enabled port pins generate a trigger.

DMA	Trigger	Functional Unit	Description
Number	Name	- Functional Unit	Description
14	URX0	USART 0	USART 0 RX complete
15	UTX0	USART 0	USART 0 TX complete
16	URX1	USART 1	USART 1 RX complete
17	UTX1	USART 1	USART 1 TX complete
18	FLASH	Flash controller	Flash data write complete
19	RADIO	Radio (not in CC2540)	CC253x: RF packet byte received (see Section 23.3) CC2541: Radio DMA trigger 0 (see Section 25.3.2)
20	ADC_CHALL	ADC	ADC end of a conversion in a sequence, sample ready
21	ADC_CH11	ADC	ADC end of conversion channel 0 in sequence, sample ready
22	ADC_CH21	ADC	ADC end of conversion channel 1 in sequence, sample ready
23	ADC_CH32	ADC	ADC end of conversion channel 2 in sequence, sample ready
24	ADC_CH42	ADC	ADC end of conversion channel 3 in sequence, sample ready
25	ADC_CH53	ADC	ADC end of conversion channel 4 in sequence, sample ready
26	ADC_CH63	ADC	ADC end of conversion channel 5 in sequence, sample ready
27	ADC_CH74	ADC	ADC end of conversion channel 6 in sequence, sample ready
28	ADC_CH84	ADC	ADC end of conversion channel 7 in sequence, sample ready
29	ENC_DW	AES	AES encryption processor requests download of input data
30	ENC_UP	AES	AES encryption processor requests upload of output data
31	DBG_BW	Debug interface	Debug interface burst write

 Table 8-1. DMA Trigger Sources (continued)

## Table 8-2. DMA Configuration Data Structure

Byte Offset	Bit	Name	Description
0	7:0	SRCADDR[15:8]	DMA channel source address, high
1	7:0	SRCADDR[7:0]	DMA channel source address, low
2	7:0	DESTADDR[15:8]	DMA channel destination address, high. Note that flash memory is not directly writable.
3	7:0	DESTADDR[7:0]	DMA channel destination address, low. Note that flash memory is not directly writable.
4	7:5	VLEN[2:0]	Variable-length transfer mode. In word mode, bits 12:0 of the first word are considered as the transfer length.
			000: Use LEN for transfer count
			001: Transfer the number of bytes/words specified by the first byte/word + 1 (up to a maximum specified by LEN). Thus, the transfer count excludes the length byte/word.
			010: Transfer the number of bytes/words specified by the first byte/word (up to a maximum specified by LEN). Thus, the transfer count includes the length byte/word.
			011: Transfer the number of bytes/words specified by the first byte/word + 2 (up to a maximum specified by LEN).
			100: Transfer the number of bytes/words specified by the first byte/word + 3 (up to a maximum specified by LEN).
			101: Reserved
			110: Reserved
			111: Alternative for using LEN as the transfer count
4	4:0	LEN[12:8]	The DMA channel transfer count
			Used as the maximum allowable length when VLEN differs from 000 and 111. The DMA channel counts in words when in WORDSIZE mode, and in bytes otherwise.
5	7:0	LEN[7:0]	The DMA channel transfer count
			Used as the maximum allowable length when VLEN differs from 000 and 111. The DMA channel counts in words when in WORDSIZE mode, and in bytes otherwise.
6	7	WORDSIZE	Selects whether each DMA transfer is 8-bit (0) or 16-bit (1).



Byte Offset	Bit	Name	Description
6	6:5	TMODE[1:0]	The DMA channel transfer mode
			00: Single
			01: Block
			10: Repeated single
			11: Repeated block
6	4:0	TRIG[4:0]	Selects one of the triggers shown in Table 8-1
7	7:6	<pre>SRCINC[1:0]</pre>	Source address increment mode (after each transfer):
			00: 0 bytes/words
			01: 1 byte/word
			10: 2 bytes/word
			11: -1 byte/word
7	5:4	DESTINC[1:0]	Destination address increment mode (after each transfer):
			00: 0 bytes/words
			01: 1 byte/word
			10: 2 bytes/words
			11: –1 byte/word
7	3	IRQMASK	Interrupt mask for this channel.
			0: Disable interrupt generation
			1: Enable interrupt generation on DMA channel done
7	2	м8	Mode of 8th bit for VLEN transfer length; only applicable when WORDSIZE = 0 and VLEN differs from 000 and 111.
			0: Use all 8 bits for transfer count
			1: Use 7 LSB for transfer count
7	1:0	PRIORITY[1:0]	The DMA channel priority:
			00: Low, CPU has priority.
			01: Assured, DMA at least every second try
			10: High, DMA has priority
			11: Reserved

Table 8-2. DMA Configuration Data Structure (continued)

### 8.8 DMA Registers

This section describes the SFR registers associated with the DMA controller.

Bit	Name	Reset	R/W	Description
7	ABORT	0	R0/W	DMA abort. This bit is used to stop ongoing DMA transfers. Writing a 1 to this bit aborts all channels which are selected by setting the corresponding DMAARM bit to 1.
				0: Normal operation
				1: Abort all selected channels
6:5	-	00	R/W	Reserved
4	DMAARM4	0	R/W1	DMA arm channel 4
				This bit must be set in order for any DMA transfers to occur on the channel. For nonrepetitive transfer modes, the bit is automatically cleared on completion.
3	DMAARM3	0	R/W1	DMA arm channel 3
				This bit must be set in order for any DMA transfers to occur on the channel. For nonrepetitive transfer modes, the bit is automatically cleared on completion.
2	DMAARM2	0	R/W1	DMA arm channel 2
				This bit must be set in order for any DMA transfers to occur on the channel. For nonrepetitive transfer modes, the bit is automatically cleared on completion.
1	DMAARM1	0	R/W1	DMA arm channel 1
				This bit must be set in order for any DMA transfers to occur on the channel. For nonrepetitive transfer modes, the bit is automatically cleared on completion.
0	DMAARM0	0	R/W1	DMA arm channel 0
				This bit must be set in order for any DMA transfers to occur on the channel. For nonrepetitive transfer modes, the bit is automatically cleared on completion.

#### DMAARM (0xD6) – DMA Channel Arm

#### DMAREQ (0xD7) – DMA Channel Start Request and Status

Bit	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved
4	DMAREQ4	0	R/W1 H0	DMA transfer request, channel 4
				When set to 1, activate the DMA channel (has the same effect as a single trigger event). This bit is cleared when DMA transfer is started.
3	DMAREQ3	0	R/W1 H0	DMA transfer request, channel 3
				When set to 1, activate the DMA channel (has the same effect as a single trigger event). This bit is cleared when DMA transfer is started.
2	DMAREQ2	0	R/W1 H0	DMA transfer request, channel 2
				When set to 1, activate the DMA channel (has the same effect as a single trigger event). This bit is cleared when DMA transfer is started.
1	DMAREQ1	0	R/W1 H0	DMA transfer request, channel 1
				When set to 1, activate the DMA channel (has the same effect as a single trigger event). This bit is cleared when DMA transfer is started.
0	DMAREQ0	0	R/W1 H0	DMA transfer request, channel 0
				When set to 1, activate the DMA channel (has the same effect as a single trigger event). This bit is cleared when DMA transfer is started.

#### DMA0CFGH (0xD5) – DMA Channel-0 Configuration Address High Byte

Bit	Name	Reset	R/W	Description
7:0	DMA0CFG[15:8]	0x00	R/W	The DMA channel-0 configuration address, high-order

#### DMA0CFGL (0xD4) – DMA Channel-0 Configuration Address Low Byte

Bit	Name	Reset	R/W	Description
7:0	DMA0CFG[7:0]	0x00	R/W	The DMA channel 0 configuration address, low-order



DMA	DMA1CFGH (0xD3) – DMA Channel 1–4 Configuration Address High Byte					
Bit	Name	Reset	R/W	Description		
7:0	DMA1CFG[15:8]	0x00	R/W	The DMA channel 1–4 configuration address, high-order		

#### DMA1CFGL (0xD2) – DMA Channel 1–4 Configuration Address Low Byte

Bit	Name	Reset	R/W	Description
7:0	DMA1CFG[7:0]	0x00	R/W	The DMA channel 1–4 configuration address, low-order

#### DMAIRQ (0xD1) – DMA Interrupt Flag

Bit	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved
4	DMAIF4	0	R/W0	DMA channel-4 interrupt flag
				0: DMA channel transfer not complete
				1: DMA channel transfer complete/interrupt pending
3	DMAIF3	0	R/W0	DMA channel-3 interrupt flag
				0: DMA channel transfer not complete
				1: DMA channel transfer complete/interrupt pending
2	DMAIF2	0	R/W0	DMA channel-2 interrupt flag
				0: DMA channel transfer not complete
				1: DMA channel transfer complete/interrupt pending
1	DMAIF1	0	R/W0	DMA channel-1 interrupt flag
				0: DMA channel transfer not complete
				1: DMA channel transfer complete/interrupt pending
0	DMAIF0	0	R/W0	DMA channel-0 interrupt flag
				0: DMA channel transfer not complete
				1: DMA channel transfer complete/interrupt pending



# Timer 1 (16-Bit Timer)

Timer 1 is an independent 16-bit timer which supports typical timer/counter functions such as input capture, output compare, and PWM functions. The timer has five independent capture/compare channels. The timer uses one I/O pin per channel. The timer is used for a wide range of control and measurement applications, and the availability of up/down count mode with five channels allows, for example, implementation of motor-control applications.

The features of Timer 1 are as follows:

- · Five capture/compare channels
- Rising, falling, or any-edge input capture
- Set, clear or toggle output compare
- Free-running, modulo, or up/down counter operation
- Clock prescaler for divide by 1, 8, 32, or 128
- · Interrupt request generated on each capture/compare and terminal count
- DMA trigger function

#### Topic

#### Page

9.1	16-Bit Counter	108
9.2	Timer 1 Operation	108
9.3	Free-Running Mode	108
9.4	Modulo Mode	108
9.5	Up/Down Mode	109
9.6	Channel-Mode Control	
9.7	Input Capture Mode	109
9.8	Output Compare Mode	
9.9	IR Signal Generation and Learning	115
9.10	Timer 1 Interrupts	117
9.11	Timer 1 DMA Triggers	117
9.12	Timer 1 Registers	118
9.13	Accessing Timer 1 Registers as Array	123



#### 9.1 16-Bit Counter

The timer consists of a 16-bit counter that increments or decrements at each active clock edge. The period of the active clock edges is defined by the register bits, CLKCONCMD.TICKSPD, which set the global division of the system clock, giving a variable clock-tick frequency from 0.25 MHz to 32 MHz (given the use of the 32-MHz XOSC as clock source). This frequency is further divided in Timer 1 by the prescaler value set by TICTL.DIV. This prescaler value can be 1, 8, 32, or 128. Thus, the lowest clock frequency used by Timer 1 is 1953.125 Hz and the highest is 32 MHz when the 32 MHz XOSC is used as system clock source. When the 16-MHz RCOSC is used as system clock source, then the highest clock frequency used by Timer 1 is 16 MHz.

The counter operates as a free-running counter, a modulo counter, or an up/down counter for use in center-aligned PWM.

It is possible to read the 16-bit counter value through the two 8-bit SFRs, T1CNTH and T1CNTL, containing the high-order byte and low-order byte, respectively. When T1CNTL is read, the high-order byte of the counter at that instant is buffered in T1CNTH so that the high-order byte can be read from T1CNTH. Thus, T1CNTL must always be read first, before reading T1CNTH.

All write accesses to the T1CNTL register reset the 16-bit counter.

The counter produces an interrupt request when the terminal count value (overflow) is reached. It is possible to start and halt the counter with T1CTL control register settings. The counter is started when a value other than 00 is written to T1CTL.MODE. If 00 is written to T1CTL.MODE, the counter halts at its present value.

#### 9.2 Timer 1 Operation

In general, control register T1CTL is used to control the timer operation. The status register T1STAT holds the interrupt flags. The various modes of operation are described as follows.

#### 9.3 Free-Running Mode

In the free-running mode of operation, the counter starts from 0x0000 and increments at each active clock edge. When the counter reaches 0xFFFF (overflow), the counter is loaded with 0x0000 and continues incrementing its value as shown in Figure 9-1. When the terminal count value 0xFFFF is reached, the interrupt flag T1STAT.OVFIF is set. An interrupt request is generated if enabled, see Section 9.10 for details. The free-running mode can be used to generate independent time intervals and output signal frequencies.

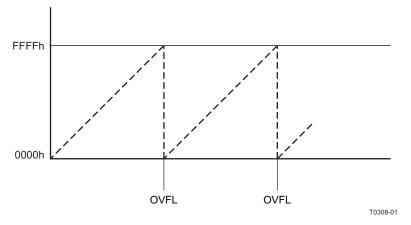


Figure 9-1. Free-Running Mode

#### 9.4 Modulo Mode

When the timer operates in modulo mode, the 16-bit counter starts at 0x0000 and increments at each active clock edge. After the counter has reached the period value T1CC0, held in registers T1CC0H:T1CC0L, the counter is reset to 0x0000 and continues to increment. If the timer is started with a



value above T1CC0, the interrupt flag T1STAT.OVFIF is set when the terminal count value (0xFFF) is reached, after which the counter wraps to 0x0000. An interrupt request is generated if enabled, see Section 9.10 for details. If a periodic interrupt is wanted at the period value, this can be obtained by enabling an output compare interrupt on channel 0, as explained in Section 9.8. The modulo mode can be used for applications where a period other then 0xFFFF is required. The counter operation is shown in Figure 9-2.

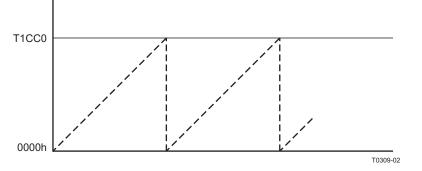


Figure 9-2. Modulo Mode

# 9.5 Up/Down Mode

In the up/down timer mode, the counter repeatedly starts from 0x0000 and counts up until the value held in T1CC0H:T1CC0L is reached, and then the counter counts down until 0x0000 is reached, as shown in Figure 9-3. This timer mode is used when symmetrical output pulses are required with a period other than 0xFFFF, and therefore allows implementation of center-aligned PWM output applications. The interrupt flag T1STAT.OVFIF is set when the counter value reaches 0x0000 in the up/down mode. An interrupt request is generated if enabled, see Section 9.10 for details.

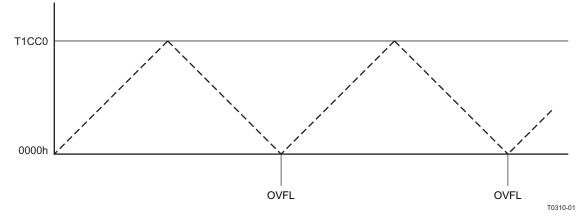


Figure 9-3. Up/Down Mode

# 9.6 Channel-Mode Control

The channel mode is set for each channel with its control and status register, T1CCTLn. The settings include input capture and output compare modes.

# 9.7 Input Capture Mode

When a channel is configured as an input capture channel, the I/O pin associated with that channel is configured as an input. After the timer has been started, a rising edge, falling edge, or any edge on the input pin triggers a capture of the 16-bit counter contents into the associated capture register. Thus, the timer is able to capture the time when an external event takes place.

**NOTE:** Before an I/O pin can be used by the timer, the required I/O pin must be configured as a Timer 1 peripheral pin.

The channel input pin is synchronized to the internal system clock. Thus, pulses on the input pin must have a minimum duration greater than the system clock period.

The content of the 16-bit capture register is read out from registers T1CCnH:T1CCnL.

When the capture takes place, the interrupt flag for the channel, T1STAT.CHnIF (n is the channel number), is set. An interrupt request is generated if enabled, see Section 9.10 for details.

# 9.8 Output Compare Mode

In output compare mode, the I/O pin associated with a channel is set as an output. After the timer has been started, the contents of the counter are compared with the contents of the channel compare register T1CCnH:T1CCnL. If the compare register equals the counter contents, the output pin is set, reset, or toggled, according to the compare output mode setting of T1CCTLn.CMP. Note that all edges on output pins are glitch-free when operating in a given output compare mode. Writing to the compare register T1CCnL is buffered, so that a value written to T1CCnL does not take effect until the corresponding high-order register, T1CCnH, is written. Writing to compare register T1CCnH:T1CCnL does not take effect on the output compare value until the counter value is 0x00.

Note that channel 0 has fewer output compare modes because T1CC0H:T1CC0L has a special function in modes 6 and 7, meaning these modes would not be useful for channel 0.

When a compare occurs, the interrupt flag for the channel, T1STAT.CHnIF (n is the channel number), is set. An interrupt request is generated if enabled, see Section 9.10 for details.

Examples of output compare modes in various timer modes are given in the following figures.

**Edge-aligned:** PWM output signals can be generated using the timer modulo mode and channels 1 and 2 in output compare mode 6 or 7 (defined by the T1CCTLn.CMP bits, where n is 1 or 2) as shown in Figure 9-4. The period of the PWM signal is determined by the setting in T1CC0, and the duty cycle is determined by T1CCn, where n is the PWM channel, 1 or 2.

The timer free-running mode may also be used. In this case, CLKCONCMD.TICKSPD and the prescaler divider value in the T1CTL.DIV bits set the period of the PWM signal. The polarity of the PWM signal is determined by whether output compare mode 6 or 7 is used.

PWM output signals can also be generated using output compare modes 4 and 5 as shown in Figure 9-4, or by using modulo mode as shown in Figure 9-5. Using output compare mode 4 or 5 is preferred for simple PWM.

**Center-aligned:** PWM outputs can be generated when the timer up/down mode is selected. The channel output compare mode 4 or 5 (defined by T1CCTLn.CMP bits, where n is 1 or 2) is selected, depending on the required polarity of the PWM signal. The period of the PWM signal is determined by T1CC0, and the duty cycle for the channel output is determined by T1CCn, where n is the PWM channel, 1 or 2.

The center-aligned PWM mode is required by certain types of motor-drive applications, and typically less noise is produced than in the edge-aligned PWM mode, because the I/O pin transitions are not lined up on the same clock edge.

In some types of applications, a defined delay or dead time is required between outputs. Typically, this is required for outputs driving an H-bridge configuration to avoid uncontrolled cross-conduction in one side of the H-bridge. The delay or dead-time can be obtained in the PWM outputs by using T1CCn as shown in the following:

Assuming that channel 1 and channel 2 are used to drive the outputs using timer up/down mode and the channels use output compare modes 4 and 5, respectively, then the timer period (in Timer 1 clock periods) is:

 $t_{P} = T1CC0 \times 2$ 

and the dead time, i.e., the time when both outputs are low, (in Timer 1 clock periods) is given by:  $t_p = T1CC1 - T1CC2$ 



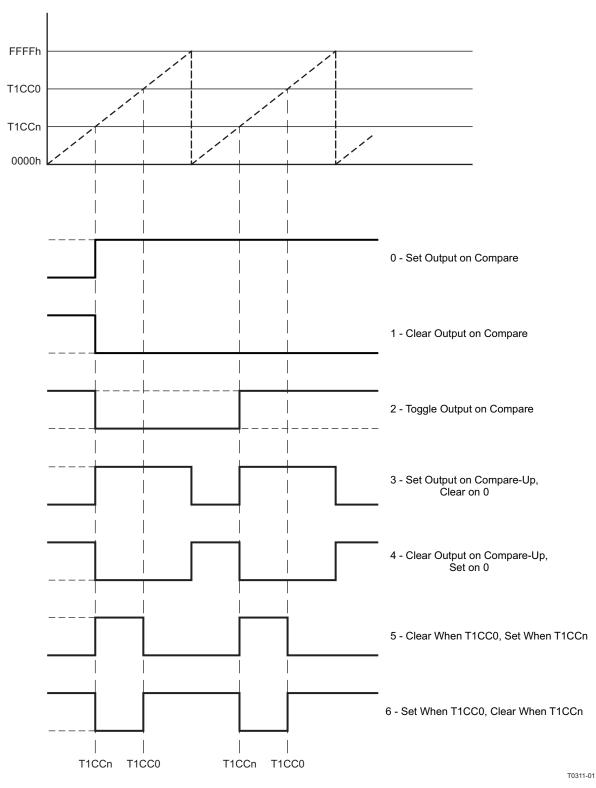
A compare output pin is initialized to the value listed in Table 9-1 when:

- a value is written to T1CNTL (all Timer 1 channels)
- 0x7 is written to T1CCTLn.CMP (channel n)

# Table 9-1. Initial Compare Output Values (Compare Mode)

Compare Mode (T1CCTLn.CMP)	Initial Compare Output
Set output on compare (000)	0
Clear output on compare (001)	1
Toggle output on compare (010)	0
Set output on compare-up, clear on compare down in up-down mode (011)	0
In other modes than up-down mode, set output on compare, clear on 0 (011)	0
Clear output on compare-up, set on compare down in up-down mode (100)	1
In other modes than up-down mode, clear output on compare, set on 0 (100)	1
Clear when equal T1CC0, set when equal T1CCn (101)	0
Set when equal T1CC0, clear when equal T1CCn (110)	1











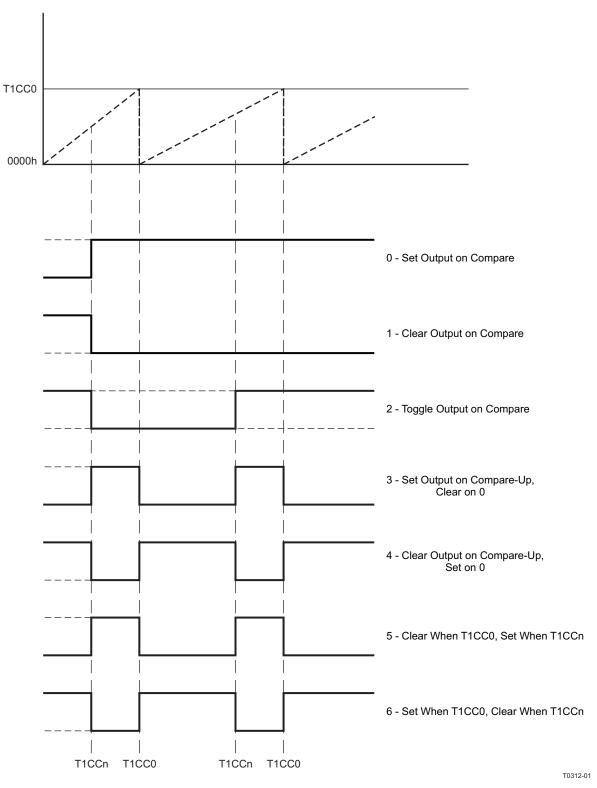
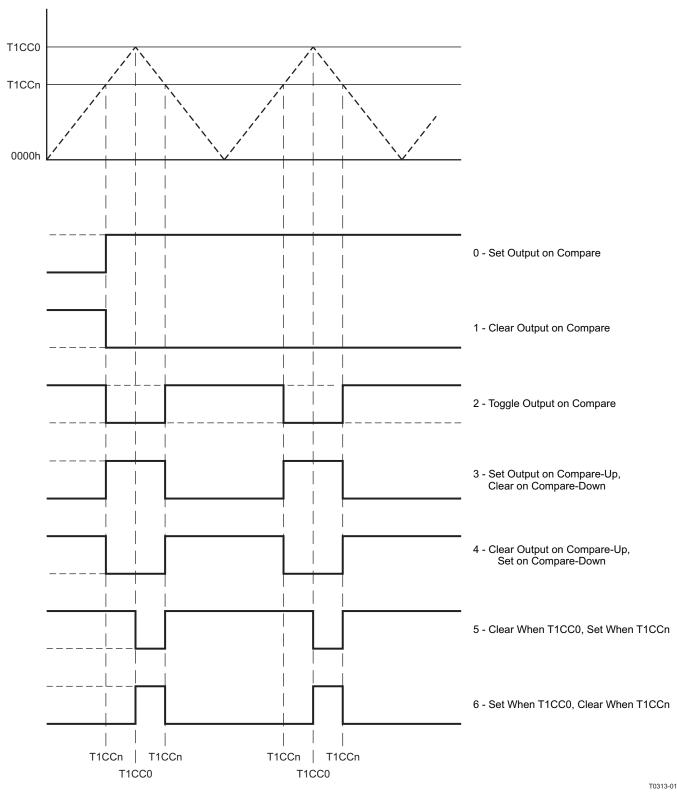


Figure 9-5. Output Compare Modes, Timer Modulo Mode



#### Output Compare Mode

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# 9.9 IR Signal Generation and Learning

This section describes how Timer 1 can be configured in IR generation mode, where it counts Timer 3 periods and the output is ANDed with the output of Timer 3 to generate modulated consumer IR signals with minimal CPU interaction.

# 9.9.1 Introduction

Generation of IR signals for remote control is generally done in one of two ways:

- Modulated codes
- Non-modulated codes (C-codes, flash codes)

The device includes flexible timer functionality to implement generation and learning of both types of IR signals with minimal CPU interaction. Most IR protocols can be implemented with only one CPU intervention per command.

# 9.9.2 Modulated Codes

Modulated codes can be generated using Timer 1 (16-bit) and Timer 3 (8-bit). Timer 3 in modulo mode is used to generate the carrier. Timer 3 has an individual prescaler for its input. Its period is set using T3CC0. Timer 3 channel 1 is used for PWM output. The duty cycle of the carrier is set using T3CC1. Channel 1 uses compare mode: *Clear output on compare, set on 0x00* (T3CCTL1.CMP = 100). Table 9-2 shows the frequency error calculation for a 38-kHz carrier using Timer 3.

Description	Value		
System clock frequency	32,000 kHz		
IR carrier frequency	38 kHz		
System clock period	0.00003125 ms		
IR carrier period	0.026315789 ms		
Timer prescaler	4		
Timer period	0.000125 ms		
Ideal timer value	210.5263158		
True timer value	211		
True timer period	0.026375 ms		
True timer frequency	37.91469194 kHz		
Period error	59.21052632 ns		
Frequency error	85.30805687 Hz		
Frequency error %	0.2245%		

# Table 9-2. Frequency Error Calculation for 38-kHz Carrier

The IRCTL.IRGEN register bit enables IR generation mode in Timer 1. When the IRGEN bit is set, Timer 1 takes the output of the Timer 3 channel 1 compare signal as tick instead of the system tick. The Timer 1 period is set using T1CC0 with Timer 1 in modulo mode (T1CTL.MODE = 10) and channel 0 in compare mode (T1CCTL0.MODE = 1). Channel 1 compare mode *Clear output on compare, set on 0x0000* (T1CCTL1.CMP = 100) is used for output of the gating signal.

The number of *mark* carrier periods is set by T1CC1. T1CC1 must be updated every Timer 1 period by the DMA or CPU. Note that an update to T1CC1 is buffered and does not take effect before Timer 1 reaches 0x0000.

The number of *space* carrier periods is set by T1CC0. Its value should be set to the total number of *mark* and *space* carrier periods wanted. The compare values are buffered until the timer hits 0x0000.

The output of Timer 1 channel 1 is ANDed with that of Timer 3 channel 1 to form the IR output as shown in Figure 9-7



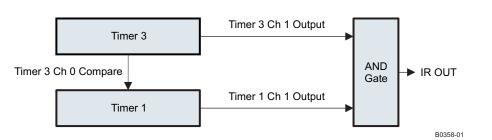


Figure 9-7. Block Diagram of Timers in IR-Generation Mode

The timing of the Timer 3 channel 1 output and Timer 1 channel 1 output signals is synchronized such that no glitches are produced on the IR Out signal.

When the IRGEN bit is set, the IR out signal is routed to pins instead of the normal Timer 1 channel 1 output (see also Section 7.6.1).

Figure 9-8 shows the example of Timer 3 being initialized to a 33% duty cycle (T3CC0 =  $3 \times T3CC1$ ). Timer 1 has been initialized to 3.

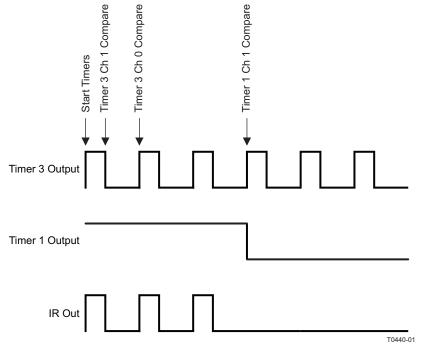


Figure 9-8. Modulated Waveform Example

To achieve a period of *space* only, TICC1 should be set to 0x00.

# 9.9.3 Non-Modulated Codes

To generate non-modulated IR codes, Timer 1 is used in modulo mode. The period of the signal is given by TICCO, and the pulse duration is given by TICCI. TICCI gives the length of the *mark* period, and TICCO gives the total number of *mark* and *space* periods. The compare values are buffered until the timer hits 0x0000. The compare values must be updated once every period by the DMA or CPU if they are not to be kept the same.



# 9.9.4 Learning

Learning is done by using the capture function of Timer 1 (16-bit) and Timer 3 (8-bit). Timer 3 can handle the carrier frequency detection and Timer 1 can handle the code learning from the demodulated signal. The circuit could be set up as described in Figure 9-9

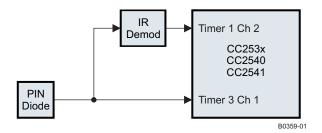


Figure 9-9. IR Learning Board Diagram

# 9.9.4.1 Carrier Frequency Detection

Timer 3 is used to capture and detect the carrier frequency with input directly from the IR PIN diode. The timer should sample the carrier a limited number of times. If a carrier is detected, the frequency detected should contribute to the average number, which is what can be stored in the database.

# 9.9.4.2 Demodulated Code Learning

The output from the IR PIN diode is demodulated by an appropriate circuit. The output from this circuit is used as input to one of the Timer 1 channels in capture mode.

# 9.9.5 Other Considerations

The IR output pin should be placed in the high-impedance state or pulled down during reset to avoid unnecessary power consumption from illuminating the IR LED. Note that only the P1.1 output for Timer 1 channel 1 is placed in the high-impedance state with no pullup during and after reset.

# 9.10 Timer 1 Interrupts

One interrupt vector is assigned to the timer. An interrupt request is generated when one of the following timer events occurs:

- Counter reaches terminal count value (overflow, or turns around zero).
- Input capture event
- Output compare event

The status register, T1STAT, contains the source interrupt flags for the terminal-count value event and the five channel compare/capture events. A source interrupt flag is set when the corresponding event occurs, regardless of interrupt mask bits. The CPU interrupt flag IRCON.T1IF is set when one of the events occurs if the corresponding interrupt mask bit is equal to 1. The interrupt mask bits are T1CCTLn.IM for the five channels and TIMIF.OVFIM for the overflow event. The CPU interrupt flag IRCON.T1IF is also set when a Timer 1 source interrupt flag is being cleared and one or more other Timer 1 source interrupt flags are still set while the corresponding interrupt mask bit is set. An interrupt request is generated when IRCON.T1IF goes from 0 to 1 if IEN1.T1IEN and IEN0.EA are both equal to 1.

# 9.11 Timer 1 DMA Triggers

There are three DMA triggers associated with Timer 1. These are DMA triggers T1 CH0, T1 CH1, and T1_CH2, which are generated on timer compare events as follows:

- T1 CH0 Channel 0 compare
- T1_CH1 Channel 1 compare
- T1 CH2 Channel 2 compare



There are no triggers associated with channels 3 and 4.

# 9.12 Timer 1 Registers

This section describes the Timer 1 registers, which consist of the following registers:

- T1CNTH Timer 1 count high
- T1CNTL Timer 1 count low
- T1CTL Timer 1 control
- T1STAT Timer 1 status
- T1CCTLn Timer 1 channel n capture/compare control
- T1CCnH Timer 1 channel n capture/compare value high
- T1CCnL Timer 1 channel n capture/compare value low

The TIMIF.OVFIM register bit resides in the TIMIF register, which is described together with the Timer 3 and Timer 4 registers.

#### T1CNTH (0xE3) – Timer 1 Counter High

Bit	Name	Reset	R/W	Description
7:0	CNT[15:8]	0x00	R	Timer count high-order byte. Contains the high byte of the 16-bit timer counter buffered at the time T1CNTL is read

#### T1CNTL (0xE2) - Timer 1 Counter Low

Bit	Name	Reset	R/W	Description
7:0	CNT[7:0]	0x00	R/W	Timer count low-order byte. Contains the low byte of the 16-bit timer counter. Writing anything to this register results in the counter being cleared to 0x0000 and initializes all output pins of associated channels.

#### T1CTL (0xE4) - Timer 1 Control

Bit	Name	Reset	R/W	Description
7:4	_	0000	R0	Reserved
3:2	DIV[1:0]	00	R/W	Prescaler divider value. Generates the active clock edge used to update the counter as follows:
				00: Tick frequency/1
				01: Tick frequency/8
				10: Tick frequency/32
				11: Tick frequency/128
1:0	MODE [1:0]	00	R/W	Timer 1 mode select. The timer operating mode is selected as follows:
				00: Operation is suspended.
				01: Free-running, repeatedly count from 0x0000 to 0xFFFF.
				10: Modulo, repeatedly count from 0x0000 to TICC0.
				11: Up/down, repeatedly count from 0x0000 to TICC0 and from TICC0 down to 0x0000.

T1S	TAT (0xAF)	– Timer 1	Status	
Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5	OVFIF	0	R/W0	Timer 1 counter-overflow interrupt flag. Set when the counter reaches the terminal count value in free-running or modulo mode, and when zero is reached counting down in up-down mode. Writing a 1 has no effect.
4	CH4IF	0	R/W0	Timer 1 channel 4 interrupt flag. Set when the channel 4 interrupt condition occurs. Writing a 1 has no effect.
3	CH3IF	0	R/W0	Timer 1 channel 3 interrupt flag. Set when the channel 3 interrupt condition occurs. Writing a 1 has no effect.
2	CH2IF	0	R/W0	Timer 1 channel 2 interrupt flag. Set when the channel 2 interrupt condition occurs. Writing a 1 has no effect.
1	CH1IF	0	R/W0	Timer 1 channel 1 interrupt flag. Set when the channel 1 interrupt condition occurs. Writing a 1 has no effect.
0	CH0IF	0	R/W0	Timer 1 channel 0 interrupt flag. Set when the channel 0 interrupt condition occurs. Writing a 1 has no effect.

# T1CCTL0 (0xE5) – Timer 1 Channel 0 Capture/Compare Control

Bit	Name	Reset	R/W	Description		
7	RFIRQ	0	R/W	When set, use RF interrupt for capture instead of regular capture input.		
6	IM	1	R/W	Channel 0 interrupt mask. Enables interrupt request when set.		
5:3	CMP[2:0]	000	R/W	Channel 0 compare-mode select. Selects action on output when timer value equals compare value in ${\tt T1CC0}$		
				000: Set output on compare		
				001: Clear output on compare		
				010: Toggle output on compare		
				011: Set output on compare-up, clear on 0		
				100: Clear output on compare-up, set on 0		
				101: Reserved		
				110: Reserved		
				111: Initialize output pin. CMP[2:0] is not changed.		
2	MODE	0	R/W	Mode. Select Timer 1 channel 0 capture or compare mode		
				0: Capture mode		
				1: Compare mode		
1:0	CAP[1:0]	00	R/W	Channel 0 capture-mode select		
				00: No capture		
				01: Capture on rising edge		
				10: Capture on falling edge		
				11: Capture on all edges		

#### T1CC0H (0xDB) - Timer 1 Channel 0 Capture/Compare Value, High

Bit	Name	Reset	R/W	Description
7:0	T1CC0[15:8]	0x00	R/W	Timer 1 channel 0 capture/compare value high-order byte. Writing to this register when T1CCTL0.MODE = 1 (compare mode) causes the T1CC0[15:0] update to the written value to be delayed until T1CNT = 0x0000.

#### T1CC0L (0xDA) – Timer 1 Channel 0 Capture/Compare Value, Low

Bit	Name	Reset	R/W	Description
7:0	T1CC0[7:0]	0x00		Timer 1 channel 0 capture/compare value low-order byte. Data written to this register is stored in a buffer but not written to T1CC0[7:0] until, and at the same time as, a later write to T1CC0H takes effect.



Timer 1 Registers

Bit	Name	Reset	R/W	Description		
7	RFIRQ	0	R/W	When set, use RF interrupt for capture instead of regular capture input.		
6	IM	1	R/W	Channel 1 interrupt mask. Enables interrupt request when set.		
5:3	CMP[2:0]	000	R/W	Channel 1 compare-mode select. Selects action on output when timer value equals compare value in T1CC1.		
				000: Set output on compare		
				001: Clear output on compare		
				010: Toggle output on compare		
				011: Set output on compare-up, clear on compare-down in up-down mode. Otherwise set output on compare, clear on 0.		
				100: Clear output on compare-up, set on compare-down in up-down mode. Otherwise clear output on compare, set on 0.		
				101: Clear when equal T1CC0, set when equal T1CC1		
				110: Set when equal T1CC0, clear when equal T1CC1		
				111: Initialize output pin. CMP[2:0] is not changed.		
2	MODE	0	R/W	Mode. Select Timer 1 channel 1 capture or compare mode		
				0: Capture mode		
				1: Compare mode		
1:0	CAP[1:0]	00	R/W	Channel 1 capture-mode select		
				00: No capture		
				01: Capture on rising edge		
				10: Capture on falling edge		
				11: Capture on all edges		

# T1CCTL1 (0xE6) – Timer 1 Channel 1 Capture/Compare Control

#### T1CC1H (0xDD) – Timer 1 Channel 1 Capture/Compare Value, High

Bit	Name	Reset	R/W	Description
7:0	T1CC1[15:8]	0x00	R/W	Timer 1 channel 1 capture/compare value high-order byte. Writing to this register when T1CCTL1.MODE = 1 (compare mode) causes the T1CC1[15:0] update to the written value to be delayed until T1CNT = 0x0000.

#### T1CC1L (0xDC) – Timer 1 Channel 1 Capture/Compare Value, Low

Bit	Name	Reset	R/W	Description
7:0	T1CC1[7:0]	0x00	R/W	Timer 1 channel 1 capture/compare value low-order byte. Data written to this register is stored in a buffer but not written to T1CC1[7:0] until, and at the same time as, a later write to T1CC1H takes effect.



Bit	Name	Reset	R/W	Description
7	RFIRQ	0	R/W	When set, use RF interrupt for capture instead of regular capture input.
6	IM	1	R/W	Channel 2 interrupt mask. Enables interrupt request when set.
5:3	CMP[2:0]	000	R/W	Channel 2 compare mode select. Selects action on output when timer value equals compare value in ${\tt T1CC2}$ .
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set output on compare-up, clear on compare-down in up-down mode. Otherwise set output on compare, clear on 0.
				100: Clear output on compare-up, set on compare-down in up-down mode. Otherwise clear output on compare, set on 0.
				101: Clear when equal T1CC0, set when equal T1CC2
				110: Set when equal T1CC0, clear when equal T1CC2
				111: Initialize output pin. CMP[2:0] is not changed.
2	MODE	0	R/W	Mode. Select Timer 1 channel 2 capture or compare mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Channel 2 capture-mode select
				00: No capture
				01: Capture on rising edge
				10: Capture on falling edge
				11: Capture on all edges

# T1CCTL2 (0xE7) – Timer 1 Channel 2 Capture/Compare Control

#### T1CC2H (0xDF) - Timer 1 Channel 2 Capture/Compare Value, High

Bit	Name	Reset	R/W	Description
7:0	T1CC2[15:8]	0x00	R/W	Timer 1 channel 2 capture/compare value high-order byte. Writing to this register when T1CCTL2.MODE = 1 (compare mode) causes the T1CC2[15:0] update to the written value to be delayed until T1CNT = 0x0000.

#### T1CC2L (0xDE) – Timer 1 Channel 2 Capture/Compare Value, Low

Bit	Name	Reset	R/W	Description
7:0	T1CC2[7:0]	0x00	R/W	Timer 1 channel 2 capture/compare value low-order byte. Data written to this register is stored in a buffer but not written to T1CC2[7:0] until, and at the same time as, a later write to T1CC2H takes effect.

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Timer 1 Registers

Bit	Name	Reset	R/W	Description
7	RFIRQ	0	R/W	When set, use RF interrupt for capture instead of regular capture input.
6	IM	1	R/W	Channel 3 interrupt mask. Enables interrupt request when set.
5:3	CMP[2:0]	000	R/W	Channel 3 compare mode select. Selects action on output when timer value equals compare value in T1CC3.
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set output on compare-up, clear on compare-down in up-down mode. Otherwise set output on compare, clear on 0.
				100: Clear output on compare-up, set on compare down in up-down mode. Otherwise clear output on compare, set on 0.
				101: Clear when equal T1CC0, set when equal T1CC3
				110: Set when equal T1CC0, clear when equal T1CC3
				111: Initialize output pin. CMP[2:0] is not changed.
2	MODE	0	R/W	Mode. Select Timer 1 channel 3 capture or compare mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Channel 3 capture-mode select
				00: No capture
				01: Capture on rising edge
				10: Capture on falling edge
				11: Capture on all edges

#### T1CCTL3 (0x62A3) – Timer 1 Channel 3 Capture/Compare Control

#### T1CC3H (0x62AD) – Timer 1 Channel 3 Capture/Compare Value, High

Bit	Name	Reset	R/W	Description
7:0	T1CC3[15:8]	0x00	R/W	Timer 1 channel 3 capture/compare value high-order byte. Writing to this register when T1CCTL3.MODE = 1 (compare mode) causes the T1CC3[15:0] update to the written value to be delayed until T1CNT = 0x0000.

#### T1CC3L (0x62AC) – Timer 1 Channel 3 Capture/Compare Value, Low

Bit	Name	Reset	R/W	Description
7:0	T1CC3[7:0]	0x00		Timer 1 channel 3 capture/compare value low-order byte. Data written to this register is stored in a buffer but not written to T1CC3[7:0] until, and at the same time as, a later write to T1CC3H takes effect.



Bit	Name	Reset	R/W	Description
7	RFIRQ	0	R/W	When set, use RF interrupt for capture instead of regular capture input.
6	IM	1	R/W	Channel 4 interrupt mask. Enables interrupt request when set.
5:3	CMP[2:0]	000	R/W	Channel 4 compare mode select. Selects action on output when timer value equals compare value in ${\tt T1CC4}$ .
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set output on compare-up, clear on compare down in up-down mode. Otherwise set output on compare, clear on 0.
				100: Clear output on compare-up, set on compare down in up-down mode. Otherwise clear output on compare, set on 0.
				101: Clear when equal T1CC0, set when equal T1CC4
				110: Set when equal T1CC0, clear when equal T1CC4
				111: Initialize output pin. CMP[2:0] is not changed.
2	MODE	0	R/W	Mode. Select Timer 1 channel 4 capture or compare mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Channel 4 capture-mode select
				00: No capture
				01: Capture on rising edge
				10: Capture on falling edge
				11: Capture on all edges

#### T1CCTL4 (0x62A4) – Timer 1 Channel 4 Capture/Compare Control

#### T1CC4H (0x62AF) - Timer 1 Channel 4 Capture/Compare Value, High

Bit	Name	Reset	R/W	Description
7:0	T1CC4[15:8]	0x00	R/W	Timer 1 channel 4 capture/compare value high-order byte. Writing to this register when T1CCTL4.MODE = 1 (compare mode) causes the T1CC4[15:0] update to the written value to be delayed until T1CNT = 0x0000.

#### T1CC4L (0x62AE) – Timer 1 Channel 4 Capture/Compare Value, Low

Bit	Name	Reset	R/W	Description
7:0	T1CC4[7:0]	0x00	R/W	Timer 1 channel 4 capture/compare value low-order byte. Data written to this register is stored in a buffer but not written to T1CC4[7:0] until, and at the same time as, a later write to T1CC4H takes effect.

#### IRCTL (0x6281) - Timer 1 IR Generation Control

Bit	Name	Reset	R/W	Description
7:1	-	0000 000	R/W	Reserved
0	IRGEN	0	R/W	When this bit is set, a connection between Timer 3 channel 1 and Timer 1 tick input is made so that the timers can be used to generate modulated IR codes (see also Section 9.9).

# 9.13 Accessing Timer 1 Registers as Array

The Timer 1 capture/compare channel registers can be accessed as a contiguous region in the XDATA memory space. This facilitates accessing the registers as a simple indexed structure. The five capture/compare control registers are mapped to 0x62A0–0x62A4. The 16-bit capture/compare values are mapped to 0x62A6–0x62AF; 0x62A5 is unused.



Page

# Timer 3 and Timer 4 (8-Bit Timers)

Timer 3 and Timer 4 are two 8-bit timers. Each timer has two independent capture/compare channels, each using one I/O pin per channel.

Features of Timer 3 and Timer 4 are as follows:

- · Two capture/compare channels
- Set, clear, or toggle output compare
- Clock prescaler for divide by 1, 2, 4, 8, 16, 32, 64, 128
- Interrupt request generated on each capture/compare and terminal-count event
- DMA trigger function

#### Topic

# 10.1 8-Bit Timer Counter 126 10.2 Timer 3/Timer 4 Mode Control 126 10.3 Channel Mode Control 126 10.4 Input Capture Mode 127 10.5 Output Compare Mode 127 10.6 Timer 3 and Timer 4 Interrupts 127 10.7 Timer 3 and Timer 4 DMA Triggers 128 10.8 Timer 3 and Timer 4 Registers 128

#### **10.1 8-Bit Timer Counter**

All timer functions are based on the main 8-bit counter found in Timer 3 and Timer 4. The counter increments or decrements at each active clock edge. The period of the active clock edges, as defined by the register bits CLKCONCMD.TICKSPD[2:0], is further multiplied (the frequency is divided) by the prescaler value set by TxCTL.DIV[2:0] (where x refers to the timer number, 3 or 4). The counter operates as either a free-running counter, a down counter, a modulo counter, or an up/down counter.

It is possible to read the 8-bit counter value through the SFR register TxCNT, where x refers to the timer number, 3 or 4.

The possibility to clear and halt the counter is given with TxCTL control register settings. The counter is started when a 1 is written to TxCTL. START. If a 0 is written to TxCTL. START, the counter halts at its present value.

# 10.2 Timer 3/Timer 4 Mode Control

In general, the control register TxCTL is used to control the timer operation.

#### 10.2.1 Free-Running Mode

In the free-running mode of operation, the counter starts from 0x00 and increments at each active clock edge. When the counter reaches 0xFF, the counter is loaded with 0x00 and continues incrementing its value. When the terminal count value 0xFF is reached (i.e., an overflow occurs), the interrupt flag TIMIF.TxOVFIF is set. An interrupt request is generated if enabled, see Section 10.6 for details. The free-running mode can be used to generate independent time intervals and output-signal frequencies.

#### 10.2.2 Down Mode

In the down mode, after the timer has been started, the counter is loaded with the contents in TxCC0. The counter then counts down to 0x00. The interrupt flag TIMIF.TxOVFIF is set when 0x00 is reached. An interrupt request is generated if enabled, see Section 10.6 for details. The timer down mode can generally be used in applications where an event timeout interval is required.

#### 10.2.3 Modulo Mode

When the timer operates in modulo mode, the 8-bit counter starts at 0x00 and increments at each active clock edge. After the count has reached the period value held in register TxCC0, the counter is reset to 0x00 and continues to increment. If the timer started with a value above TxCC0, the interrupt flag TIMIF.TxOVFIF is set when the terminal value (0xFF) is reached, after which the counter wraps to 0x00. An interrupt request is generated if enabled, see Section 10.6 for details. If a periodic interrupt is wanted at the period value, this can be obtained by enabling an output compare interrupt on channel 0, as explained in Section 10.5. The modulo mode can be used for applications where a period other than 0xFF is required.

#### 10.2.4 Up/Down Mode

In the up/down timer mode, the counter repeatedly starts from 0x00 and counts up until the value held in TxCC0 is reached, and then the counter counts down until 0x00 is reached. This timer mode is used when symmetrical output pulses are required with a period other than 0xFF, allowing implementation of center-aligned PWM output applications. The interrupt flag TIMIF.TxOVFIF is set when the counter value reaches 0x00 in the up/down mode. An interrupt request is generated if enabled, see Section 10.6 for details.

Clearing the counter by writing to TxCTL.CLR also resets the count direction to the count-up-from-0x00 mode.

#### 10.3 Channel Mode Control

The channel modes for each channel, 0 and 1, are set by the control and status registers TxCCTLn, where n is the channel number, 0 or 1. The settings include capture and compare modes.



# 10.4 Input Capture Mode

When a channel is configured as an input capture channel, the I/O pin associated with that channel is configured as an input. After the timer has been started, a rising edge, falling edge, or any edge on the input pin triggers a capture of the 8-bit counter contents into the associated capture register. Thus, the timer is able to capture the time when an external event takes place.

**NOTE:** Before an I/O pin can be used by the timer, the required I/O pin must be configured as a Timer 3/Timer 4 peripheral pin.

The channel input pin is synchronized to the internal system clock. Thus, pulses on the input pin must have a minimum duration greater than the system clock period.

The content of the 8-bit capture register for channel n is read out from register T3CCn/T4CCn.

When the capture takes place, the interrupt flag for the channel, TIMIF.TxCHnIF (x is 3 or 4, n is the channel number), is set. An interrupt request is generated if enabled, see Section 10.6 for details.

# 10.5 Output Compare Mode

In output-compare mode, the I/O pin associated with a channel must be set to an output. After the timer has been started, the content of the counter is compared with the contents of channel compare register TxCCOn. If the compare register equals the counter contents, the output pin is set, reset, or toggled according to the compare output mode setting of TxCCTL.CMP1:0. Note that all edges on output pins are glitch-free when operating in a given compare output mode.

For simple PWM use, output compare modes 4 and 5 are preferred.

Writing to compare register TxCC0 or TxCC1 does not take effect on the output compare value until the counter value is 0x00.

When the capture takes place, the interrupt flag for the channel, TIMIF.TxCHnIF (x is 3 or 4, n is the channel number), is set. An interrupt request is generated if enabled, see Section 10.6 for details.

A compare output pin is initialized to the value listed in Table 9-1 when:

- a 1 is written to TxCNTR.CLR (All Timer x channels)
- 0x7 is written to TxCCTLn.CMP (Timer x, channel n)

#### Table 10-1. Initial Compare Output Values (Compare Mode)

Compare Mode (TxCCTLn.CMP)	Initial Compare Output
Set output on compare (000)	0
Clear output on compare (001)	1
Toggle output on compare (010)	0
Set output on compare-up, clear on compare-down in up-down mode (011)	0
In other modes than up-down mode, set output on compare, clear on 0 (011)	0
Clear output on compare-up, set on compare-down in up-down mode (100)	1
In other modes than up-down mode, clear output on compare, set on 0 (100)	1
Set output on compare, clear on 0xFF (101)	0
Clear output on compare, set on 0x00 (110)	1

# 10.6 Timer 3 and Timer 4 Interrupts

One interrupt vector is assigned to each of the timers. These are T3 and T4. An interrupt request is generated when one of the following timer events occurs:

- Counter reaches terminal count value.
- Compare event
- Capture event

Input Capture Mode



#### Timer 3 and Timer 4 DMA Triggers

The SFR register TIMIF contains all interrupt flags for Timer 3 and Timer 4. The register bits TIMIF.TxOVFIF and TIMIF.TxCHnIF contain the source interrupt flags for the two terminal-count value events and the four channel-compare events, respectively. A source interrupt flag is set when the corresponding event occurs, regardless of interrupt mask bits. The CPU interrupt flag IRCON.T3IF or IRCON.T4IF is set when one of the events occurs if the corresponding interrupt mask bit is equal to 1. The interrupt mask bits are TxCCTLn.IM for the four channels and TxCTL.OVFIM for the overflow events. The CPU interrupt flag IRCON.T3IF or IRCON.T4IF is also set when a Timer 3 or Timer 4 source interrupt flag is being cleared and one or more other source interrupt flags for the same timer are still set while the corresponding interrupt mask bit is set. An interrupt request is generated when IRCON.TxIF goes from 0 to 1 if IEN1.TxIEN and IEN0.EA are both equal to 1 (x is 3 or 4).

# 10.7 Timer 3 and Timer 4 DMA Triggers

Two DMA triggers are associated with Timer 3, and two DMA triggers are associated with Timer 4.

- T3_CH0: Timer 3 channel 0 capture/compare
- T3_CH1: Timer 3 channel 1 capture/compare
- T4_CH0: Timer 4 channel 0 capture/compare
- T4_CH0: Timer 4 channel 1 capture/compare

# 10.8 Timer 3 and Timer 4 Registers

#### T3CNT (0xCA) – Timer 3 Counter

	()			
Bit	Name	Reset	R/W	Description
7:0	CNT[7:0]	0x00	R	Timer count byte. Contains the current value of the 8-bit counter

#### T3CTL (0xCB) - Timer 3 Control

Bit	Name	Reset	R/W	Description
7:5	DIV[2:0]	000	R/W	Prescaler divider value. Generates the active clock edge used to clock the timer from CLKCONCMD.TICKSPD as follows:
				000: Tick frequency/1
				001: Tick frequency/2
				010: Tick frequency/4
				011: Tick frequency/8
				100: Tick frequency16
				101: Tick frequency /32
				110: Tick frequency/64
				111: Tick frequency/128
4	START	0	R/W	Start timer. Normal operation when set, suspended when cleared
3	OVFIM	1	R/W	Overflow interrupt mask
				0: Interrupt is disabled.
				1: Interrupt is enabled.
2	CLR	0	R0/W1	Clear counter. Writing a 1 to CLR resets the counter to 0x00 and initializes all output pins of associated channels. Always read as 0.
1:0	MODE[1:0]	00	R/W	Timer 3 mode. Select the mode as follows:
				00: Free-running, repeatedly count from 0x00 to 0xFF
				01: Down, count from T3CC0 to 0x00
				10: Modulo, repeatedly count from 0x00 to T3CC0
				11: Up/down, repeatedly count from 0x00 to T3CC0 and down to 0x00



Bit	Name	Reset	R/W	Description
7	_	0	R0	Reserved
6	IM	1	R/W	Channel 0 interrupt mask
				0: Interrupt is disabled.
				1: Interrupt is enabled.
5:3	CMP[2:0]	000	R/W	Channel 0 compare output mode select. Specified action occurs on output when timer value equals compare value in T3CC0
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set output on compare-up, clear on 0
				100: Clear output on compare-up, set on 0
				101: Set output on compare, clear on 0xFF
				110: Clear output on compare, set on 0x00
				111: Initialize output pin. CMP[2:0] is not changed.
2	MODE	0	R/W	Mode. Select Timer 3 channel 0 mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Capture mode select
				00: No capture
				01: Capture on rising edge
				10: Capture on falling edge
				11: Capture on both edges

# T3CCTL0 (0xCC) – Timer 3 Channel 0 Capture/Compare Control

# T3CC0 (0xCD) – Timer 3 Channel 0 Capture/Compare Value

Bit	Name	Reset	R/W	Description
7:0	VAL[7:0]	0x00	R/W	Timer capture/compare value channel 0. Writing to this register when T3CCTL0.MODE=1 (compare mode) causes the T3CC0.VAL[7:0] update to the written value to be delayed until T3CNT.CNT[7:0]=0x00.



#### Timer 3 and Timer 4 Registers

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	IM	1	R/W	Channel 1 interrupt mask
				0: Interrupt is disabled.
				1: Interrupt is enabled.
5:3	CMP[2:0]	000	R/W	Channel 1 compare output-mode select. Specified action on output when timer value equals compare value in T3CC1
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set on compare-up, clear on compare-down in up-down mode. Otherwise, set output on compare, clear on 0.
				100: Clear output on compare-up, set on compare-down in up-down mode. Otherwise clear output on compare, set on 0.
				101: Set output on compare, clear on 0xFF
				110: Clear output on compare, set on 0x00
				111: Initialize output pin. CMP[2:0] is not changed
2	MODE	0	R/W	Mode. Select Timer 3 channel 1 mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Capture mode select
				00: No capture
				01: Capture on rising edge
	1			10: Capture on falling edge
	1			11: Capture on both edges

# T3CCTL1 (0xCE) – Timer 3 Channel 1 Capture/Compare Control

# T3CC1 (0xCF) – Timer 3 Channel 1 Capture/Compare Value

Bit	Name	Reset	R/W	Description
7:0	VAL[7:0]	0x00	R/W	Timer 3 capture/compare value, channel 1. Writing to this register when T3CCTL1.MODE = 1 (compare mode) causes the T3CC1.VAL[7:0] update to the written value to be delayed until T3CNT.CNT[7:0] = 0x00.

#### T4CNT (0xEA) - Timer 4 Counter

Bit	Name	Reset	R/W	Description
7:0	CNT[7:0]	0x00	R	Timer count byte. Contains the current value of the 8-bit counter



Bit	Name	Reset	R/W	Description
7:5	DIV[2:0]	000	R/W	Prescaler divider value. Generates the active clock edge used to clock the timer from CLKCONCMD.TICKSPD as follows:
				000: Tick frequency/1
				001: Tick frequency/2
				010: Tick frequency/4
				011: Tick frequency/8
				100: Tick frequency/16
				101: Tick frequency/32
				110: Tick frequency/64
				111: Tick frequency/128
4	START		R/W	Start timer. Normal operation when set, suspended when cleared
3	OVFIM	1	R/W	Overflow interrupt mask
2	CLR	0	R0/W1	Clear counter. Writing a 1 to CLR resets the counter to 0x00 and initialize all output pins of associated channels. Always read as 0.
1:0	MODE[1:0]	0	R/W	Timer 4 mode. Select the mode as follows:
				00: Free running, repeatedly count from 0x00 to 0xFF
				01: Down, count from T4CC0 to 0x00
				10: Modulo, repeatedly count from 0x00 to T4CC0
				11: Up/down, repeatedly count from 0x00 to T4CC0 and down to 0x00

# T4CCTL0 (0xEC) – Timer 4 Channel 0 Capture/Compare Control

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	IM	1	R/W	Channel 0 interrupt mask
5:3	CMP[2:0]	000	R/W	Channel 0 compare output-mode select. Specified action occurs on output when timer value equals compare value in T4CC0.
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set output on compare-up, clear on 0
				100: Clear output on compare-up, set on 0
				101: Set output on compare, clear on 0xFF
				110: Clear output on compare, set on 0x00
				111: Initialize output pin. CMP[2:0] is not changed
2	MODE	0	R/W	Mode. Select Timer 4 channel 0 mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Capture mode select. 00 – No capture, 01 – Capture on rising edge, 10 – Capture on falling edge, 11 – Capture on both edges

#### T4CC0 (0xED) - Timer 4 Channel 0 Capture/Compare Value

Bit	Name	Reset	R/W	Description
7:0	VAL[7:0]	0x00	R/W	Timer 4 capture/compare value, channel 0. Writing to this register when T4CCTL0.MODE = 1 (compare mode) causes the T4CC0.VAL[7:0] update to the written value to be delayed until T4CNT.CNT[7:0] = 0x00.



#### Timer 3 and Timer 4 Registers

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	IM	1	R/W	Channel 1 interrupt mask
5:3	CMP[2:0]	000	R/W	Channel 1 compare output-mode select. Specified action on output when timer value equals compare value in T4CC1
				000: Set output on compare
				001: Clear output on compare
				010: Toggle output on compare
				011: Set on compare-up, clear on compare-down in up-down mode. Otherwise, set output or compare, clear on 0.
				100: Clear output on compare-up, set on compare-down in up-down mode. Otherwise clear output on compare, set on 0.
				101: Set output on compare, clear on 0xFF
				110: Clear output on compare, set on 0x00
				111: Initialize output pin. CMP[2:0] is not changed.
2	MODE	0	R/W	Mode. Select Timer 4 channel 1 mode
				0: Capture mode
				1: Compare mode
1:0	CAP[1:0]	00	R/W	Capture mode select. 00 – No Capture, 01 – Capture on rising edge, 10 – Capture on falling edge, 11 – Capture on both edges

#### T4CCTL1 (0xEE) – Timer 4 Channel 1 Capture/Compare Control

#### T4CC1 (0xEF) – Timer 4 Channel 1 Capture/Compare Value

Bit	Name	Reset	R/W	Description
7:0	VAL[7:0]	0x00	R/W	Timer capture/compare value, channel 1. Writing to this register when T4CCTL1.MODE = 1 (compare mode) causes the T4CC1.VAL[7:0] update to the written value to be delayed until T4CNT.CNT[7:0] = 0x00.

#### TIMIF (0xD8) - Timer 1/3/4 Interrupt Mask/Flag

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	OVFIM	1	R/W	Timer 1 overflow interrupt mask
5	T4CH1IF	0	R/W0	Timer 4 channel 1 interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.
4	T4CH0IF	0	R/W0	Timer 4 channel 0 interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.
3	T40VFIF	0	R/W0	Timer 4 overflow interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.
2	T3CH11F	0	R/W0	Timer 3 channel 1 interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.
1	T3CH0IF	0	R/W0	Timer 3 channel 0 interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.
0	T30VFIF	0	R/W0	Timer 3 overflow interrupt flag
				0: No interrupt is pending.
				1: Interrupt is pending.



# Sleep Timer

Page

The Sleep Timer is used to set the period during which the system enters and exits low-power modes PM1 and PM2. The Sleep Timer is also used to maintain timing in Timer 2 when entering power mode PM1 or PM2.

The main features of the Sleep Timer are the following:

- 24-bit timer up-counter operating at 32-kHz clock rate
- 24-bit compare with interrupt and DMA trigger
- 24-bit capture

#### Topic

 11.1
 General
 134

 11.2
 Timer Compare
 134

 11.3
 Timer Capture
 134

 11.4
 Sleep Timer Registers
 135



General

# 11.1 General

The Sleep Timer is a 24-bit timer running on the 32-kHz clock (either RCOSC or XOSC). The timer starts running immediately after a reset and continues to run uninterrupted.

The current value of the timer can be read from SFR registers ST2:ST1:ST0. When ST0 is read, the current value of the 24-bit counter is latched. Thus, the ST0 register must be read before ST1 and ST2 to read a correct Sleep Timer count value.

The Sleep Timer is running when operating in all power modes except PM3. The value of the Sleep Timer is not preserved in PM3. When returning from PM1 or PM2 (where the system clock is shut down), the Sleep Timer value in ST2:ST1:ST0 is not up-to-date until a positive edge on the 32-kHz clock has been detected after the system clock restarted. To ensure an updated value is read, wait for a positive transition on the 32-kHz clock by polling the SLEEPSTA.CLK32K bit, before reading the Sleep Timer value.

Note that if supply voltage drops below 2 V while in PM2, the sleep interval might be affected.

#### 11.2 Timer Compare

A timer compare event occurs when the timer value is equal to the 24-bit compare value and there is a positive edge on the 32-kHz clock. The compare value is set by writing to registers ST2:ST1:ST0. Writing to ST0 while STLOAD.LDRDY is 1 initiates loading of the new compare value, i.e., the most-recent values written to the ST2, ST1, and ST0 registers. This means that when writing a compare value, ST2 and ST1 must be written before ST0.STLOAD.LDRDY is 0 during the load, and software must not start a new load until STLOAD.LDRDY has flipped back to 1.

When setting a new compare value, the value should be at least 5 more than the current sleep timer value. Otherwise, the timer compare event may be lost.

The interrupt enable bit for the ST interrupt is IEN0.STIE, and the interrupt flag is IRCON.STIF. When a timer compare event occurs, the interrupt flag IRCON.STIF is asserted.

In PM1 and PM2, the Sleep Timer compare event may be used to wake up the device and return to active operation in active mode. The default value of the compare value after reset is 0xFF FFFF.

For all devices except the CC2540/41, the Sleep Timer compare event can also be used as a DMA trigger (DMA trigger 11 in Table 8-1).

Note that if supply voltage drops below 2 V while in PM2, the sleep interval might be affected.

#### **11.3 Timer Capture**

The timer capture occurs when the interrupt flag for a selected I/O pin is set and this event has been dectected by the 32-kHz clock. Sleep Timer capture is enabled by setting STCC.PORT[1:0] and STCC.PIN[2:0] to the I/O pin that is to be used to trigger the capture. When STCS.VALID goes high, the capture value in STCV2:STCV1:STCV0 can be read. The captured value is one more than the value at the instant for the event on the I/O pin. Software should therefore subtract one from the captured value if abolute timing is required. To enable a new capture, follow these steps:

- 1. Clear STCS.VALID.
- 2. Wait until SLEEPSTA.CLK32K is low.
- 3. Wait until SLEEPSTA.CLK32K is high.
- 4. Clear the pin interrupt flag in the POIFG/P1IFG/P2IFG register.

This sequence, using the rising edge on P0.0 as an example, is shown in Figure 11-1. Failure to follow the procedure may cause the capture functionality to stop working until a chip reset.



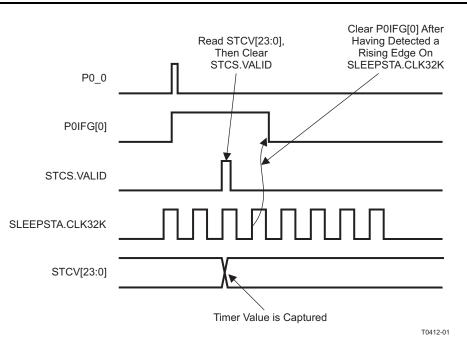


Figure 11-1. Sleep Timer Capture (Example Using Rising Edge on P0_0)

It is not possible to switch the input-capture pin while capture is enabled. Capture must be disabled before a new input-capture pin can be selected. To disable capture, follow these steps (note that interrupts will be disabled for up to half of a 32-kHz cycle, or 15.26 µs):

- 1. Disable interrupts.
- 2. Wait until SLEEPSTA.CLK32K is high.
- 3. Set STCC.PORT[1:0] to 3. This disables capture.

# 11.4 Sleep Timer Registers

The registers used by the Sleep Timer are:

- ST2 Sleep Timer 2
- ST1 Sleep Timer 1
- ST0 Sleep Timer 0
- STLOAD Sleep Timer load status
- STCC Sleep Timer capture control
- STCS Sleep Timer capture status
- STCV0 Sleep Timer capture value byte 0
- STCV1 Sleep Timer capture value byte 1
- STCV2 Sleep Timer capture value byte 2

Bit	Name	Reset	R/W	Description
7:0	ST2[7:0]	0x00	R/W	Sleep Timer count/compare value. When read, this register returns the high bits [23:16] of the Sleep Timer count. When writing, this register sets the high bits [23:16] of the compare value. The value read is latched at the time of reading register ST0. The value written is latched when ST0 is written.



#### Sleep Timer Registers

ST1 (	ST1 (0x96) – Sleep Timer 1							
Bit	Name	Reset	R/W	Description				
7:0	ST1[7:0]	0x00	R/W	Sleep Timer count/compare value. When read, this register returns the middle bits [15:8] of the Sleep Timer count. When writing, this register sets the middle bits [15:8] of the compare value. The value read is latched at the time of reading register ST0. The value written is latched when ST0 is written.				

#### ST0 (0x95) - Sleep Timer 0

Bit	Name	Reset	R/W	Description
7:0	ST0[7:0]	0x00	R/W	Sleep Timer count/compare value. When read, this register returns the low bits [7:0] of the Sleep Timer count. When writing, this register sets the low bits [7:0] of the compare value. Writes to this register are ignored unless STLOAD.LDRDY is 1.

#### STLOAD (0xAD) - Sleep Timer Load Status

Bit	Name	Reset	R/W	Description
7:1	-	0000 00 0	R0	Reserved
0	LDRDY	1	R	Load ready. This bit is 0 while the Sleep Timer loads the 24-bit compare value and 1 when the Sleep Timer is ready to start loading a new compare value.

#### STCC (0x62B0) – Sleep Timer Capture Control

Bit	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved
4:3	PORT[1:0]	11	R	Port select. Valid settings are 0–2. Capture is disabled when set to 3, i.e. an invalid setting is selected.
2:0	PIN[2:0]	111		Pin select. Valid settings are 0–7 when PORT[1:0] is 0 or 1, 0–5 when PORT[1:0] is 2. Capture is disabled when an invalid setting is selected.

#### STCS (0x62B1) - Sleep Timer Capture Status

Bit	Name	Reset	R/W	Description
7:1	-	0000 00 0	R0	Reserved
0	VALID	0	R/W0	Capture valid flag. Set to 1 when capture value in STCV has been updated. Clear explicitly to allow new capture.

# STCV0 (0x62B2) – Sleep Timer Capture Value Byte 0

Bit	Name	Reset	R/W	Description
7:0	STCV[7:0]	0x00	R	Bits [7:0] of Sleep Timer capture value

#### STCV1 (0x62B3) – Sleep Timer Capture Value Byte 1

Bit	Name	Reset	R/W	Description
7:0	STCV[15:	0x00	R	Bits [15:8] of Sleep Timer capture value
	8]			

#### STCV2 (0x62B4) – Sleep Timer Capture Value Byte 2

Bit	Name	Reset	R/W	Description
7:0	STCV[23: 16]	0x00	R	Bits [23:16] of Sleep Timer capture value



Page

The ADC (in the CC2530/CC2531/CC2540/CC2541) supports 14-bit analog-to-digital conversion with up to 12 effective number of bits (ENOB). It includes an analog multiplexer with up to eight individually configurable channels and a reference voltage generator. Conversion results can be written to memory through DMA. Several modes of operation are available.

#### Topic

12.1	ADC Introduction	138
12.2	ADC Operation	138

#### **12.1 ADC Introduction**

The ADC supports up to 14-bit analog-to-digital conversion with up to 12 bits ENOB (Effective Number Of Bits). It includes an analog multiplexer with up to eight individually configurable channels and a reference voltage generator. Conversion results can be written to memory through DMA. Several modes of operation are available.

The main features of the ADC are as follows:

- Selectable decimation rates which also set the effective resolution (7 to 12 bits).
- Eight individual input channels, single-ended or differential
- · Reference voltage selectable as internal, external single-ended, external differential, or AVDD5
- Interrupt request generation
- DMA triggers at end of conversions
- Temperature sensor input
- · Battery measurement capability

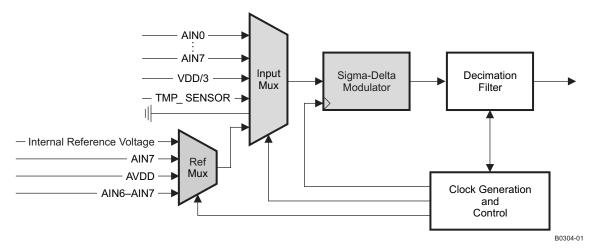


Figure 12-1. ADC Block Diagram

# 12.2 ADC Operation

This section describes the general setup and operation of the ADC and describes the use of the ADC control and status registers accessed by the CPU.

# 12.2.1 ADC Inputs

The signals on the port-0 pins can be used as ADC inputs. In the following, these port pins are referred to as the AIN0–AIN7 pins. The input pins AIN0–AIN7 are connected to the ADC.

It is possible to configure the inputs as single-ended or differential inputs. In the case where differential inputs are selected, the differential inputs consist of the input pairs AIN0–AIN1, AIN2–AIN3, AIN4–AIN5, and AIN6–AIN7. Note that no negative supply can be applied to these pins, nor a supply higher than VDD (unregulated power). It is the difference between the pins of each pair that is converted in differential mode.

In addition to the input pins AINO–AIN7, the output of an on-chip temperature sensor can be selected as an input to the ADC for temperature measurements. In order to do so, the registers TR0.ADCTM and ATEST.ATESTCTRL must be set as described in the register descriptions in Section 12.2.10 and Section 23.15.3 (CC253x) or Section 24.1 (CC2540), or Section 25.12 (CC2541).

It is also possible to select a voltage corresponding to AVDD5/3 as an ADC input. This input allows the implementation of, e.g., a battery monitor in applications where this feature is required. Note that the reference in this case must not be dependent on the battery voltage; for instance, the AVDD5 voltage must not be used as a reference.

The single-ended inputs AIN0 through AIN7 are represented by channel numbers 0 to 7. Channel numbers 8 through 11 represent the differential inputs consisting of AIN0–AIN1, AIN2–AIN3, AIN4–AIN5, and AIN6–AIN7. Channel numbers 12 through 15 represent GND (12), temperature sensor (14), and AVDD5/3 (15), with channel 13 being reserved. These values are used in the ADCCON2.SCH and ADCCON3.SCH fields.

The ADC input is a switched capacitance stage which draws current during the conversion. As an example, the equivalent input impedance of a typical device was found to be 176 k $\Omega$  when used with an input voltage of 3 V, a 512× decimation rate, and the internal reference.

# 12.2.2 ADC Conversion Sequences

The ADC can perform a sequence of conversions and move the results to memory (through DMA) without any interaction from the CPU.

The conversion sequence can be influenced with the APCFG register (see Section 7.6.6), in that the eight analog inputs to the ADC come from I/O pins that are not necessarily programmed to be analog inputs. If a channel should normally be part of a sequence, but the corresponding analog input is disabled in the APCFG register, then that channel is skipped. When using differential inputs, both pins in a differential pair must set as analog input pins in the APCFG register.

The ADCCON2.SCH register bits are used to define an ADC conversion sequence from the ADC inputs. If ADCCON2.SCH is set to a value less than 8, the conversion sequence contains a conversion from each channel from 0 up to and including the channel number programmed in ADCCON2.SCH. When ADCCON2.SCH is set to a value between 8 and 12, the sequence consists of differential inputs, starting at channel 8 and ending at the programmed channel. For ADCCON2.SCH greater than or equal to 12, the sequence consists of the selected channel only.

# 12.2.3 Single ADC Conversion

In addition to this sequence of conversions, the ADC can be programmed to perform a single conversion from any channel. Such a conversion is triggered by writing to the ADCCON3 register. The conversion starts immediately unless a conversion sequence is already ongoing, in which case the single conversion is performed as soon as that sequence is finished.

# 12.2.4 ADC Operating Modes

This section describes the operating modes and initialization of conversions.

The ADC has three control registers: ADCCON1, ADCCON2, and ADCCON3. These registers are used to configure the ADC and to report status.

The ADCCON1.EOC bit is a status bit that is set high when a conversion ends and cleared when ADCH is read.

The ADCCON1.ST bit is used to start a sequence of conversions. A sequence starts when this bit is set high, ADCCON1.STSEL is 11, and no conversion is currently running. When the sequence is completed, this bit is automatically cleared.

The ADCCON1.STSEL bits select the event that starts a new sequence of conversions. The options which can be selected are rising edge on external pin P2.0, end of previous sequence, a Timer 1 channel 0 compare event, or ADCCON1.ST is 1.

The ADCCON2 register controls how the sequence of conversions is performed.

ADCCON2.SREF is used to select the reference voltage. The reference voltage should only be changed when no conversion is running.

The ADCCON2.SDIV bits select the decimation rate, thereby also the resolution and time required to complete a conversion, and hence the sample rate. The decimation rate should only be changed when no conversion is running.

The last channel of a sequence is selected with the ADCCON2.SCH bits as described previously.



#### ADC Operation

The ADCCON3 register controls the channel number, reference voltage, and decimation rate for a single conversion. The single conversion takes place immediately after the ADCCON3 register is written to, or if a conversion sequence is ongoing, immediately after the sequence has ended. The coding of the register bits is exactly as for ADCCON2.

# 12.2.5 ADC Conversion Results

The digital conversion result is represented in 2s-complement form. For single-ended configurations, the result can be expected to be positive. This is because the result is the difference between the input signal and ground, which is always positively signed (Vconv = Vinp – Vinn, where Vinn = 0 V). The maximum value is reached when the input signal is equal to VREF, the selected voltage reference. For differential configurations, the difference between two pins is converted, and this difference can be negatively signed. For example, with a decimation rate of 512 using only the 12 MSBs of the digital conversion result register, the maximum value of 2047 is reached when the analog input (Vconv) is equal to VREF, and minimum value of -2048 is reached when the analog input is equal to -VREF.

The digital conversion result is available in ADCH and ADCL when ADCCON1.EOC is set to 1. Note that the conversion result always resides in the MSB section of the combined ADCH and ADCL registers.

When the ADCCON2.SCH bits are read, they indicate the channel on which conversion is ongoing. The results in ADCL and ADCH normally apply to the previous conversion. If the conversion sequence has ended, ADCCON2.SCH has a value of one more than the last channel number, but if the channel number last written to ADCCON2.SCH was 12 or more, the same value is read back.

# 12.2.6 ADC Reference Voltage

The positive reference voltage for analog-to-digital conversions is selectable as either an internally generated voltage, the AVDD5 pin, an external voltage applied to the AIN7 input pin, or a differential voltage applied to the AIN6–AIN7 inputs.

The accuracy of the conversion results depend on the stability and noise properties of the reference voltage. Offset from the wanted voltage introduces a gain error in the ADC proportional to the ratio of the wanted voltage and the actual voltage. Noise on the reference must be lower than quantization noise of the ADC to ensure the specified SNR is achieved.

# 12.2.7 ADC Conversion Timing

The ADC should only be used with the 32-MHz XOSC, and no system clock division should be implemented by the user. The actual ADC sampling frequency of 4 MHz is generated by fixed internal division. The time required to perform a conversion depends on the selected decimation rate. In general, the conversion time is given by:

Tconv = (decimation rate + 16) × 0.25  $\mu$ s.

# 12.2.8 ADC Interrupts

The ADC generates an interrupt when a single conversion triggered by writing to ADCCON3 has completed. No interrupt is generated when a conversion from the sequence is completed.

# 12.2.9 ADC DMA Triggers

The ADC generates a DMA trigger every time a conversion from the sequence has completed. When a single conversion completes, no DMA trigger is generated.

There is one DMA trigger for each of the eight channels defined by the first eight possible settings for ADCCON2.SCH. The DMA trigger is active when a new sample is ready from the conversion for the channel. The DMA triggers are named ADC_CH**sd** in Table 8-1, where **s** is single-ended channel and **d** is differential channel.

In addition, one DMA trigger, ADC_CHALL, is active when new data is ready from any of the channels in the ADC conversion sequence.



# 12.2.10 ADC Registers

This section describes the ADC registers.

#### ADCL (0xBA) - ADC Data, Low

Bit	Name	Reset	R/W	Description
7:2	ADC[5:0]	0000 00	R	Least-significant part of ADC conversion result
1:0	-	00	R0	Reserved. Always read as 0

# ADCH (0xBB) - ADC Data, High

			-	
Bit	Name	Reset	R/W	Description
7:0	ADC[13:6]	0x00	R	Most-significant part of ADC conversion result

#### ADCCON1 (0xB4) – ADC Control 1

Bit	Name	Reset	R/W	Description
7	EOC	0	R/H0	End of conversion. Cleared when ADCH has been read. If a new conversion is completed before the previous data has been read, the EOC bit remains high.
				0: Conversion not complete
				1: Conversion completed
6	ST	0	R/W1/ H0	Start conversion. Read as 1 until conversion has completed
				0: No conversion in progress
				1: Start a conversion sequence if ADCCON1.STSEL = 11 and no sequence is running.
5:4	STSEL[1:0]	11	R/W	Start select. Selects the event that starts a new conversion sequence
				00: External trigger on P2.0 pin
				01: Full speed. Do not wait for triggers
				10: Timer 1 channel 0 compare event
				11: ADCCON1.ST = 1
3:2	-	00	R/W	Controls the 16-bit random-number generator. See ADCCON1 (0xB4) – ADC Control 1 description in Section 14.3.
1:0	-	11	R/W	Reserved. Always set to 11



ADC Operation

Bit	Name	Reset	R/W	Description
7:6	SREF[1:0]	00	R/W	Selects reference voltage used for the sequence of conversions
				00: Internal reference
				01: External reference on AIN7 pin
				10: AVDD5 pin
				11: External reference on AIN6–AIN7 differential input
5:4	SDIV[1:0]	01	R/W	Sets the decimation rate for channels included in the sequence of conversions. The decimation rate also determines the resolution and time required to complete a conversion.
				00: 64 decimation rate (7 bits ENOB setting)
				01: 128 decimation rate (9 bits ENOB setting)
				10: 256 decimation rate (10 bits ENOB setting)
				11: 512 decimation rate (12 bits ENOB setting)
3:0	SCH[3:0]	0000	R/W	Sequence channel select. Selects the end of the sequence. A sequence can either be from AIN0 to AIN7 (SCH $\leq$ 7) or from differential input AIN0–AIN1 to AIN6–AIN7 (8 $\leq$ SCH $\leq$ 11). For other settings, only one conversions is performed.
				When read, these bits indicate the channel number on which a conversion is ongoing.
				0000: AIN0
				0001: AIN1
				0010: AIN2
				0011: AIN3
				0100: AIN4
				0101: AIN5
				0110: AIN6
				0111: AIN7
				1000: AIN0–AIN1
				1001: AIN2–AIN3
				1010: AIN4–AIN5
				1011: AIN6–AIN7
				1100: GND
				1101: Reserved
				1110: Temperature sensor
				1111: VDD/3



Bit	Name	Reset	R/W	Description
7:6	EREF[1:0]	00	R/W	Selects reference voltage used for the extra conversion
				00: Internal reference
				01: External reference on AIN7 pin
				10: AVDD5 pin
				11: External reference on AIN6–AIN7 differential input
5:4	EDIV[1:0]	00	R/W	Sets the decimation rate used for the extra conversion. The decimation rate also determines the resolution and the time required to complete the conversion.
				00: 64 decimation rate (7 bits ENOB)
				01: 128 decimation rate (9 bits ENOB)
				10: 256 decimation rate (10 bits ENOB)
				11: 512 decimation rate (12 bits ENOB)
3:0	ECH[3:0]	0000	R/W	Single channel select. Selects the channel number of the single conversion that is triggered by writing to ADCCON3.
				0000: AIN0
				0001: AIN1
				0010: AIN2
				0011: AIN3
				0100: AIN4
				0101: AIN5
				0110: AIN6
				0111: AIN7
				1000: AIN0–AIN1
				1001: AIN2–AIN3
				1010: AIN4–AIN5
				1011: AIN6–AIN7
				1100: GND
				1101: Reserved
				1110: Temperature sensor
				1111: VDD/3

# TR0 (0x624B) – Test Register 0

Bit	Name	Reset	R/W	Description
7:1	-	000 000	R0	Reserved. Write as 0.
0	ADCTM	0	R/W	Set to 1 to connect the temperature sensor to the SOC_ADC. See also ATEST register description to enable the temperature sensor in Section 23.15.3 (CC253x) or Section 24.1 (CC2540) or Chapter 25 (CC2541).



# **Battery Monitor**

The battery monitor (in the CC2533 only) enables simple voltage monitoring in the devices that do not include an ADC. It is designed such that it is accurate in the voltage areas around 2 V, with lower resolution at higher voltages. The registers BATTMON and MONMUX are used to access and control the functionality of the battery monitor.

The battery monitor can also be used to do simple temperature monitoring by connecting it to the chip internal temperature sensor instead of the supply voltage. The input is controlled using the MONMUX register.

Topic

13.1	Functionality and Usage of the Battery Monitor	146
13.2	Using the Battery Monitor for Temperature Monitoring	146
13.3	Battery Monitor Registers	147



Functionality and Usage of the Battery Monitor

#### 13.1 Functionality and Usage of the Battery Monitor

The battery monitor makes it possible to check whether the supply voltage (AVDD5) is above or below a certain programmable level. Its usage is controlled by the BATTMON register in the following manner:

BATTMON_VOLTAGE is used to set the trigger point for the battery monitor. Note the fact that the step size is different for different voltage ranges (see the register description in Section 13.3 for details). This is done to achieve good accuracy in the voltage areas around 2 V, with lower resolution at higher voltages.

BATTMON_PD is used to enable/disable the battery monitor.

After enabling the battery monitor by setting  $BATTMON_PD = 0$  and waiting for at least 2 µs, the value of  $BATTMON_OUT$  indicates whether the voltage is above or below the trigger point (set by  $BATTMON_VOLTAGE$ ).

**NOTE:** One should turn the battery monitor off (BATTMON_PD = 1) after reading the measurement BATTMON_OUT in order to save power, as the battery monitor consumes power when enabled ( = 0).

Recommended usage of the battery monitor can be summarized in the following way:

- 1. Set **BATTMON_VOLTAGE** to the value to be monitored.
- 2. Enable the battery monitor by setting BATTMON_PD = 0.
- 3. Wait for at least 2 µs.
- 4. Read the BATTMON_OUT result to see whether the voltage level is above or below the value set in BATTMON_VOLTAGE.
- 5. Disable the battery monitor (BATTMON_PD = 1) to avoid unnecessary current consumption.

#### 13.2 Using the Battery Monitor for Temperature Monitoring

The battery monitor can also be used to do some simple temperature monitoring. When the battery monitor is connected to the internal temperature sensor instead of the supply voltage AVDD5 (see the description of MONMUX in Section 13.3), it can indicate whether the temperature is above or below a certain level. This is done by comparing the voltage coming from the temperature sensor to the voltage trigger point of the battery monitor. The controls for this measurement are the same as for the normal use of the battery monitor (see the description of BATTMON in Section 13.3).

It is important to understand that due to the nature of the battery monitor (optimized for voltages arround 2 V) and the output voltage range of the temperature sensor, there are only about 8 temperature trigger values in the temperature range of -40°C to 125°C (see Table 13-1). As a result, the battery monitor gives only a rough indication of the temperature range, but this is useful for doing temperature compensation on analog components in a system. See the device's data sheet (Appendix C) for performance characteristics details.

Temperature	BATTMON_VOLTAGE			
-40°C	22			
-26°C	21			
-11°C	20			
7°C	19			
25°C	18			
47°C	17			
70°C	16			
97°C	15			
128°C	14			

#### Table 13-1. Values Showing How Different Temperatures Relate to BATTMON_VOLTAGE for a Typical Device



Battery Monitor Registers

The temperature sensor is inversely proportional to BATTMON_VOLTAGE. The temperature (in °C) corresponding to a given BATTMON_VOLTAGE is given by:

$$Temp = \frac{A}{BATTMON_VOLTAGE<4:0>} - B$$

(1)

Assuming BATTMON_VOLTAGE < 27, and only valid for  $-40^{\circ}$ C < Temp < 125°C, A and B for a typical device are given in Table 13-2.

#### Table 13-2. Values for A and B (for a Typical Device) When Using the Battery monitor for Temperature Monitoring

Constant	Тур
A	6470
В	334

Note that A should be relatively constant for all devices, but B is not. Information that can be used to calculate B for a given chip is included in the chip's information page (see Section 2.2.3 for info about the information page).

Example:

Find the BATTMON_VOLTAGE setting that tells whether the temperature is above or below 75°C.

BATTMON_VOLTAGE<4:0> = 
$$\frac{6470}{75+334}$$
 = 15.82

0470

(2)

The closest setting is 16, which corresponds to ~70°C (see Table 13-1). By writing 16 to BATTMON_VOLTAGE, an output of BATTMON_OUT = 1 tells that the temperature is above 70°C, whereas BATTMON_OUT = 0 tells that it is below 70°C.

#### 13.3 Battery Monitor Registers

This section describes the battery monitor registers.



#### Battery Monitor Registers

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	TMON (0x6264) – Bat			
Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Always read 0
6	BATTMON_OUT	0	R	Result from the battery monitor.
				1: Voltage is above value set in BATTMON_VOLTAGE.
				0: Voltage is below the value set in BATTMON_VOLTAGE.
				Note that the value of BATTMON_OUT is undefined except when BATTMON_PD is 0 and has
			_	been 0 for 2 us.
5:1	BATTMON_VOLTAGE	11100	R/W	Controls the trigger point for the battery monitor. The step size is 24 mV for the first 23 settings, and then 169 mV (unless temperature-sense mode is enabled; see Section 13.2 for details). Range to be used: 3–31
				3: 1.93 V
				4: 1.93 V + (4 − 3) × 0.024 V = 1.954 V
				5: 1.93 V + (5 − 3) × 0.024 V = 1.978 V
				6: 1.93 V + (6 − 3) × 0.024 V = 2.002 V
				7: 1.93 V + (7 − 3) × 0.024 V = 2.026 V
				8: 1.93 V + (8 − 3) × 0.024 V = 2.050 V
				9: 1.93 V + (9 − 3) × 0.024 V = 2.074 V
				10: 1.93 V + (10 − 3) × 0.024 V = 2.098 V
				11: 1.93 V + (11 – 3) × 0.024 V = 2.122 V
				12: 1.93 V + (12 – 3) × 0.024 V = 2.146 V
				13: 1.93 V + (13 – 3) × 0.024 V = 2.170 V
				14: 1.93 V + (14 – 3) × 0.024 V = 2.194 V
				15: 1.93 V + (15 – 3) × 0.024 V = 2.218 V
				16: 1.93 V + (16 − 3) × 0.024 V = 2.242 V
				17: 1.93 V + (17 – 3) × 0.024 V = 2.266 V
				18: 1.93 V + (18 – 3) × 0.024 V = 2.290 V
				19: 1.93 V + (19 – 3) × 0.024 V = 2.314 V
				20: 1.93 V + (20 – 3) × 0.024 V = 2.338 V
				21: 1.93 V + (21 – 3) × 0.024 V = 2.362 V
				22: 1.93 V + (22 – 3) × 0.024 V = 2.386 V
				23: 1.93 V + (23 – 3) × 0.024 V = 2.410 V
				24: 1.93 V + (24 − 3) × 0.024 V = 2.434 V
				25: 1.93 V + (25 – 3) × 0.024 V = 2.458 V
				26: 1.93 V + (26 - 3) × 0.024 V = 2.482 V
				27: 2.482 V + (27 – 26) × 0.169 V = 2.651 V
				28: 2.482 V + (28 – 26) × 0.169 V = 2.820 V
				29: 2.482 V + (29 – 26) × 0.169 V = 2.989 V
				30: 2.482 V + (30 – 26) × 0.169 V = 3.158 V
				31 : 2.48 2V + (31 – 26) × 0.169 V = 3.327 V
0	BATTMON_PD	1	R/W	Turns on the battery monitor. Wait at least 2 µs before reading BATTMON_OUT.
	_			0: Enable the battery monitor.
				1: Disable the battery monitor.
				One should turn the battery monitor off (BATTMON_PD = 1) after reading out the measurement BATTMON_OUT in order to save power, as the battery monitor consumes power when enabled

#### MONMUX (0x61A6) – Monitor MUX

Bit	Name	Reset	R/W	Description			
7:1	-	-	-	Reserved			
0	BATTMON_INPUT	0	R/W	etermines the input to the battery moitor:			
				: Supply voltage (AVDD5)			
				1: Voltage from the temperature sensor, which needs to be enabled using the ATEST.ATESTCTRL register; described in Section 23.15.3			



# **Random-Number Generator**

This chapter provides more information about the random-number generator and its usage.

#### Topic

opic		Page
14.1	Introduction	150
14.2	Random-Number-Generator Operation	150
14.3	Random-Number-Generator Registers	151



#### 14.1 Introduction

The random-number generator has the following features.

- Generates pseudorandom bytes which can be read by the CPU or used directly by the command strobe processor (see Section 23.14).
- Calculates CRC16 of bytes that are written to RNDH.
- Seeded by value written to RNDL.

The random-number generator is a 16-bit linear-feedback shift register (LFSR) with polynomial  $X^{16} + X^{15} + X^2 + 1$  (i.e., CRC16). It uses different levels of unrolling depending on the operation it performs. The basic version (no unrolling) is shown in Figure 14-1.

The random-number generator is turned off when ADCCON1.RCTRL = 11.

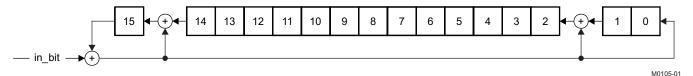


Figure 14-1. Basic Structure of the Random-Number Generator

#### 14.2 Random-Number-Generator Operation

The operation of the random-number generator is controlled by the ADCCON1.RCTRL bits (see also Section 12.2.10). The current value of the 16-bit shift register in the LFSR can be read from the RNDH and RNDL registers.

#### 14.2.1 Pseudorandom Sequence Generation

The default operation (ADCCON1.RCTRL is 00) is to clock the LFSR once (13× unrolling; where clocking with 13× unrolling means performing an operation equivalent to doing 13 shifts with feedback) each time the command strobe processor (Section 23.14) reads the random value. This leads to the availability of a fresh pseudorandom byte from the LSB end of the LFSR.

Another way to update the LFSR is to set ADCCON1.RCTRL to 01. This clocks the LFSR once (13× unrolling), and the ADCCON1.RCTRL bits are automatically cleared when the operation has completed.

#### 14.2.2 Seeding

The LFSR can be seeded by writing to the RNDL register twice. Each time the RNDL register is written, the 8 LSBs of the LFSR are copied to the 8 MSBs and the 8 LSBs are replaced with the new data byte that was written to RNDL.

For the CC253x, when a random value is required, the LFSR should be seeded by writing RNDL with random bits from the IF_ADC in the RF receive path. To use this seeding method, the radio must first be powered on. The radio should be placed in the infinite RX state to avoid possible sync detect in the RX state. The random bits from the IF_ADC are read from the least significant bit position of the RF register RFRND. These bits should be concatenated over time to form the bytes needed for the random-number-generator seed. See Section 23.12 for a description of the randomness of these numbers. Note that this cannot be done while the radio is in use for normal tasks.

Note that a seed value of 0x0000 or 0x8003 always leads to an unchanged value in the LFSR after clocking, as no values are pushed in via in_bit (see Figure 14-1); hence, neither of these seed values should not be used for random-number generation.

#### 14.2.3 CRC16

The LFSR can also be used to calculate the CRC value of a sequence of bytes. Writing to the RNDH register triggers a CRC calculation. The new byte is processed from the MSB end and an 8× unrolling is used, so that a new byte can be written to RNDH every clock cycle.



Note that the LFSR must be properly seeded by writing to RNDL before the CRC calculations start. Usually, the seed value for CRC calculations should be 0x0000 or 0xFFFF.

#### 14.3 Random-Number-Generator Registers

This section describes the random-number-generator registers.

Bit	Name	Reset	R/W	Description	
7:0	RNDL[7:0]	0xFF	R/W	Random value/seed or CRC result, low byte	
				When used for random-number generation, writing to this register twice seeds the random-number generator. Writing to this register copies the 8 LSBs of the LFSR to the 8 MSBs and replaces the 8 LSBs with the data value written.	
				he value returned when reading from this register is the 8 LSBs of the LFSR.	
				When used for random-number generation, reading this register returns the 8 LSBs of the random number. When used for CRC calculations, reading this register returns the 8 LSBs of the CRC result.	

#### RNDH (0xBD) – Random-Number-Generator Data, High Byte

Bit	Name	Reset	R/W	Description
7:0	RNDH[7:0]	0xFF	R/W	Random value or CRC result/input data, high byte
				When written, a CRC16 calculation is triggered, and the data value written is processed starting with the MSB.
				The value returned when reading from this register is the 8 MSBs of the LFSR.
				When used for random-number generation, reading this register returns the 8 MSBs of the random number. When used for CRC calculations, reading this register returns the 8 MSBs of the CRC result.

#### ADCCON1 (0xB4) - ADC Control 1 (see also Section 12.2.10)

Bit	Name	Reset	R/W	Description				
7:4	-	0011	-	For CC2533, these bits are reserved. For the other devices, see the <b>ADCCON1 (0xB4)</b> – <b>ADC Control 1</b> description in Section 12.2.10.				
3:2	RCTRL[1:0]	00	R/W	Controls the 16-bit random-number generator (Chapter 14). When 01 is written, the setting automatically returns to 00 when the operation has completed.				
				00: Normal operation. (13× unrolling)				
				01: Clock the LFSR once (13× unrolling)				
				10: Reserved				
				11: Stopped. Random-number generator is turned off.				
1:0	-	11	R/W	Reserved. Always set to 11				



## **AES Coprocessor**

Page

The Advanced Encryption Standard (AES) coprocessor allows encryption/decryption to be performed with minimal CPU usage.

The coprocessor has the following features:

- Supports all security suites in IEEE 802.15.4
- ECB, CBC, CFB, OFB, CTR, and CBC-MAC modes
- Hardware support for CCM mode
- 128-bit key and IV/nonce
- DMA transfer trigger capability

#### Topic

# 15.1 AES Operation 154 15.2 Key and IV 154 15.3 Padding of Input Data 154 15.4 Interface to CPU 154 15.5 Modes of Operation 154 15.6 CBC-MAC 154 15.7 CCM Mode 155 15.8 AES Interrupts 157 15.9 AES DMA Triggers 157 15.10 AES Registers 157

#### 15.1 AES Operation

To encrypt a message, the following procedure must be followed (ECB, CBC):

- Load key
- Load initialization vector (IV)
- Download and upload data for encryption/decryption.

The AES coprocessor works on blocks of 128 bits. A block of data is loaded into the coprocessor, encryption is performed, and the result must be read out before the next block can be processed. Before each block is loaded, a dedicated start command must be sent to the coprocessor.

#### 15.2 Key and IV

Before a key or IV/nonce load starts, an appropriate load key or IV/nonce command must be issued to the coprocessor. When loading the IV, it is important also to set the correct mode.

A key load or IV load operation aborts any processing that could be running. The key, once loaded, stays valid until a key reload takes place.

The IV must be downloaded before the beginning of each message (not each block).

Both the key and IV values are cleared by a reset of the device and when PM2 or PM3 is entered.

#### 15.3 Padding of Input Data

The AES coprocessor works on blocks of 128 bits. If the last block contains less than 128 bits, it must be padded with zeros when written to the coprocessor.

#### 15.4 Interface to CPU

The CPU communicates with the coprocessor using three SFR registers:

- ENCCS, encryption control and status register
- ENCDI, encryption input register
- ENCDO, encryption output register

Read/write to the status register is done directly by the CPU, whereas access to the input/output registers should be performed using direct memory access (DMA).

When using DMA with the AES coprosessor, two DMA channels must be used, one for input data and one for output data. The DMA channels must be initialized before a start command is written to ENCCS. Writing a start command generates a DMA trigger, and the transfer is started. After each block is processed, an interrupt is generated. The interrupt is used to issue a new start command to ENCCS.

#### 15.5 Modes of Operation

When using CFB, OFB, or CTR mode, the 128-bit blocks are divided into four 32-bit blocks. The 32 bits are loaded into the AES coprocessor, and the resulting 32 bits are read out. This continues until all 128 bits have been encrypted. The only time one must consider this is if data is loaded/read directly using the CPU. When using DMA, this is handled automatically by the DMA triggers generated by the AES coprocessor; thus, DMA is preferred.

Both encryption and decryption are performed similarly.

The CBC-MAC mode is a variant of the CBC mode. See Section 15.6 for an explanation.

CCM is a combination of CBC-MAC and CTR. Parts of the CCM must therefore be done in software. The following section gives a short explanation of the necessary steps to be done.

#### 15.6 CBC-MAC

When performing CBC-MAC encryption, data is downloaded to the coprocessor in CBC-MAC mode one block at a time, except for the last block. Before the last block is loaded, the mode is changed to CBC. The last block is downloaded and the block uploaded is the message MAC.



CBC-MAC decryption is similar to encryption. The message MAC uploaded must be compared with the MAC to be verified.

#### 15.7 CCM Mode

To encrypt a message in CCM mode, the following sequence can be conducted (key is already loaded):

#### Message Authentication Phase

This phase takes place during the following steps 1–6.

- 1. The software loads the IV with zeros.
- 2. The software creates block B0. The layout of block B0 is shown in Figure 15-1.

	Name B0	ne					Designation First Block for Authentication in CCM Mode									
Byte	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
Name	Flag		Nonce										L	М		

#### Figure 15-1. Message Authentication Phase Block B0

There is no restriction on the nonce value. L_M is the message length in bytes.

For 802.15.4, nonce is 13 bytes and L_M is 2 bytes.

The content of the authentication flag byte is described in Figure 15-2.

L is set to 6 in this example. So, L - 1 is set to 5. M and A_Data can be set to any value.

	Name FLAG/B0			Designation Authentication Flag Field for CCM mode						
Bit	7	6	5	4	3	2	1	0		
Name	Reserved	A_Data		(M – 2)/2		L – 1				
Value	0	х	х	x	х	1	0	1		

#### Figure 15-2. Authentication Flag Byte

- If some additional authentication data (denoted a, following) is needed (that is, A_Data = 1), the software creates the A_Data length field, called L(a) by:
  - (a) If I(a) = 0, (that is, A_Data = 0), then L(a) is the empty string. Note that I(a) is the length of a in octets.
  - (b) If  $0 < I(a) < 2^{16} 2^8$ , then L(a) is the 2-octet encoding of I(a).

The additional authentication data is appended to the A_Data length field L(a). The additional authentication blocks are padded with zeros until the last additional authentication block is full. There is no restriction on the length of a.

AUTH-DATA = L(a) + Authentication Data + (zero padding)

- 4. The last block of the message is padded with zeros until full (that is, if its length is not an integral multiple of 128 bits).
- 5. The software concatenates block B0, the additional authentication blocks if any, and the message;

Input message = B0 + AUTH-DATA + Message + (zero padding of message)

 Once the input message authentication by CBC-MAC is finished, the software leaves the uploaded buffer contents unchanged (M = 16), or keeps only the higher-M bytes of the buffer unchanged, while setting the lower bits to 0 (M != 16).

The result is called T.

#### Message Encryption

7. The software creates the key stream block A0. Note that L = 6, with the current example of the CTR generation. The content is shown in Figure 15-3.

Note that when encrypting authentication data T to generate U in OFB mode, the CTR value must be zero. When encrypting message blocks using CTR mode, the CTR value must be any value but zero. The content of the encryption-flag byte is described in Figure 15-4.

SWRU191C–April 2009–Revised January 2012 Submit Documentation Feedback



CCM Mode

	Name A0						Designation First CTR Value for CCM Mode									
Byte	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
Name	Flag		Nonce CTR													

#### Figure 15-3. Message Encryption Phase Block

	Name FLAG/A0		Designation Encryption Flag Field for CCM Mode					
Bit	7	6	5	4	3	2	1	0
Name	Rese	erved		—			L – 1	
Value	0	0	0	0	0	1	0	1

#### Figure 15-4. Encryption Flag Byte

- 8. The software loads A0 by selecting a Load IV/nonce command. To do so, it sets the mode to CFB or OFB at the same time it selects the Load IV/nonce command.
- The software calls a CFB or an OFB encryption on the authenticated data T. The uploaded buffer contents stay unchanged (M = 16), or only its first M bytes stay unchanged, the others being set to 0 (M 16). The result is U, which is used later.
- 10. The software calls a CTR-mode encryption immediately on the still-padded message blocks. It must reload the IV when the CTR value is any value but zero.
- 11. The encrypted authentication data U is appended to the encrypted message. This gives the final result, c.

Result C = encrypted message(m) + U



#### Message Decryption

#### CCM Mode Decryption

In the coprocessor, the automatic generation of CTR works on 32 bits; therefore, the maximum length of a message is  $128 \times 2^{32}$  bits, that is  $2^{36}$  bytes, which can be written in a 6-bit word. So, the value L is set to 6. To decrypt a CCM-mode processed message, the following sequence can be conducted (key is already loaded).

#### Message Parsing Phase

- 1. The software parses the message by separating the M rightmost octets, namely U, and the other octets, namely string C.
- 2. C is padded with zeros until it can fill an integral number of 128-bit blocks.
- 3. U is padded with zeros until it can fill a 128-bit block.
- 4. The software creates the key stream block A0. It is done the same way as for CCM encryption.
- 5. The software loads A0 by selecting a Load IV/nonce command. To do so, it sets the mode to CFB or OFB at the same time as it selects the IV load.
- The software calls a CFB or an OFB encryption on the encrypted authenticated data U. The uploaded buffer contents stay unchanged (M = 16), or only its first M bytes stay unchanged, the others being set to 0 (M != 16). The result is T.
- 7. The software calls a CTR-mode decryption immediately on the encrypted message blocks C. Reloading the IV/CTR is not necessary.

#### **Reference Authentication Tag Generation**

This phase is identical to the authentication phase of CCM encryption. The only difference is that the result is named MACTag (instead of T).

#### Message Authentication Checking Phase

The software compares T with MACTag.

#### 15.8 AES Interrupts

The AES interrupt, ENC, is produced when encryption or decryption of a block is completed. The interrupt enable bit is IEN0.ENCIE, and the interrupt flag is SOCON.ENCIF.

#### 15.9 AES DMA Triggers

Two DMA triggers are associated with the AES coprocessor. These are ENC_DW, which is active when input data must be downloaded to the ENCDI register, and ENC_UP, which is active when output data must be uploaded from the ENCDO register.

The ENCDI and ENCDO registers should be set as destination and source locations for DMA channels used to transfer data to or from the AES coprocessor.

#### 15.10 AES Registers

The AES coprocessor registers have the layout shown in this section.

The registers return to their reset value when the chip enters PM2 or PM3.



#### AES Registers

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved, always read as 0
6:4	MODE[2:0]	000	R/W	Encryption/decryption mode
				000: CBC
				001: CFB
				010: OFB
				011: CTR
				100: ECB
				101: CBC MAC
				110: Reserved
				111: Reserved
3	RDY	1	R	Encryption/decryption ready status
				0: Encryption/decryption in progress
				1: Encryption/decryption is completed.
2:1	CMD[1:0]	0	R/W	Command to be performed when a 1 is written to ST
				00: Encrypt block
				01: Decrypt block
				10: Load key
				11: Load IV/nonce
0	ST	0	R/W1 H0	Start processing command set by CMD. Must be issued for each command or 128-bit block of data. Cleared by hardware.

#### ENCDI (0xB1) – Encryption Input Data

Bit	Name	Reset	R/W	Description	
7:0	DIN[7:0]	0x00	R/W	Encryption input data	

#### ENCDO (0xB2) - Encryption Output Data

Bit	Name	Reset	R/W	Description
7:0	DOUT[7:0]	0x00	R/W	Encryption output data



# Watchdog Timer

The Watchdog Timer (WDT) is intended as a recovery method in situations where the CPU may be subjected to a software upset. The WDT resets the system when software fails to clear the WDT within the selected time interval. The watchdog can be used in applications that are subject to electrical noise, power glitches, electrostatic discharge, etc., or where high reliability is required. If the watchdog function is not needed in an application, it is possible to configure the Watchdog Timer to be used as an interval timer that can be used to generate interrupts at selected time intervals.

The features of the Watchdog Timer are as follows:

- · Four selectable timer intervals
- Watchdog mode
- Timer mode
- Interrupt request generation in timer mode

The WDT is configured as either a Watchdog Timer or as a timer for general-purpose use. The operation of the WDT module is controlled by the WDCTL register. The Watchdog Timer consists of a 15-bit counter clocked by the 32-kHz clock source. Note that the contents of the 15-bit counter are not user-accessible. The contents of the 15-bit counter are retained during all power modes, and the Watchdog Timer continues counting when entering active mode again.

Горіс		Page
16.2	Watchdog Mode Timer Mode Watchdog Timer Register	160



#### 16.1 Watchdog Mode

The WDT is disabled after a system reset. To start the WDT in watchdog mode, the WDCTL.MODE[1:0] bits must be set to 10. The Watchdog Timer counter then starts incrementing from 0. When the timer is enabled in watchdog mode, it is not possible to disable the timer. Therefore, writing 00 or 01 to WDCTL.MODE[1:0] has no effect if the WDT is already operating in Watchdog mode.

The WDT operates with a Watchdog Timer clock frequency of 32.768 kHz (when the 32-kHz XOSC is used). This clock frequency gives time-out periods equal to 1.9 ms, 15.625 ms, 0.25 s, and 1 s, corresponding to the count value settings 64, 512, 8192, and 32,768, respectively.

If the counter reaches the selected timer interval value, the Watchdog Timer generates a reset signal for the system. If a watchdog clear sequence is performed before the counter reaches the selected timer interval value, the counter is reset to 0 and continues incrementing its value. The watchdog clear sequence consists of writing 0xA to WDCTL.CLR[3:0], followed by writing 0x5 to the same register bits within one watchdog clock period. If this complete sequence is not performed before the end of the watchdog period, the Watchdog Timer generates a reset signal for the system.

When the WDT has been enabled in watchdog mode, it is not possible to change the mode by writing to the WDCTL.MODE[1:0] bits, and the timer interval value cannot be changed.

In watchdog mode, the WDT does not produce interrupt requests.

#### 16.2 Timer Mode

To start the WDT in timer mode, the WDCTL.MODE[1:0] bits must be set to 11. The timer is started and the counter starts incrementing from 0. When the counter reaches the selected interval value, the timer produces an interrupt request (IRCON2.WDTIF/IEN2.WDTIE).

In timer mode, it is possible to clear the timer contents by writing a 1 to WDCTL.CLR[0]. When the timer is cleared, the content of the counter is set to 0. Writing 00 to WDCTL.MODE[1:0] stops the timer and clears it to 0.

The timer interval is set by the WDCTL.INT[1:0] bits. The interval cannot be changed during timer operation, and should be set when the timer is started. In timer mode, a reset is not produced when the timer interval has been reached.

Note that if the watchdog mode is selected, the timer mode cannot be selected before the chip is reset.

#### 16.3 Watchdog Timer Register

This section describes the register, WDCTL, for the Watchdog Timer.



Bit	Name	Reset	R/W	Description
7:4	CLR[3:0]	0000	R0/W	Clear timer. In watchdog mode, when 0xA followed by 0x5 is written to these bits, the timer is cleared (i.e. loaded with 0). Note that the timer is only cleared when 0x5 is written within one watchdog clock period after 0xA was written. Writing these bits when the Watchdog Timer is IDLE has no effect. When operating in timer mode, the timer can be cleared to 0x0000 (but not stopped) by writing 1 to CLR[0] (the other 3 bits are don't care).
3:2	MODE[1:0]	00	R/W	Mode select. These bits are used to start the WDT in watchdog mode or timer mode. Setting these bits to IDLE stops the timer when in timer mode. Note: to switch to watchdog mode when operating in timer mode, first stop the WDT - then start the WDT in Watchdog mode. When operating in Watchdog mode, writing these bits has no effect.
				00: IDLE
				01: Reserved
				10: Watchdog mode
				11: Timer mode
1:0	INT[1:0]	00	R/W	Timer interval select. These bits select the timer interval, which is defined as a given number of 32-kHz oscillator periods. Note that the interval can only be changed when the WDT is IDLE, so the interval must be set at the same time as the timer is started.
				00: Clock period × 32,768 (~1 s) when running the 32-kHz XOSC
				01: Clock period × 8192 (~0.25 s)
				10: Clock period × 512 (~15.625 ms)
				11: Clock period × 64 (~1.9 ms)
				For CC253x and CC2540, when clock division is enabled through CLKCONCMD.CLKSPD, the length of the watchdog timer interval is reduced by a factor equal to the current oscillator clock frequency divided by the set clock speed. E.g., if 32-MHz crystal is selected and clock speed is set to 4 MHz, then the watchdog timeout is reduced by a factor o 32 MHz/4 MHz = 8. If the watchdog interval set by WDCTL.INT was 1 s, nominally it is 1/8 s with this clock division factor. For CC2541 the watchdog timer interval is independent of the clock division rate.



USART 0 and USART 1 are serial communications interfaces that can be operated separately in either asynchronous UART mode or in synchronous SPI mode. The two USARTs have identical function, and are assigned to separate I/O pins. See Section 7.6 for I/O configuration.

Торіс		Page
	UART Mode	
	SPI Mode	
17.3	SSN Slave-Select Pin	166
17.4	Baud-Rate Generation	166
17.5	USART Flushing	167
17.6	USART Interrupts	167
17.7	USART DMA Triggers	167
17.8	USART Registers	167

#### 17.1 UART Mode

For asynchronous serial interfaces, the UART mode is provided. In the UART mode, the interface uses a two-wire or four-wire interface consisting of the pins RXD and TXD, and optionally RTS and CTS. The UART mode of operation includes the following features:

- 8 or 9 payload bits
- Odd, even, or no parity
- · Configurable start- and stop-bit levels
- Configurable LSB- or MSB-first transfer
- · Independent receive and transmit interrupts
- Independent receive and transmit DMA triggers
- Parity and framing error status

The UART mode provides full-duplex asynchronous transfers, and the synchronization of bits in the receiver does not interfere with the transmit function. A UART byte transfer consists of a start bit, eight data bits, an optional ninth data or parity bit, and one or two stop bits. Note that the data transferred is referred to as a byte, although the data can actually consist of eight or nine bits.

The UART operation is controlled by the USART control and status registers, UxCSR, and the UART control registers, UxUCR, where x is the USART number, 0 or 1.

The UART mode is selected when UxCSR.MODE is set to 1.

#### 17.1.1 UART Transmit

A UART transmission is initiated when the USART receive/transmit data buffers, UxDBUF, are written. The byte is transmitted on the TXDx output pins. The UxDBUF registers are double-buffered.

The UxCSR.ACTIVE bit goes high when the byte transmission starts and low when it ends. When the transmission ends, the UxCSR.TX_BYTE bit is set to 1. An interrupt request is generated when the UxDBUF register is ready to accept new transmit data. This happens immediately after the transmission has been started; hence, a new data byte value can be loaded into the data buffer while the byte is being transmitted.

#### 17.1.2 UART Receive

Data reception on the UART is initiated when a 1 is written to the UxCSR.RE bit. The UART then searches for a valid start bit on the RXDx input pin and sets the UxCSR.ACTIVE bit high. When a valid start bit has been detected, the received byte is shifted into the receive register. The UxCSR.RX_BYTE bit is set and a receive interrupt is generated when the operation has completed. At the same time, UxCSR.ACTIVE goes low.

The received data byte is available through the UxDBUF register. When UxDBUF is read, UxCSR.RX_BYTE is cleared by hardware.

**NOTE:** When the application has read UxDBUF, it is important that it does not clear UxCSR.RX_BYTE. Clearing UxCSR.RX_BYTE implicitly makes the UART believe that the UART RX shift register is empty, even though it might hold pending data (typically due to back-to-back transmission). Consequently, the UART asserts (TTL low) the RT/RTS line, which allows flow into the UART, leading to potential overflow. Hence, the UxCSR.RX_BYTE flag integrates closely with the automatic RT/RTS function and must therefore be controlled solely by the SoC UART itself. Otherwise, the application could typically experience that the RT/RTS line remains asserted (TTL low), even though a back-to-back transmission clearly suggests it ought to intermittently pause the flow.

#### 17.1.3 UART Hardware Flow Control

Hardware flow control is enabled when the UXUCR.FLOW bit is set to 1. The RTS output is driven low when the receive register is empty and reception is enabled. Transmission of a byte does not occur before the CTS input goes low.



#### 17.1.4 UART Character Format

If the BIT9 and PARITY bits in register UxUCR are set high, parity generation and detection is enabled. The parity is computed and transmitted as the ninth bit, and during reception, the parity is computed and compared to the received ninth bit. If there is a parity error, the UxCSR.ERR bit is set high. This bit is cleared when UxCSR is read.

The number of stop bits to be transmitted is set to one or two bits, as determined by the register bit UxUCR.SPB. The receiver always checks for one stop bit. If the first stop bit received during reception is not at the expected stop bit level, a framing error is signaled by setting register bit UxCSR.FE high. UxCSR.FE is cleared when UxCSR is read. The receiver checks both stop bits when UxUCR.SPB is set. Note that the RX interrupt is set when the first stop bit is checked OK. If second stop bit is not OK, there is a delay in setting the framing error bit, UxCSR.FE. This delay is baud-rate dependent (bit duration).

#### 17.2 SPI Mode

This section describes the SPI mode of operation for synchronous communication. In SPI mode, the USART communicates with an external system through a three-wire or four-wire interface. The interface consists of the pins MOSI, MISO, SCK, and SS_N. See Section 7.6 for a description of how the USART pins are assigned to the I/O pins.

The SPI mode includes the following features:

- Three-wire (master) and four-wire SPI interface
- Master and slave modes
- Configurable SCK polarity and phase
- · Configurable LSB- or MSB-first transfer

The SPI mode is selected when UxCSR.MODE is set to 0.

In SPI mode, the USART can be configured to operate either as a SPI master or as a SPI slave by writing the  $\tt UxCSR.SLAVE$  bit.

#### 17.2.1 SPI Master Operation

A SPI byte transfer in master mode is initiated when the UxDBUF register is written. The USART generates the SCK serial clock using the baud-rate generator (see Section 17.4) and shifts the provided byte from the transmit register onto the MOSI output. At the same time, the receive register shifts in the received byte from the MISO input pin.

The UxCSR.ACTIVE bit goes high when the transfer starts and low when the transfer ends. When the transfer ends, the UxCSR.TX_BYTE bit is set to 1.

The polarity and clock phase of the serial clock SCK is selected by UxGCR.CPOL and UxGCR.CPHA. The order of the byte transfer is selected by the UxGCR.ORDER bit.

At the end of the transfer, the received data byte is available for reading from the UxDBUF. A receive interrupt is generated when this new data is ready in the UxDBUF USART receive/transmit data register.

A transmit interrupt is generated when the unit is ready to accept another data byte for transmission. Because UxDBUF is double-buffered, this happens just after the transmission has been initiated. Note that data should not be written to UxDBUF until UxCSR.TX_BYTE is 1. For DMA transfers, this is handled automatically. For back-to-back transmits using DMA, the UxGCR.CPHA bit must be set to zero; if not, transmitted bytes can become corrupted. For systems requiring setting of UxGCR.CPHA, polling UxCSR.TX_BYTE is needed.

Also, note the difference between transmit interrupt and receive interrupt, as the former arrives approximately eight bit-periods prior to the latter.

SPI master-mode operation as described previously is a three-wire interface. No select input is used to enable the master. If the external slave requires a slave-select signal, this can be implemented through software using a general-purpose I/O pin.



SSN Slave-Select Pin

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#### 17.2.2 SPI Slave Operation

A SPI byte transfer in slave mode is controlled by the external system. The data on the MOSI input is shifted into the receive register controlled by the serial clock, SCK, which is an input in slave mode. At the same time, the byte in the transmit register is shifted out onto the MISO output.

The UXCSR.ACTIVE bit goes high when the transfer starts and low when the transfer ends. Then the UXCSR.RX_BYTE bit is set and a receive interrupt is generated.

The expected polarity and clock phase of SCK is selected by UXGCR.CPOL and UXGCR.CPHA. The expected order of the byte transfer is selected by the UXGCR.ORDER bit.

At the end of the transfer, the received data byte is available for reading from UxDBUF.

The transmit interrupt is generated at the start of the operation.

#### 17.3 SSN Slave-Select Pin

When the USART is operating in SPI mode, configured as a SPI slave, a four-wire interface is used with the slave-select (SSN) pin as an input to the SPI. When SSN is low, the SPI slave is active, receives data on the MOSI input, and outputs data on the MISO output. When SSN is high, the SPI slave is inactive and does not receive data. The MISO output is in the high-impedance state when SSN is high. Also note that the release of SSN (SSN going high) must be aligned to the end of the byte received or sent. If released during a byte, the next received byte is not received properly, as information about the previous byte is present in the SPI system. A USART flush can be used to remove this information.

In SPI master mode, the SSN pin is not used. When the USART operates as a SPI master and a slave-select signal is required by an external SPI slave device, then a general-purpose I/O pin should be used to implement the slave-select signal function in software.

#### 17.4 Baud-Rate Generation

An internal baud-rate generator sets the UART baud rate when operating in UART mode and the SPI master clock frequency when operating in SPI mode.

The UxBAUD.BAUD_M[7:0] and UxGCR.BAUD_E[4:0] registers define the baud rate used for UART transfers and the rate of the serial clock for SPI transfers. The baud rate is given by the following equation:

Baud Rate = 
$$\frac{(256 + BAUD_M) \times 2^{BAUD_E}}{2^{28}} \times f$$

(3)

where f is the system clock frequency, 16 MHz for the RCOSC or 32 MHz for the XOSC.

The register values required for standard baud rates are shown in Table 17-1 for a typical system clock set to 32 MHz. The table also gives the difference in actual baud rate to standard baud rate value as a percentage error.

The maximum baud rate for the UART mode is f/16 when BAUD_E is 16 and BAUD_M is 0, and where f is the system clock frequency.

See the device data sheet for the maximum baud rate in SPI mode.

Note that the baud rate must be set through the UxBAUD and UxGCR registers before any other UART or SPI operations take place. If the baud rate is changed while in UART mode, it may take up to one bit period of the old baud rate before the change takes effect.

Baud Rate (bps)	UxBAUD.BAUD_M	UxGCR.BAUD_E	Error (%)
2400	59	6	0.14
4800	59	7	0.14
9600	59	8	0.14
14,400	216	8	0.03
19,200	59	9	0.14

 Table 17-1. Commonly Used Baud-Rate Settings for 32 MHz System Clock

Baud Rate (bps)	UxBAUD.BAUD_M	UxGCR.BAUD_E	Error (%)
28,800	216	9	0.03
38,400	59	10	0.14
57,600	216	10	0.03
76,800	59	11	0.14
115,200	216	11	0.03
230,400	216	12	0.03

# Table 17-1. Commonly Used Baud-Rate Settings for 32 MHz System Clock (continued)

#### 17.5 USART Flushing

The current operation can be aborted by setting the UxUCR.FLUSH register bit. This event stops the current operation and clears all data buffers. It should be noted that when setting the flush bit in the middle of a TX/RX bit, the flushing does not take place until this bit has ended (buffers are cleared immediately, but timers keeping knowledge of bit duration are not). Thus, using the flush bit should either be aligned with USART interrupts or use a wait time of one bit duration at the current baud rate before updated data or configuration can be received by the USART.

#### 17.6 USART Interrupts

Each USART has two interrupts. These are the RX complete interrupt (URXx) and the TX interrupt (UTXx). The TX interrupt is triggered when transmission starts and the data buffer is offloaded.

The USART interrupt enable bits are found in the IEN0 and IEN2 registers. The interrupt flags are located in the TCON and IRCON2 registers. See Section 2.5 for details of these registers. The interrupt enables and flags are summarized as follows.

Interrupt enables:

- USARTO RX: IENO.URX0IE
- USART1 RX: IEN0.URX1IE
- USARTO TX: IEN2.UTX0IE
- USART1 TX: IEN2.UTX1IE

Interrupt flags:

- USARTO RX: TCON.URX0IF
- USART1 RX: TCON.URX11F
- USARTO TX: IRCON2.UTX0IF
- USART1 TX: IRCON2.UTX1IF

#### 17.7 USART DMA Triggers

There are two DMA triggers associated with each USART. The DMA triggers are activated by RX complete and TX complete events, i.e., the same events as the USART interrupt requests. A DMA channel can be configured using a USART receive/transmit buffer, UxDBUF, as source or destination address.

See Table 8-1 for an overview of the DMA triggers.

#### 17.8 USART Registers

The registers for the USART are described in this section. For each USART there are five registers consisting of the following (x refers to the USART number, i.e., 0 or 1):

- UxCSR, USART x control and status
- UXUCR, USART x UART control



- UxGCR, USART x generic control
- UxDBUF, USART x receive/transmit data buffer
- UxBAUD, USART x baud-rate control

Bit	Name	Reset	R/W	Description
7	MODE	0	R/W	USART mode select
				0: SPI mode
				1: UART mode
6	RE	0	R/W	UART receiver enable. Note: Do not enable receive before UART is fully configured.
				0: Receiver disabled
				1: Receiver enabled
5	SLAVE	0	R/W	SPI master or slave mode select
				0: SPI master
				1: SPI slave
4	FE	0	R/W0	UART framing error status. This bit is automatically cleared on a read of the UOCSR register or bits in the UOCSR register.
				0: No framing error detected
				1: Byte received with incorrect stop-bit level
3	ERR	0	R/W0	UART parity error status. This bit is automatically cleared on a read of the UOCSR register or bits in the UOCSR register.
				0: No parity error detected
				1: Byte received with parity error
2	RX_BYTE	0	R/W0	Receive byte status. UART mode and SPI slave mode. This bit is automatically cleared when reading U0DBUF; clearing this bit by writing 0 to it effectively discards the data in U0DBUF.
				0: No byte received
				1: Received byte ready
1	TX_BYTE	0	R/W0	Transmit byte status. UART mode and SPI master mode
				0: Byte not transmitted
				1: Last byte written to data-buffer register has been transmitted
0	ACTIVE	0	R	USART transmit/receive active status. In SPI slave mode, this bit equals slave select.
				0: USART idle
				1: USART busy in transmit or receive mode

Bit	Name	Reset	R/W	Description
7	FLUSH	0	R0/W1	Flush unit. When set, this event stops the current operation and returns the unit to the idle state.
6	FLOW	0	R/W	UART hardware flow enable. Selects use of hardware flow control with RTS and CTS pins
				0: Flow control disabled
				1: Flow control enabled
5	D9	0	R/W	If parity is enabled (see PARITY, bit 3 in this register), then this bit sets the parity level as follows:
				0: Odd parity
				1: Even parity
4	BIT9	0	R/W	Set this bit to 1 in order to enable the parity bit tranfer (as 9th bit). The content of this 9th bit is given by D9, if parity is enabled by PARITY.
				0: 8-bit transfer
				1: 9-bit transfer
3	PARITY	0	R/W	UART parity enable. One must set BIT9 in addition to setting this bit for parity to be calculated.
				0: Parity disabled
				1: Parity enabled
2	SPB	0	R/W	UART number of stop bits. Selects the number of stop bits to transmit
				0: 1 stop bit
				1: 2 stop bits
1	STOP	1	R/W	UART stop-bit level must be different from the start-bit level
				0: Low stop bit
				1: High stop bit
0	START	0	R/W	UART start-bit level. Ensure that the polarity of the start bit is opposite the level of the idle line.
				0: Low start bit
				1: High start bit

#### U0GCR (0xC5) - USART 0 Generic Control

Bit	Name	Reset	R/W	Description	
7	CPOL	0	R/W	SPI clock polarity	
				0: Negative clock polarity	
				1: Positive clock polarity	
6	СРНА	0	R/W	SPI clock phase	
				0: Data is output on <i>MOSI</i> when <i>SCK</i> goes from <b>CPOL</b> inverted to <b>CPOL</b> , and data input is sampled on <i>MISO</i> when <i>SCK</i> goes from <b>CPOL</b> to <b>CPOL</b> inverted.	
				1: Data is output on <i>MOSI</i> when <i>SCK</i> goes from <b>CPOL</b> to <b>CPOL</b> inverted, and data input is sampled on <i>MISO</i> when <i>SCK</i> goes from <b>CPOL</b> inverted to <b>CPOL</b> .	
5	ORDER	0	R/W	Bit order for transfers	
				0: LSB first	
				1: MSB first	
4:0	BAUD_E[4:0]	0 0000	R/W	Baud rate exponent value. BAUD_E along with BAUD_M determines the UART baud rate and the SPI master SCK clock frequency.	

#### U0DBUF (0xC1) - USART 0 Receive/Transmit Data Buffer

Bit	Name	Reset	R/W	Description
7:0	DATA[7:0]	0x00	R/W	USART receive and transmit data. When writing this register, the data written is written to the internal transmit-data register. When reading this register, the data from the internal read-data register is read.

#### U0BAUD (0xC2) - USART 0 Baud-Rate Control

Bit	Name	Reset	R/W	Description	
7:0	BAUD_M[7:0]	0x00	R/W	Baud-rate mantissa value. BAUD_E along with BAUD_M decides the UART baud rate and the SPI master SCK clock frequency.	



USART Registers

Bit	Name	Reset	R/W	Description
7	MODE	0	R/W	USART mode select
				0: SPI mode
				1: UART mode
6	RE	0	R/W	UART receiver enable. Note: Do not enable receive before UART is fully configured.
				0: Receiver disabled
				1: Receiver enabled
5	SLAVE	0	R/W	SPI master- or slave-mode select
				0: SPI master
				1: SPI slave
4	FE	0	R/W0	UART framing error status. This bit is automatically cleared on a read of the UICSR register or bits in the UICSR register.
				0: No framing error detected
				1: Byte received with incorrect stop-bit level
3	ERR	0	R/W0	UART parity error status. This bit is automatically cleared on a read of the UICSR register or bits in the UICSR register.
				0: No parity error detected
				1: Byte received with parity error
2	RX_BYTE	0	R/W0	Receive byte status. UART mode and SPI slave mode. This bit is automatically cleared when reading UldBUF; clearing this bit by writing 0 to it effectively discards the data in UldBUF.
				0: No byte received
				1: Received byte ready
1	TX_BYTE	0	R/W0	Transmit byte status. UART mode and SPI master mode
				0: Byte not transmitted
				1: Last byte written to data buffer register has been transmitted
0	ACTIVE	0	R	USART transmit/receive active status. In SPI slave mode, this bit equals slave select.
				0: USART idle
				1: USART busy in transmit or receive mode

#### U1UCR (0xFB) – USART 1 UART Control

Bit	Name	Reset	R/W	Description
7	FLUSH	0	R0/W1	Flush unit. When set, this event stops the current operation and returns the unit to the idle state.
6	FLOW	0	R/W	UART hardware flow enable. Selects use of hardware flow control with RTS and CTS pins
				0: Flow control disabled
				1: Flow control enabled
5	D9	0	R/W	If parity is enabled (see PARITY, bit 3 in this register), then this bit sets the parity level as follows.
				0: Odd parity
				1: Even parity
4	BIT9	0	R/W	Set this bit to 1 in order to enable the parity bit tranfer (as 9th bit). The content of this 9th bit is given by D9, if parity is enabled by PARITY.
				0: 8-bit transfer
				1: 9-bit transfer
3	PARITY	0	R/W	UART parity enable. One must set BIT9 in addition to setting this bit for parity to be calculated.
				0: Parity disabled
				1: Parity enabled
2	SPB	0	R/W	UART number of stop bits. Selects the number of stop bits to transmit
				0: 1 stop bit
				1: 2 stop bits
1	STOP	1	R/W	UART stop-bit level must be different from start-bit level.
				0: Low stop bit
				1: High stop bit
0	START	0	R/W	UART start-bit level. Ensure that the polarity of the start bit is opposite the level of the idle line.
				0: Low start bit
				1: High start bit



U1G	CR (0xFC) – USA	RT 1 Gen	eric Con	trol	
Bit	Name	Reset	R/W	Description	
7	CPOL	0	R/W	SPI clock polarity	
				0: Negative clock polarity	
				1: Positive clock polarity	
6	CPHA	0	R/W	SPI clock phase	
				0: Data is output on <i>MOSI</i> when <i>SCK</i> goes from <b>CPOL</b> inverted to <b>CPOL</b> , and data input is sampled on <i>MISO</i> when <i>SCK</i> goes from <b>CPOL</b> to <b>CPOL</b> inverted.	
				1: Data is output on <i>MOSI</i> when <i>SCK</i> goes from <b>CPOL</b> to <b>CPOL</b> inverted, and data input is sampled on <i>MISO</i> when <i>SCK</i> goes from <b>CPOL</b> inverted to <b>CPOL</b> .	
5	ORDER	0	R/W	Bit order for transfers	
				0: LSB first	
				1: MSB first	
4:0	BAUD_E[4:0]	0 0000	R/W	Baud rate exponent value. BAUD_E along with BAUD_M determines the UART baud rate and the SPI master SCK clock frequency.	

#### U1DBUF (0xF9) – USART 1 Receive/Transmit Data Buffer

Bit	Name	Reset	R/W	Description
7:0	DATA[7:0]	0x00	R/W	USART receive and transmit data. When writing this register, the data written is written to the internal transmit-data register. When reading this register, the data from the internal read-data register is read.

#### U1BAUD (0xFA) – USART 1 Baud-Rate Control

Bit	Name	Reset	R/W	Description	
7:0	BAUD_M[7:0]	0x00		Baud rate mantissa value. BAUD_E along with BAUD_M determines the UART baud rate and the SPI master SCK clock frequency.	



# **Operational Amplifier**

The operational amplifier (in the CC2530, CC2531, and CC2540) has the following features:

- Low offset
- Ideal for use in combination with the onboard ADC in sensor applications

# Topic Page 18.1 Description 174 18.2 Calibration 174 18.3 Clock Source 174 18.4 Registers 174

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Description

#### 18.1 Description

The operational amplifier is connected to the I/O pins as follows:

- The positive input pin is connected to P0_0.
- The negative input pin is connected to P0_1.
- The output is connected to P0_2.

The pins used by the operational amplifier must be configured as analog pins, by setting bits APCFG[2:0] to 1. The OPAMPC.EN bit is used to enable/disable the operational amplifier. When power mode 2/3 is entered, the operational amplifier is shut down automatically and must be restarted when entering PMO again.

#### 18.2 Calibration

The operational amplifier must be calibrated. A calibration is started by writing 1 to OPAMPC.CAL. During calibration, OPAMPS.CAL_BUSY is 1. A new calibration is not accepted before OPAMPS.CAL_BUSY goes low. Every time after enabling the operational amplifier, calibration must be performed.

#### 18.3 Clock Source

The operational amplifier uses a divided version of the system clock. The division factor depends on which clock source is used, HSOSC or XOSC. While the operational amplifier is enabled, the clock source should not be changed.

#### 18.4 Registers

This section describes the registers for the operational amplifier.

Bit	Name	Reset	R/W	Description
7:2	-	0000 00	R/W	Reserved. Always write 0000 00.
1:0	MODE	00	R/W	Operational amplifier mode
				00 and 01: Non-chop mode – Higher offset (~500 $\mu$ V), but no chopper ripple. Use in conjunction with Mode 10 if offset cancellation is required. Offset for these two modes is the opposite of the offset seen in Mode 10.
				10: Non-chop mode – Higher offset (~500 $\mu$ V), but no chopper ripple. Use in conjunction with Mode 00 or Mode 01 to double sample and correct for the offset by averaging the two samples.
				11: Chop mode – Very low offset (~ 50 $\mu V$ ), and very low noise (1/f noise shifted to 1 MHz due to chopping), and 1 MHz ripple

#### OPAMPMC (CC2530/CC2531: 0x61A6. CC2540: 0x61AD) – Operational Amplifier Mode Control

#### **OPAMPC (0x62C0)** – Operational Amplifier Control

Bit	Name	Reset	R/W	Description
7:2	-	0000 00	R0	Reserved
1	CAL	0	W1/R0	Start calibration. Calibration only starts if OPAMPC.EN is 1.
0	EN	0	R/W	Operational amplifier enable

#### **OPAMPS (0x62C1)** – Operational Amplifier Status

Bit	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved
0	CAL_BUSY	0	R	Calibration in progress



# **Analog Comparator**

The analog comparator (in the CC2530, CC2531, CC240 and CC2541) has the following features:

- Low-power operation
- Wake-up source

Topic

Page

19.1	Description	176
19.2	Register	176

#### 19.1 Description

The analog comparator is connected to the I/O pins as follows:

- The positive input pin is connected to P0_5.
- The negative input pin is connected to P0 4.
- The output can be read from CMPCTL.OUTPUT.

The comparator pins must be configured as analog pins by setting bits APCFG[5:4] to 1. The CMPCTL.EN bit is used to enable/disable the comparator. The output from the comparator is connected internally to the edge detector that controls POIFG[5]. This makes it possible to associate an I/O interrupt with a rising/falling edge on the comparator output. When enabled, the comparator remains active while in power mode 2 or 3. Thus, it is possible to wake up from power mode 2/3 on a rising or falling edge on the comparator output.

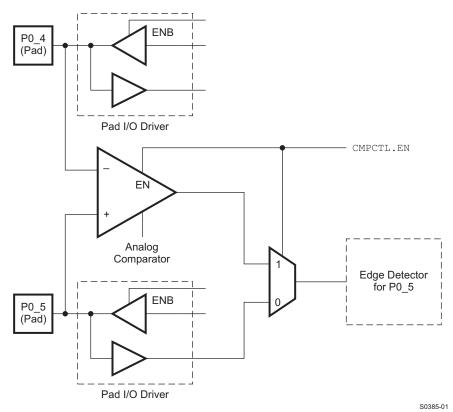


Figure 19-1. Analog Comparator

#### 19.2 Register

This section describes the registers for the analog comparator.

	own one (oxeed) Analog comparator control and clauds								
Bit	Name	Reset	R/W	Description					
7:2	-	0000 00	R0	Reserved					
1	EN	0	R/W	Comparator enable					
0	OUTPUT	0	R	Comparator output					

CMPCTL (0x62D0) - Analog Comparator Control and Status



Page

The  $l^2C$  module (in the CC2533 and CC2541) provides an interface between the device and  $l^2C$ -compatible devices connected by the two-wire  $l^2C$  serial bus. External components attached to the  $l^2C$  bus serially transmit and/or receive serial data to/from the  $l^2C$  module through the two-wire  $l^2C$  interface.

The I²C module features include:

- Compliance with the I²C specification v2.1 (published by Philips Semiconductor)
- 7-bit device addressing modes
- General call
- START/RESTART/STOP
- Multi-master transmitter/receiver mode
- Slave receiver/transmitter mode
- Standard mode up to 100-kbps and fast mode up to 400-kbps support

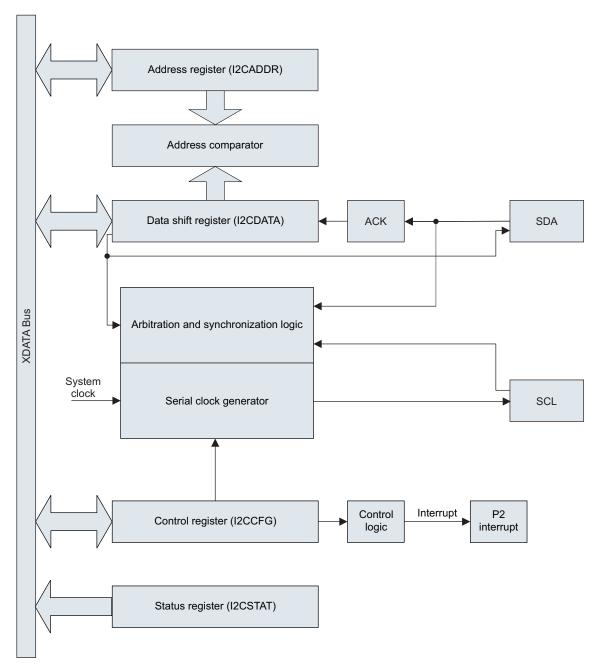
Figure 20-1 shows the block diagram of the I²C module.

On the CC2533 and CC2541, the  $I^2C$  module is connected to pins 2 and 3 and uses the P2 interrupt to the CPU. Pins 2 and 3 can alternatively be controlled as two GPIO pins if they are not used by the  $I^2C$  module.

#### Topic

20.1	Operation	178
20.2	I ² C Registers	188







#### 20.1 Operation

The I²C module supports any slave or master I²C-compatible device. Figure 20-2 shows an example of an I²C bus. Each I²C device is recognized by a unique address and can operate as either a transmitter or a receiver. A device connected to the I²C bus can be considered as the master or the slave when performing data transfers. A master initiates a data transfer and generates the clock signal, SCL. Any device addressed by a master is considered a slave.

I²C data is communicated using the serial data (SDA) pin and the serial clock (SCL) pin. Both SDA and SCL are bidirectional and must be connected to a positive supply voltage using a pullup resistor.



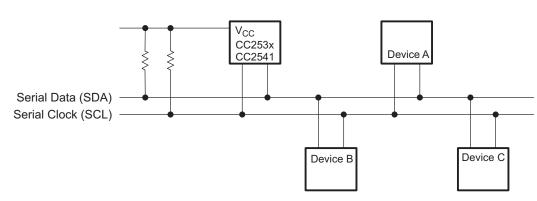


Figure 20-2. I²C Bus Connection Diagram

#### 20.1.1 fC Initialization and Reset

The I²C module is enabled by setting the I2CCFG.ENS1 bit. It is then in the not-addressed slave state.

The I²C configuration and state is not retained in power modes PM2 and PM3. It must be reconfigured after coming out of sleep mode.

The I²C module is not reset when disabled, and retains its internal state until the next time I2CCFG.ENS1 is set.

#### 20.1.2 PC Serial Data

One clock pulse is generated by the master device for each data bit transferred. The I²C module operates with byte data. Data is transferred MSB first as shown in Figure 20-3.

The first byte after a START condition consists of a 7-bit slave address and the R/W bit. When R/W = 0, the master transmits data to a slave. When R/W = 1, the master receives data from a slave. The ACK bit is sent from the receiver after each byte on the ninth SCL clock.

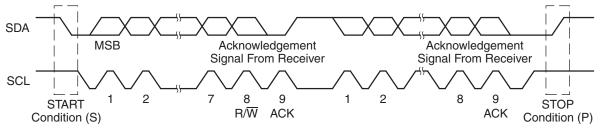


Figure 20-3. I²C Module Data Transfer

START and STOP conditions are generated by the master and are shown in Figure 20-3. A START condition is a high-to-low transition on the SDA line while SCL is high. A STOP condition is a low-to-high transition on the SDA line while SCL is high.

Data on SDA must be stable during the high period of SCL (see Figure 20-4). The state of SDA can only change when SCL is low, otherwise a START or STOP condition is generated.



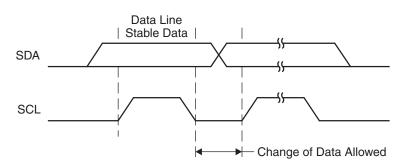


Figure 20-4. Bit Transfer on I²C Bus

#### 20.1.3 PC Addressing Modes

The I²C module supports 7-bit addressing mode.

#### 20.1.3.1 7-Bit Addressing

In the 7-bit addressing format (see Figure 20-5), the first byte is the 7-bit slave address and the R/W bit. The ACK bit is sent from the receiver after each byte.

1	← 7 →	1	1	8	1	8	1	1
S	Slave Address	R/W	ACK	Data	ACK	Data	ACK	Ρ

#### Figure 20-5. I²C Module 7-Bit Addressing Format

#### 20.1.3.2 Repeated Start Conditions

The direction of data flow on SDA can be changed by the master, without first stopping a transfer, by issuing a repeated START condition. This is called a RESTART. After a RESTART is issued, the slave address is again sent out with the new data direction specified by the R/W bit. The RESTART condition is shown in Figure 20-6.

1	₹ 7 •	1	1	<b>₩</b> 8	1	1	7	1	1	8	▶ 1  1
S	Slave Address	R/W	ACK	Data	ACK	S	Slave Addres	s R/W	ACK	Data	ACK P
	4			Any Numb			<b> ↓</b> 1			Any Numbe	er

Figure 20-6. I²C Module Addressing Format With Repeated START Condition

#### 20.1.4 PC Module Operating Modes

The I²C module can operate in master transmitter, master receiver, slave transmitter, or slave receiver mode. The modes are discussed in the following sections.

#### 20.1.4.1 Slave Mode

Initially, the I²C module is configured in receiver mode by setting the <code>l2CCFG.ENS1</code> bit to receive the I²C address. Afterwards, transmit and receive operations are controlled automatically, depending on the R/W bit received, together with the slave address.

The  $l^2C$  slave address is programmed with the <code>l2CADDR.ADDR</code> bits. The value of the <code>l2CADDR.GC</code> bit determines whether the slave responds to a general call.



When a START condition is detected on the bus, the I²C module receives the transmitted address and compares it against its own address stored in I2CADDR. ADDR. If the compare is successful, an interrupt is generated and the I2CCFG.SI bit is set. The same is done for a general call address match if the I2CADDR.GC bit is set.

## 20.1.4.1.1 ^fC Slave Transmitter Mode

Slave transmitter mode is entered when the slave address transmitted by the master is identical to this device's own address with a set R/W bit. The slave transmitter shifts the serial data out on SDA with the clock pulses that are generated by the master device. The slave device does not generate the clock, but it does hold SCL low while intervention of the CPU is required after a byte has been transmitted.

If the master requests data from the slave, the I²C module is automatically configured as a transmitter, and I2CCFG.SI is set. The SCL line is held low until the first data to be sent is written into the data buffer I2CDATA. Then the address is acknowledged and the data is transmitted. After the data is acknowledged by the master, the bus is stalled during the acknowledge cycle by holding SCL low until new data is written into I2CDATA. If the master sends a NACK the I²C module returns to the not-addressed slave state.

Table 20-1 provides more details regarding the slave transmitter operation.

Status	Status of the	Application S	Softwa	re Res	ponse		
Code (Value of				<b>To</b> 12	2CCFG		Next Action Taken by I ² C Hardware
12CSTAT. STAC)	I ² C	To/From 12CDATA	STA	STO	SI	AA	
0xA8	Own SLA+R has been received; ACK has been returned.	Load data byte or load data byte	X X	0 0	0 0	0 1	Last data byte is transmitted and ACK is received. Data byte is transmitted; ACK is received.
0xB0	Arbitration lost in SLA+R/W as master; own SLA+R has been received; ACK has been returned.	SLA+R/W as ister; own A+R has en received; K has been		0	0	0 1	Last data byte is transmitted and ACK is received. Data byte is transmitted; ACK is received.
0xB8	Data byte has been transmitted; ACK has been received.	Load data byte or load data byte	X X	0 0	0 0	0 1	Last data byte is transmitted and ACK is received Data byte is transmitted; ACK is received.
0xC0	Data byte has been transmitted; not-ACK has been received.	No action or no action or no action	0 0 1	0 0 0	0 0 0	0 1 0	Switched to not-addressed SLV mode; no recognition of own SLA or general call address Switched to not-addressed SLV mode; own SLA or general call address is recognized. Switched to not-addressed SLV mode; no recognition of own SLA or general call address;
		or no action	1	0	0	1	START condition is transmitted when the bus becomes free. Switched to not-addressed SLV mode; own SLA or general-call address is recognized; START condition is transmitted when the bus becomes free.

Table 20-1. Slave Transmitter Mode

Status		Application S	Softwa	re Res	ponse		
Code (Value of	Status of the			ToI	2CCFG		Next Action Taken by I ² C Hardware
I2CSTAT. STAC)	I ² C	To/From 12CDATA	STA	STO	SI	AA	
0xC8	Last data byte has been	No action	0	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general call address
	transmitted; ACK has been received.	or no action	0	0	0	1	Switched to not-addressed SLV mode; own SLA or general call address is recognized.
		or no action	1	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general call address; START condition is transmitted when the bus becomes free.
		or no action	1	0	0	1	Switched to not-addressed SLV mode; own SLA or general call address is recognized; START condition is transmitted when the bus becomes free.

#### Table 20-1. Slave Transmitter Mode (continued)

## 20.1.4.1.2 *PC Slave Receiver Mode*

Slave receiver mode is entered when the slave address transmitted by the master is identical to its own address and a cleared R/W bit is received. In slave receiver mode, serial data bits received on SDA are shifted in with the clock pulses that are generated by the master device. The slave device does not generate the clock, but it can hold SCL low if intervention of the CPU is required after a byte has been received.

If the slave interrupt is triggered from the master, the  $l^2C$  module is automatically configured as a receiver and I2CCFG.SI is set. After the first data byte is received, the interrupt flag I2CCFG.SI is set again. The  $l^2C$  module automatically acknowledges the received data.

While the I2CCFG.SI flag is set, the bus is stalled by holding SCL low.

When the master generates a STOP condition, the I2CCFG.STO flag is set.

If the master generates a repeated START condition, the I²C state machine returns to its address reception state.

Table 20-2 provides more details regarding slave receiver operation.

Status		Application S	Softwa	re Res	ponse		
Code (Value of	Status of the			<b>To</b> 12	2CCFG		Next Action Taken by I ² C Hardware
12CSTAT. STAC)	l²C	To/From 12CDATA	STA	STO	SI	AA	
0x60	Own SLA+W	No action	Х	0	0	0	Data byte is received and not-ACK is returned
	has been received; ACK has been returned.	or no action	х	0	0	1	Data byte is received and ACK is returned.
i	Arbitration lost in SLA+R/W as master; own SLA+W has been received, ACK returned.	No action	Х	0	0	0	Data byte is received and not-ACK is returned.
		or no action	x	0	0	1	Data byte is received and ACK is returned.
0x70	General-call	No action	Х	0	0	0	Data byte is received and not-ACK is returned.
	address (0x00) has been received; ACK has been returned	or no action	x	0	0	1	Data byte is received and ACK is returned.

Table 20-2. Slave Receiver Mode

Status		Application S	Softwa	re Res	ponse		
Code (Value of	Status of the			<b>To</b> 12	2CCFG	T	Next Action Taken by I ² C Hardware
12CSTAT. STAC)	l²C	To/From I2CDATA	STA	STO	SI	AA	
0x78	Arbitration lost in SLA+R/W as master; general-call address has been received, ACK returned.	No action or no action	X X	0 0	0	0 1	Data byte is received and not-ACK is returned. Data byte is received and ACK is returned.
0x80	Previously addressed with own SLV address; DATA has been received, ACK returned.	Read data byte or read data byte	X X	0	0	0 1	Data byte is received and not-ACK is returned. Data byte is received and ACK is returned.
0x88	Previously addressed with own SLA;	Read data byte	0	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address
	DATA byte has been received,	or read data byte	0	0	0	1	Switched to not-addressed SLV mode; own SLA or general call address is recognized.
	not-ACK returned.	or read data byte	1	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address; START condition is transmitted when the bus becomes free.
		or read data byte	1	0	0	1	Switched to not-addressed SLV mode; own SLA or general-call address is recognized; START condition is transmitted when the bus becomes free.
0x90	Previously	Read data byte	Х	0	0	0	Data byte is received and not-ACK is returned.
	addressed with general-call address; DATA has been received, ACK returned.	or read data byte	x	0	0	1	Data byteis received and ACK is returned.
0x98	Previously addressed with	Read data byte	0	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address
	own SLA; DATA byte has been received,	or read data byte	0	0	0	1	Switched to not-addressed SLV mode; own SLA or general-call address is recognized.
	not-ACK returned.	or read data byte	1	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address; START condition is transmitted when the bus becomes free.
		or read data byte	1	0	0	1	Switched to not-addressed SLV mode; own SLA or general-call address is recognized; START condition is transmitted when the bus becomes free.
0xA0	A STOP condition or	No action	0	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address
	repeated START condition has	or no action	0	0	0	1	Switched to not-addressed SLV mode; own SLA or general-call address is recognized.
	been received while still addressed as SLV/REC or	or no action	1	0	0	0	Switched to not-addressed SLV mode; no recognition of own SLA or general-call address; START condition is transmitted when the bus becomes free.
	SLV/TRX.	or no action	1	0	0	1	Switched to not-addressed SLV mode; own SLA o general-call address is recognized; START condition is transmitted when the bus becomes free.

Table 20-2. Slave Receiver Mode (continued)



### 20.1.4.2 Master Mode

The I²C module is configured as an I²C master by setting the <code>l2CCFG.ENS1</code> and <code>l2CCFG.STA</code> bits. When the master is part of a multi-master system, its own address must be programmed into the <code>l2CADDR.ADDR</code> register. The value of the <code>l2CADDR.GC</code> bit determines whether the I²C module responds to a general call.

#### 20.1.4.2.1 PC Master Transmitter Mode

To enable master transmit mode, set the I2CCFG. ENS1 and I2CCFG. STA bits. The I²C module then waits until the I²C bus is free. When the I²C bus is free, it generates a START condition, sends the slave address, and transfers a transmit direction bit. It then generates an interrupt, and the first byte of data can be written to the I2CDATA register. The I²C core sends I2CDATA content if arbitration is not lost, and then generates another interrupt. The I2CSTAT register contains a value of 0x18 or 0x20, depending on the received ACK bit (see Table 20-3). If a not-ACK is received from the slave, the master must react with either a repeated START condition or a STOP condition. Setting I2CCFG.STA during transmission causes a repeated START condition to be transmitted. Setting I2CCFG.STO during transmission causes a STOP condition to be transmitted and the I2CCFG.STO bit to be reset.

Table 20-3 provides more details regarding the master transmitter operation.

Status		Application S	Softwa	re Res	ponse		
Code (Value of	Status of the			ToI	2CCFG		Next Action Taken by I ² C Hardware
I2CSTAT. STAC)	l ² C	To/From 12CDATA	STA	STO	SI	AA	
0x08	A START condition has been transmitted.	Load SLA+W	х	0	0	х	SLA+W is transmitted. ACK is received.
0x10	A repeated	Load SLA+W	Х	0	0	Х	As for START condition (0x08)
	START condition has been transmitted.	or Ioad SLA+R	х	0	0	х	SLA+W is transmitted; I ² C is switched to MST/REC mode.
0x18	SLA+W has	Load data byte	0	0	0	Х	Data byte is transmitted; ACK is received.
	been transmitted; ACK has been	or no action	1	0	0	х	Repeated START is transmitted.
	received.	or no action	0	1	0	х	STOP condition is transmitted; STO flag is reset.
		or no action	1	1	0	х	STOP condition followed by a START condition is transmitted; STO flag is reset.
0x20	SLA+W has	Load data byte	0	0	0	Х	Data byte is transmitted; ACK is received.
	been transmitted; not-ACK has	or no action	1	0	0	х	Repeated START is transmitted.
	been received.	or no action	0	1	0	х	STOP condition is transmitted; STO flag is reset.
		or no action	1	1	0	х	STOP condition followed by a START condition is transmitted; STO flag is reset.
0x28	Data byte is	Load data byte	0	0	0	Х	Data byte is transmitted; ACK is received.
	transmitted; ACK is received.	or no action	1	0	0	х	Repeated START is transmitted.
	10001104.	or no action	0	1	0	х	STOP condition is transmitted; STO flag is reset.
		or no action	1	1	0	х	STOP condition followed by a START condition is transmitted; STO flag is reset.

#### Table 20-3. Master Transmitter Mode

Status		Application S	Softwa	re Res	ponse	)	
Code (Value of	Status of the		To 12CCFG				Next Action Taken by I ² C Hardware
I2CSTAT. STAC)	l²C	To/From 12CDATA	STA	STO	SI	АА	
0x30	Data byte in	Data byte	0	0	0	Х	Data byte is transmitted; ACK is received.
	I2CDATA has been transmitted.	or no action	1	0	0	х	Repeated START is transmitted.
	transmitted.	or no action	0	1	0	х	STOP condition is transmitted; STO flag is reset.
		or no action	1	1	0	х	STOP condition followed by a START condition is transmitted; STO flag is reset.
0x38	Arbitration lost	No action	0	0	0	Х	I ² C bus is released; not-addressed slave is entered.
	in SLA+R/W or data bytes	or no action	1	0	0	х	A START condition is transmitted when the bus becomes free.

Table 20-3. Master Transmitter Mode (contin	ued)
---------------------------------------------	------

## 20.1.4.2.2 f^C Master Receiver Mode

To enable master receive mode, set the <code>l2CCFG.ENS1</code> and the <code>l2CCFG.STA</code> bits. The <code>l²C</code> module then waits until the <code>l²C</code> bus is free. When the <code>l²C</code> bus is free, it generates a START condition, sends the slave address, and transfers a receive direction bit. It then generates an interrupt, and the first byte is received.

Table 20-4 provides more details regarding the master receiver operation.

Status		Application S	Softwa	re Res	ponse	1	
Code (Value of	Status of the			To I2	2CCFG		Next Action Taken by I ² C Hardware
12CSTAT. STAC)	I2C	To/From 12CDATA	STA	STO	SI	AA	
0x08	A START condition has been transmitted.	Load SLA+R	X	0	0	X	SLA+R is transmitted. ACK is received.
0x10	A repeated	Load SLA+R	Х	0	0	Х	As above
	START condition has been transmitted.	or Ioad SLA+W	х	0	0	Х	SLA+W is transmitted; I ² C is switched to MST/TRX mode.
0x38	Arbitration lost in not-ACK bit.	No action	0	0	0	Х	I ² C bus is released; I ² C enters slave mode.
		or no action	1	0	0	х	A start condition is transmitted when the bus becomes free.
0x40	SLA+R has	No action	0	0	0	0	Data byte is received; not-ACK is returned.
	been transmitted; ACK has been received.	or no action	0	0	0	1	Data byte is received; ACK is returned.
0x48	SLA+R has	No action	1	0	0	Х	Repeated START condition is transmitted.
	been transmitted; not-ACK has	or no action	0	1	0	x	STOP condition is transmitted; STO flag is reset.
	been received.	or no action	1	1	0	x	STOP condition followed by a START condition is transmitted; STO flag is reset.
0x50	Data byte has	Read data byte	0	0	0	0	Data byte is received; not-ACK is returned.
	been received; ACK has been returned.	or read data byte	0	0	0	1	Data byte is received; ACK is returned

#### Table 20-4. Master Receiver Mode



Status	Status of the	Application S	Softwar	e Res	ponse		
Code (Value of				<b>To</b> 12	CCFG		Next Action Taken by I ² C Hardware
12CSTAT. STAC)	I2C	To/From I2CDATA	STA	STO	SI	AA	
0x58	Data byte has	Read data byte	1	0	0	Х	Repeated START condition is transmitted.
	been received; not-ACK has been returned.	or read data byte	0	1	0	х	STOP condition is transmitted; STO flag is reset.
		or read data byte	1	1	0	х	STOP condition followed by a START condition is transmitted; STO flag is reset.

 Table 20-4. Master Receiver Mode (continued)

## 20.1.4.3 Arbitration

If two or more master transmitters simultaneously start a transmission on the bus, an arbitration procedure is invoked. Figure 20-7 shows the arbitration procedure between two devices. The arbitration procedure uses the data presented on SDA by the competing transmitters. The first master transmitter that generates a logic high is overruled by the opposing master generating a logic low. The arbitration procedure gives priority to the device that transmits the serial data stream with the lowest binary value. The master transmitter that lost arbitration switches to the slave receiver mode. If two or more devices send identical first bytes, arbitration continues on the subsequent bytes.

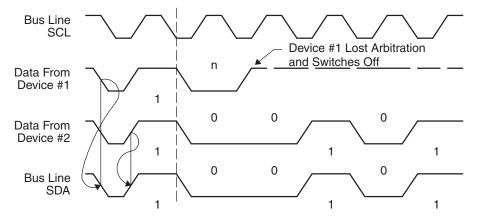


Figure 20-7. Arbitration Procedure Between Two Master Transmitters

## 20.1.5 fC Clock Generation and Synchronization

The I²C clock SCL is provided by the master on the I²C bus. When the I²C module is in master mode, the serial clock generator generates the SCL clock from the system clock. The serial clock generator is switched off when the I²C module is in slave mode.

The frequency of the SCL is determined by the system clock frequency, and the division factor given by the I2CCFG.CRx bits. Example frequencies for a 32-MHz system clock are given in the I2CCFG register description.

During the arbitration procedure, the clocks from the different masters must be synchronized. A device that first generates a low period on SCL overrules the other devices, forcing them to start their own low periods. SCL is then held low by the device with the longest low period. The other devices must wait for SCL to be released before starting their high periods. Figure 20-8 shows the clock synchronization. This allows a slow slave to slow down a fast master.



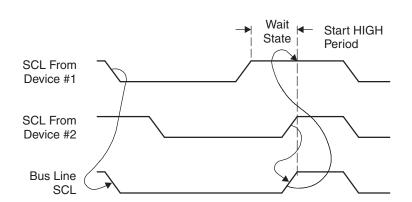


Figure 20-8. Synchronization of Two I²C Clock Generators During Arbitration

## 20.1.6 Bus Error

When an incorrect format of a frame is detected, a bus error condition is entered. The cause is that a START or STOP condition was detected during transfer of an address, data, or an acknowledge bit. When a bus error condition is entered, an interrupt is requested. The core leaves the bus error state when the I2CCFG.STO flag is set and the interrupt request is cleared. It goes into the slave mode and the I2CCFG.STO flag is automatically reset. The SDA and SCL lines are released (the STOP condition is not transmitted).

Status	Status of the I2C	Application S	Softwa	e Res	ponse		Next Action Taken by I ² C Hardware
Code (Value of		To/From 12CDATA		To 12	2CCFG		
I2CSTAT. STAC)			STA	STO	SI	AA	
0x00	Bus error during MST or selected slave modes	No action	0	1	0	Х	Only the internal hardware is affected in the MST or addressed SLV modes. In all cases, the bus is released and I ² C is switched to the not-addressed SLV mode. The Sto flag is reset.

Table 20-5. Miscellaneous States

## 20.1.7 ^fC Interrupt

The I²C module has an interrupt line to the CPU to signal that it requires service. The I²C module uses interrupt #6, which is also shared with Port 2 inputs; hence, the interrupt routine must also handle Port 2 interrupts if they are enabled.

For an interrupt request to be generated, IEN2.P2IE must be set to 1. When an interrupt request has been generated, the CPU starts executing the ISR if there are no higher-priority interrupts pending.

An interrupt is generated from the I²C module when one of the 26 out of 27 possible I²C component states is entered. The only state that does not cause an interrupt to be generated is state 0xF8, which indicates that no relevant state information is available. The corresponding I2CCFG.SI flag must be cleared by software at the end of the ISR.

## 20.1.8 *PC Pins*

The SCL and SDA pins of the I²C module are connected to pins 2 and 3, respectively, on the CC2533/CC2541. These pins are pulled up during reset to avoid floating pins. After reset, they are controlled by the I²C module and use an internal pullup resistor of 20 k $\Omega$  to hold bus signals high. If these pins are not to be used for I²C, they can be used as GPIO by setting the I2CWC.OVR bit. In this mode, pins 2 and 3 can be set up as outputs, as inputs with optional pullup, or as 4-mA drive-strength outputs like the other GPIO pads on the device by using the configuration bits in I2CWC. Their values are read or controlled using the I2CIO register. These pins cannot be configured to generate GPIO interrupts.

## 20.2 I²C Registers

This section describes all I²C registers used for control and status of the I²C module.

The registers return to their reset values when the chip enters PM2 or PM3.

I2CCFG (0x6230) - I²C Control

Bit	Name	Reset	R/W	Description				
7	CR2	0	R/W	Clock rate bit 2				
6	ENS1	0	R/W	Enable bit.				
				<ul> <li>l²C module disabled.</li> <li>SCL and SDA are set to high impedance inputs. The inputs are ignored by the l²C module.</li> <li>Note that setting ENS1 = 0 disables the l²C module but does not reset its state.</li> </ul>				
				1: I ² C module enabled.				
5	STA	0	R/W	START flag. When set, HW detects when I ² C is free and generates a START condition.				
4	STO	0	R/W1	STOP flag. When set and in master mode, a STOP condition is transmitted on the I ² C bus. HW is cleared when transmit has completed successfully.				
3	SI	0	R/W0	Interrupt flag				
2	AA	0	R/W	Assert acknowledge flag for the I ² C module.				
				When set (AA = 1), an acknowledge is returned when:				
				Slave address is recognized				
				• General call is recognized, when the I ² C module is enabled				
				• Data byte received while in master/slave receive mode				
				When not set (AA = 0), an acknowledge is returned when:				
				• Data byte is received while in master/slave receive mode				
1	CR1	0	R/W	Clock rate bit 1				
0	CR0	0	R/W	Clock rate bit 0				

## Table 20-6. Clock Rates Defined at 32 MHz

CR2	CR1	CR0	Bit Frequency (kHz)	Clock Divided by
0	0	0	123	256
0	0	1	144	244
0	1	0	165	192
0	1	1	197	160
1	0	0	33	960
1	0	1	267	120
1	1	0	533	60
1	1	1	Reserved	N/A

## I2CSTAT (0x6231) - I²C Status

Bit	Name	Reset	R/W	Description
7:3	STAC	1111 1	R	Status code. Contains the state of the $l^2C$ core. 27 states are defined: 0 to 25 and 31. Interrupt is only requested when in states 0 to 25. The value 0xF8 indicates that there is no relevant state information available and that I2CCFG.SI = 0.
2:0	-	000	R0	Reserved

#### I2CDATA (0x6232) - I²C Data

Bit	Name	Reset	R/W	Description
7:0	SD	0000 00 00	R/W	Serial data in/out (MSB is bit 7, LSB is bit 0). Contains data byte to be transmitted or byte which has just been received. Can be read or written while not in the process of shifting a byte. The register is not shadowed or double buffered, so it should only be accessed upon an interrupt.



I2CA	2CADDR (0x6233) – I ² C Own Slave Address				
Bit	Name	Reset	R/W	Description	
7:1	ADDR	0000 00 0	R/W	Own slave address	
0	GC	0	R/W	General-call address acknowledge. If set, the general-call address is recognized.	

#### I2CWC (0x6234) – Wrapper control

Bit	Name	Reset	R/W	Description
7	OVR	0	R/W	Override enable:
				0: I ² C functionality (ignore other bits in this register)
				1: GPIO functionality
6:4	-	000	R0	Reserved
3	SCLPUE	1	R/W	SCL pin pullup enable
2	SDAPUE	1	R/W	SDA pin pullup enable
1	SCLOE	0	R/W	SCL pin output enable
0	SDAOE	0	R/W	SDA pin output enable

#### I2CIO (0x6235) - GPIO

Bit	Name	Reset	R/W	Description
7:2	-	000	R0	Reserved
1	SCLD	0	R/W	SCL data value
				When I2CWC.SCLOE is set, reading SCLD reads the output register, not the pin. When I2CWC.SCLOE is cleared, reading SCLD reads the pin. Writing SCLD writes to the output register.
0	SDAD	0	R/W	SDA data value
				When I2CWC.SDAOE is set, reading SDAD reads the output register, not the pin. When I2CWC.SDAOE is cleared, reading SDAD reads the pin. Writing SDAD writes to the output register.



# **USB** Controller

Page

This section focuses on describing the functionality of the USB controller (in the CC2531/CC2540 only), and it is assumed that the reader has a good understanding of USB and is familiar with the terms and concepts used. See the Universal Serial Bus Specification for details ([8], Appendix C).

Standard USB nomenclature is used regarding IN and OUT. I.e., IN is always into the host (PC) and OUT is out of the host.

## Topic

	USB Introduction	
	USB Enable	
21.3	48-MHz USB PLL	192
21.4	USB Interrupts	193
21.5	Endpoint 0	193
21.6	Endpoint-0 Interrupts	193
21.7	Endpoints 1–5	195
	DMA	
	USB Reset	
21.10	Suspend and Resume	199
21.11	Remote Wake-Up	199
	USB Registers	



#### 21.1 USB Introduction

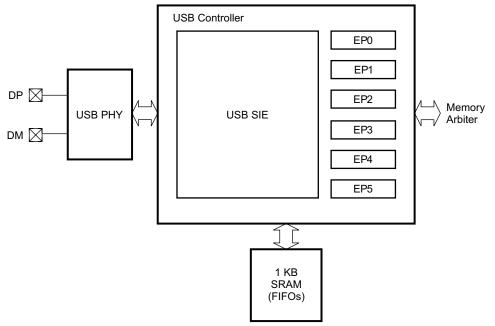
The USB controller monitors the USB for relevant activity and handles packet transfers.

Appropriate response to USB interrupts and loading/unloading of packets into/from endpoint FIFOs is the responsibility of the firmware. The firmware must be able to reply correctly to all standard requests from the USB host and work according to the protocol implemented in the driver on the PC.

The USB controller has the following features:

- Full-speed operation (up to 12 Mbps)
- Five endpoints (in addition to endpoint 0) that can be used as IN, OUT, or IN/OUT and can be configured as bulk/interrupt or isochronous.
- 1-KB SRAM FIFO available for storing USB packets
- Endpoints supporting packet sizes from 8–512 bytes
- Support for double buffering of USB packets

Figure 21-1 shows a block diagram of the USB controller. The USB PHY is the physical interface with input and output drivers. The USB SIE is the serial-interface engine, which controls the packet transfer to/from the endpoints. The USB controller is connected to the rest of the system through the memory arbiter.



B0305-01

Figure 21-1. USB Controller Block Diagram

## 21.2 USB Enable

The USB is enabled by setting <code>USBCTRL.USB_EN</code> to 1. Setting <code>USBCTRL.USB_EN</code> to 0 resets the USB controller.

## 21.3 48-MHz USB PLL

The 48-MHz internal USB PLL must be powered up and stable for the USB controller to operate correctly. It is important that the crystal oscillator is selected as souce and is stable before the USB PLL is enabled. The USB PLL is enabled by setting the USBCTRL.PLL_EN bit and waiting for the USBCTRL.PLL_LOCKED status flag to go high. When the PLL has locked, it is safe to use the USB controller.

Note: The PLL must be disabled before exiting active mode and re-enabled after entering active mode.



## 21.4 USB Interrupts

There are three interrupt flag registers with associated interrupt-enable mask registers.

Interrupt Flag	Description Associated Inter Enable Mask Reg						
USBCIF	Contains flags for common USB interrupts	USBCIE					
USBIIF	Contains interrupt flags for endpoint 0 and all the IN endpoints USBIIE						
USBOIF Contains interrupt flags for all OUT endpoints USBOIE							
Note: All interrupts except SOF and suspend are initially enabled after reset.							

#### Table 21-1. USB Interrupt Flags Interrupt-Enable Mask Registers

The USB controller uses interrupt #6 for USB interrupts. This interrupt number is shared with Port 2 inputs; hence, the interrupt routine must also handle Port 2 interrupts if they are enabled. For an interrupt request to be generated, IEN2.P2IE must be set to 1 together with the desired interrupt enable bits from the USBCIE, USBIIE, and USBOIE registers. When an interrupt request has been generated, the CPU starts executing the ISR if there are no higher-priority interrupts pending. The interrupt routine should read all the interrupt flag registers and take action depending on the status of the flags. The interrupt flag registers are cleared when they are read, and the status of the individual interrupt flags should therefore be saved in memory (typically in a local variable on the stack) to allow them to be accessed multiple times.

At the end of the ISR, after the interrupt flags have been read, the interrupt flags should be cleared to allow for new USB and P2 interrupts to be detected. The Port 2 interrupt status flags in the P2IFG register should be cleared prior to clearing IRCON2.P2IF.

When waking up from suspend (typically in PM1), the USB D+ interrupt flag, P2IFG.DPIF, is set. The D+ interrupt flag indicates that there has been a falling edge on the D+ USB data pin. This is a resume event.

## 21.5 Endpoint 0

Endpoint 0 (EP0) is a bidirectional control endpoint, and during the enumeration phase all communication is performed across this endpoint. Before the USBADDR register has been set to a value other than 0, the USB controller is only able to communicate through endpoint 0. Setting the USBADDR register to a value between 1 and 127 brings the USB function out of the default state in the enumeration phase and into the address state. All configured endpoints are then available for the application.

The EP0 FIFO is only used as either IN or OUT, and double buffering is not provided for endpoint 0. The maximum packet size for endpoint 0 is fixed at 32 bytes.

Endpoint 0 is controlled through the USBCS0 register by setting the USBINDEX register to 0. The USBCNT0 register contains the number of bytes received.

## 21.6 Endpoint-0 Interrupts

The following events may generate an EP0 interrupt request:

- A data packet has been received (USBCS0.OUTPKT_RDY = 1)
- A data packet that was loaded into the EP0 FIFO has been sent to the USB host. (USBCS0.INPKT_RDY should be set to 1 when a new packet is ready to be transferred. This bit is cleared by hardware when the data packet has been sent.)
- An IN transaction has been completed (the interrupt is generated during the status stage of the transaction).
- A STALL has been sent (USBCS0.SENT_STALL = 1)
- A control transfer ends due to a premature end-of-control transfer (USBCS0.SETUP_END = 1)

Any of these events causes USBIIF.EPOIF to be asserted, regardless of the status of the EPO interrupt mask bit USBIIE.EPOIE. If the EPO interrupt mask bit is set to 1, the CPU interrupt flag IRCON2.P2IF is also asserted. An interrupt request is only generated if IEN2.P2IE and USBIIE.EPOIE are both set to 1.



Endpoint-0 Interrupts

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#### 21.6.1 Error Conditions

When a protocol error occurs, the USB controller sends a STALL handshake. The USBCS0.SENT_STALL bit is asserted, and an interrupt request is generated if the endpoint-0 interrupt is properly enabled. A protocol error can be any of the following:

- An OUT token is received after USBCS0.DATA_END has been set to complete the OUT data stage (the host tries to send more data than expected).
- An IN token is received after USBCS0.DATA_END has been set to complete the IN data stage (the host tries to receive more data than expected).
- The USB host tries to send a packet that exceeds the maximum packet size during the OUT data stage.
- The size of the DATA1 packet received during the status stage is not 0.

The firmware can also terminate the current transaction by setting the USBCS0.SEND_STALL bit to 1. The USB controller then sends a STALL handshake in response to the next request from the USB host.

If an EP0 interrupt is caused by the assertion of the USBCS0.SENT_STALL bit, this bit should be de-asserted, and firmware should consider the transfer as aborted (and consequently free the memory buffers, etc.).

If EP0 receives an unexpected token during the data stage, the USBCS0.SETUP_END bit is asserted, and an EP0 interrupt is generated (if enabled properly). EP0 then switches to the IDLE state. Firmware should then set the USBCS0.CLR_SETUP_END bit to 1 and abort the current transfer. If USBCS0.OUTPKT_RDY is asserted, this indicates that another setup packet has been received that firmware should process.

## 21.6.2 SETUP Transactions (IDLE State)

The control transfer consists of two or three stages of transactions (setup – data – status or setup – status). The first transaction is a setup transaction. A successful setup transaction comprises three sequential packets (a token packet, a data packet, and a handshake packet), where the data field (payload) of the data packet is exactly 8 bytes long and is referred to as the setup packet. In the setup stage of a control transfer, EP0 is in the IDLE state. The USB controller rejects the data packet if the setup packet is not 8 bytes. Also, the USB controller examines the contents of the setup packet to determine whether or not there is a data stage in the control transfer. If there is a data stage, EP0 switches state to TX (IN transaction) or RX (OUT transaction) when the USBCS0.CLR_OUTPKT_RDY bit is set to 1 (if USBCS0.DATA_END = 0).

When a packet is received, the USBCS0.OUTPKT_RDY bit is asserted and an interrupt request is generated (EP0 interrupt) if the interrupt has been enabled. Firmware should perform the following when a setup packet has been received:

- 1. Unload the setup packet from the EP0 FIFO
- 2. Examine the contents and perform the appropriate operations
- 3. Set the USBCS0.CLR_OUTPKT_RDY bit to 1. This denotes the end of the setup stage. If the control transfer has no data stage, the USBCS0.DATA_END bit must also be set. If there is no data stage, the USB controller stays in the IDLE state.

## 21.6.3 IN Transactions (TX State)

If the control transfer requires data to be sent to the host, the setup stage is followed by one or more IN transactions in the data stage. In this case, the USB controller is in the TX state and only accepts IN tokens. A successful IN transaction comprises two or three sequential packets (a token packet, a data packet, and a handshake packet ⁽¹⁾). If more than 32 bytes (maximum packet size) is to be sent, the data must be split into a number of 32-byte packets followed by a residual packet. If the number of bytes to send is a multiple of 32, the residual packet is a zero-length data packet, because a packet size less than 32 bytes denotes the end of the transfer.

Firmware should load the EP0 FIFO with the first data packet and set the USBCS0.INPKT_RDY bit as

⁽¹⁾ For isochronous transfers there would not be a handshake packet from the host.

soon as possible after the USBCS0.CLR_OUTPKT_RDY bit has been set. The USBCS0.INPKT_RDY is cleared and an EPO interrupt is generated when the data packet has been sent. Firmware might then load more data packets as necessary. An EPO interrupt is generated for each packet sent. Firmware must set USBCS0.DATA_END in addition to USBCS0.INPKT_RDY when the last data packet has been loaded. This starts the status stage of the control transfer.

EP0 switches to the IDLE state when the status stage has completed. The status stage may fail if the USBCS0.SEND_STALL bit is set to 1. The USBCS0.SENT_STALL bit is then asserted, and an EP0 interrupt is generated.

If USBCS0.INPKT_RDY is not set when receiving an IN token, the USB controller replies with a NAK to indicate that the endpoint is working, but temporarily has no data to send.

## 21.6.4 OUT Transactions (RX State)

If the control transfer requires data to be received from the host, the setup stage is followed by one or more OUT transactions in the data stage. In this case, the USB controller is in the RX state and only accepts OUT tokens. A successful OUT transaction comprises two or three sequential packets (a token packet, a data packet, and a handshake packet ⁽²⁾). If more than 32 bytes (maximum packet size) is to be received, the data must be split into a number of 32-byte packets followed by a residual packet. If the number of bytes to receive is a multiple of 32, the residual packet is a zero-length data packet, because a data packet with payload less than 32 bytes denotes the end of the transfer.

The USBCS0.OUTPKT_RDY bit is set and an EP0 interrupt is generated when a data packet has been received. The firmware should set USBCS0.CLR_OUTPKT_RDY when the data packet has been unloaded from the EP0 FIFO. When the last data packet has been received (packet size less than 32 bytes) firmware should also set the USBCS0.DATA_END bit. This starts the status stage of the control transfer. The size of the data packet is kept in the USBCNT0 registers. Note that this value is only valid when USBCS0.OUTPKT_RDY = 1.

EP0 switches to the IDLE state when the status stage has completed. The status stage may fail if the DATA1 packet received is not a zero-length data packet or if the USBCS0.SEND_STALL bit is set to 1. The USBCS0.SENT_STALL bit then is asserted and an EP0 interrupt is generated.

## 21.7 Endpoints 1–5

Each endpoint can be used as an IN only, an OUT only, or IN/OUT. For an IN/OUT endpoint, there are basically two endpoints, an IN endpoint and an OUT endpoint associated with the endpoint number. Configuration and control of IN endpoints is performed through the USBCSIL and USBCSIH registers. The USBCSOL and USBCSOH registers are used to configure and control OUT endpoints. Each IN and OUT endpoint can be configured as either an isochronous (USBCSIH.ISO = 1 and/or USBCSOH.ISO = 1) or bulk/interrupt (USBCSIH.ISO = 0 and/or USBCSOH.ISO = 0) endpoint. Bulk and interrupt endpoints are handled identically by the USB controller but have different properties from a firmware perspective.

The USBINDEX register must have the value of the endpoint number before the indexed endpoint registers are accessed.

## 21.7.1 FIFO Management

Each endpoint has a certain number of FIFO memory bytes available for incoming and outgoing data packets. Table 21-2 shows the FIFO size for endpoints 1–5. The firmware is responsible for setting the USBMAXI and USBMAXO registers correctly for each endpoint to prevent data from being overwritten.

When both the IN and OUT endpoints of an endpoint number do not use double buffering, the sum of USBMAXI and USBMAXO must not exceed the FIFO size for the endpoint. Figure 21-2 a) shows how the IN and OUT FIFO memory for an endpoint is organized with single buffering. The IN FIFO grows down from the top of the endpoint memory region, whereas the OUT FIFO grows up from the bottom of the endpoint memory region.

⁽²⁾ For isochronous transfers, there is no handshake packet from the device.



When the IN or OUT endpoint of an endpoint number uses double buffering, the sum of USBMAXI and USBMAXO must not exceed half the FIFO size for the endpoint. Figure 21-2 b) illustrates the IN and OUT FIFO memory for an endpoint that uses double buffering. Notice that the second OUT buffer starts from the middle of the memory region and grows upwards. The second IN buffer also starts from the middle of the memory region but grows downwards.

To configure an endpoint as IN-only, set USBMAXO to 0, and to configure an endpoint as OUT-only, set USBMAXI to 0.

For unused endpoints, both USBMAXO and USBMAXI should be set to 0.

	EP Number	FIFO Size (in Byte	s)
	1	32	
	2	64	
	3	128	
	4	256	
	5	512	
0 BMAXI-1	IN FIFO	0 USBMAXI-1	IN FIFO (Buffer 1)
	+	USBMAXO-1	OUT FIFO

Table 21-2, FIFO Sizes for EP 1–5

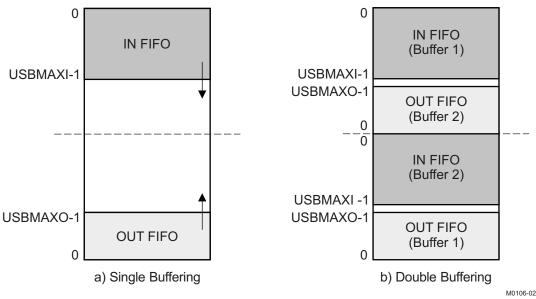


Figure 21-2. IN/OUT FIFOs

## 21.7.2 Double Buffering

To enable faster transfer and reduce the need for retransmissions, double buffering can be used. This allows two packets to be buffered in the FIFO in each direction. This is highly recommended for isochronous endpoints, which are expected to transfer one data packet every USB frame without any retransmission. For an isochronous endpoint, one data packet is sent/received every USB frame. However, the data packet may be sent/received at any time during the USB frame period, and there is a chance that two data packets may be sent/received at a few-microseconds interval. For isochronous endpoints, an incoming packet is lost if there is no buffer available, and a zero-length data packet is sent if there is no data packet ready for transmission when the USB host requests data. Double buffering is not as critical for bulk and interrupt endpoints as it is for isochronous endpoints, because packets are not lost. Double buffering, however, may improve the effective data rate for bulk endpoints.

To enable double buffering for an IN endpoint, USBCSIH.IN_DBL_BUF must be set to 1. To enable double buffering for an OUT endpoint, set USBCSOH.OUT_DBL_BUF to 1.



## 21.7.3 FIFO Access

The endpoint FIFOs are accessed by reading and writing to the registers USBF0–USBF6. Writing to a register causes the byte written to be inserted into the IN FIFO. Reading a register causes the next byte in the OUT FIFO to be extracted and the value of this byte to be returned.

When a data packet has been written to an IN FIFO, the USBCSIL.INPKT_RDY bit must be set to 1. If double buffering is enabled, the USBCSIL.INPKT_RDY bit is cleared immediately after it has been written, and another data packet can be loaded. This does not generate an IN endpoint interrupt, because an interrupt is only generated when a packet has been sent. When double buffering is used, firmware should check the status of the USBCSIL.PKT_PRESENT bit before writing to the IN FIFO. If this bit is 0, two data packets can be written. Double-buffered isochronous endpoints should only load two packets the first time the IN FIFO is loaded. After that, one packet is loaded for every USB frame. To send a zero-length data packet, USBCSIL.INPKT_RDY should be set to 1 without loading a data packet into the IN FIFO.

A data packet can be read from the OUT FIFO when the USBCSOL.OUTPKT_RDY bit is 1. An interrupt is generated when this occurs, if enabled. The size of the data packet is kept in the USBCNTH:USBCNTL registers. Note that this value is only valid when USBCSOL.OUTPKT_RDY = 1. When the data packet has been read from the OUT FIFO, the USBCSOL.OUTPKT_RDY bit must be cleared. If double buffering is enabled, there may be two data packets in the FIFO. If another data packet is ready when the USBCSOL.OUTPKT_RDY bit is cleared, the USBCSOL.OUTPKT_RDY bit is asserted immediately, and an interrupt is generated (if enabled) to signal that a new data packet has been received. The USBCSOL.FIFO_FULL bit is set when there are two data packets in the OUT FIFO.

The AutoClear feature is supported for OUT endpoints. When enabled, the USBCSOL.OUTPKT_RDY bit is cleared automatically when USBMAXO bytes have been read from the OUT FIFO. The AutoClear feature is enabled by setting USBCSOH.AUTOCLEAR = 1. The AutoClear feature can be used to reduce the time the data packet occupies the OUT FIFO buffer and is typically used for bulk endpoints.

A complementary AutoSet feature is supported for IN endpoints. When enabled, the USBCSIL.INPKT_RDY bit is set automatically when USBMAXI bytes have been written to the IN FIFO. The AutoSet feature is enabled by setting USBCSIH.AUTOSET =

1. The AutoSet feature can reduce the overall time it takes to send a data packet and is typically used for bulk endpoints.

## 21.7.4 Endpoint 1–5 Interupts

The following events may generate an IN EPx interrupt request (x indicates the endpoint number):

- A data packet that was loaded into the IN FIFO has been sent to the USB host. (USBCSIL.INPKT_RDY should be set to 1 when a new packet is ready to be transferred. This bit is cleared by hardware when the data packet has been sent.)
- A STALL has been sent (USBCSIL.SENT_STALL
  - = 1). Only bulk/interrupt endpoints can be stalled
- The IN FIFO is flushed due to the USBCSIH.FLUSH_PACKET bit being set to 1.

Any of these events causes USBIIF.INEPxIF to be asserted, regardless of the status of the IN EPx interrupt mask bit USBIIE.INEPxIE. If the IN EPx interrupt mask bit is set to 1, the CPU interrupt flag IRCON2.P2IF is also asserted. An interrupt request is only generated if IEN2.P2IE and USBIIE.INEPxIE are both set to 1. The x in the register name refers to the endpoint number, 1–5)

The following events may generate an OUT EPx interrupt request:

- A data packet has been received (USBCSOL.OUTPKT_RDY = 1).
- A STALL has been sent (USBCSIL.SENT_STALL

= 1). Only bulk/interrupt endpoints can be stalled.

Any of these events causes USBOIF.OUTEPXIF to be asserted, regardless of the status of the OUT EPx interrupt mask bit USBOIE.OUTEPXIE. If the OUT EPx interrupt mask bit is set to 1, the CPU interrupt flag IRCON2.P2IF is also asserted. An interrupt request is only generated if IEN2.P2IE and USBOIE.OUTEPXIE are both set to 1.



#### 21.7.5 Bulk/Interrupt IN Endpoint

Interrupt IN transfers occur at regular intervals, whereas bulk IN transfers use available bandwidth not allocated to isochronous, interrupt, or control transfers.

Interrupt IN endpoints may set the USBCSIH.FORCE_DATA_TOG bit. When this bit is set, the data toggle bit is continuously toggled, regardless of whether an ACK was received or not. This feature is typically used by interrupt IN endpoints that are used to communicate rate feedback for isochronous endpoints.

A bulk/interrupt IN endpoint can be stalled by setting the USBCSIL.SEND_STALL bit to 1. When the endpoint is stalled, the USB controller responds with a STALL handshake to IN tokens. The USBCSIL.SENT_STALL bit is then set, and an interrupt is generated, if enabled.

A bulk transfer longer than the maximum packet size is performed by splitting the transfer into a number of data packets of maximum size followed by a smaller data packet containing the remaining bytes. If the transfer length is a multiple of the maximum packet size, a zero-length data packet is sent last. This means that a packet with a size less than the maximum packet size denotes the end of the transfer. The AutoSet feature can be useful in this case, because many data packets are of maximum size.

#### 21.7.6 Isochronous IN Endpoint

An isochronous IN endpoint is used to transfer periodic data from the USB controller to the host (one data packet every USB frame).

If there is no data packet loaded in the IN FIFO when the USB host requests data, the USB controller sends a zero-length data packet, and the USBCSIL.UNDERRUN bit is asserted.

Double buffering requires that a data packet is loaded into the IN FIFO during the frame preceding the frame where it should be sent. If the first data packet is loaded before an IN token is received, the data packet is sent during the same frame as it was loaded and hence violates the double-buffering strategy. Thus, when double buffering is used, the USBPOW.ISO_WAIT_SOF bit should be set to 1 to avoid this. Setting this bit ensures that a loaded data packet is not sent until the next SOF token has been received.

The AutoSet feature typically is not used for isochronous endpoints, because the packet size increases or decreases from frame to frame.

## 21.7.7 Bulk/Interrupt OUT Endpoint

Interrupt OUT transfers occur at regular intervals, whereas bulk OUT transfers use available bandwidth not allocated to isochronous, interrupt, or control transfers.

A bulk/interrupt OUT endpoint can be stalled by setting the USBCSOL.SEND_STALL bit to 1. When the endpoint is stalled, the USB controller responds with a STALL handshake when the host is done sending the data packet. The data packet is discarded and is not placed in the OUT FIFO. The USB controller asserts the USBCSOL.SENT_STALL bit when the STALL handshake is sent and generates an interrupt request if the OUT endpoint interrupt is enabled.

As the AutoSet feature is useful for bulk IN endpoints, the AutoClear feature is useful for OUT endpoints, because many packets are of maximum size.

#### 21.7.8 Isochronous OUT Endpoint

An isochronous OUT endpoint is used to transfer periodic data from the host to the USB controller (one data packet every USB frame).

If there is no buffer available when a data packet is being received, the USBCSOL.OVERRUN bit is asserted and the packet data is lost. Firmware can reduce the chance for this to happen by using double buffering and using DMA to unload data packets effectively.

An isochronous data packet in the OUT FIFO may have bit errors. The hardware detects this condition and sets USBCSOL.DATA_ERROR. Firmware should therefore always check this bit when unloading a data packet.

The AutoClear feature typically is not used for isochronous endpoints, because the packet size increases or decreases from frame to frame.



#### 21.8 DMA

DMA should be used to fill the IN endpoint FIFOs and empty the OUT endpoint FIFOs. Using DMA improves the read/write performance significantly compared to using the CPU. It is therefore highly recommended to use DMA unless timing is not critical or only a few bytes are to be transferred.

There are no DMA triggers for the USB controller, meaning that DMA transfers must be triggered by firmware.

Byte-size transfer should be used.

## 21.9 USB Reset

When reset signaling is detected on the bus, the USBCIF.RSTIF flag is asserted. If USBCIE.RSTIE is enabled, IRCON2.P2IF is also asserted, and an interrupt request is generated if IEN2.P2IE = 1. The firmware should take appropriate action when a USB reset occurs. A USB reset should place the device in the default state, where it only responds to address 0 (the default address). One or more resets normally take place during the enumeration phase, immediately after the USB cable is connected.

The following actions are performed by the USB controller when a USB reset occurs:

- USBADDR is set to 0.
- USBINDEX is set to 0.
- All endpoint FIFOs are flushed.
- USBMAXI, USBCS0, USBCS1L, USBCS1H, USBMAXO, USBCS0L, USBCS0H, USBCNT0, USBCNTL, and USBCNTH are cleared.
- All interrupts, except SOF and suspend, are enabled.
- An interrupt request is generated (if IEN2.P2IE = 1 and USBCIE.RSTIE = 1).

Firmware should close all pipes and wait for a new enumeration phase when USB reset is detected.

## 21.10 Suspend and Resume

The USB controller asserts USBCIF.SUSPENDIF and enters suspend mode when the USB has been continuously idle for 3 ms, provided that USBPOW.SUSPEND_EN =

1. IRCON2.P2IF is asserted if USBCIE.SUSPENDIE is enabled, and an interrupt request is generated if IEN2.P2IE = 1.

While in suspend mode, only limited current can be sourced from the USB. See the USB 2.0 Specification [3] for details about this. To be able to meet the suspend-current requirement, the device should be taken down to PM1 when suspend is detected. The device should not enter PM2 or PM3, because this resets the USB controller. Before entering PM1, the 48-MHz USB PLL must be turned off. This is done by setting USBCTRL.PLL_EN to 0 and waiting for USBCTRL.PLL_LOCKED to be cleared.

Any valid nonidle signaling on the USB causes USBCIF.RESUMEIF to be asserted and an interrupt request to be generated, and wakes up the system if the USB resume interrupt is enabled.

When the system wakes up (enters active mode) from suspend, no USB registers except USBCTRL can be accessed before the 48-MHz USB PLL has been activated. This is done by setting USBCTRL.PLL_EN to 1 and waiting until USBCTRL.PLL_LOCKED is set.

A USB reset also wakes up the system from suspend. A USB resume interrupt request is generated if the interrupt is enabled, but the USBCIF.RSTIF interrupt flag is set instead of the USBCIF.RESUMEIF interrupt flag.

#### 21.11 Remote Wake-Up

The USB controller can resume from suspend by signaling resume to the USB hub. Resume is performed



#### USB Registers

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by setting USBPOW.RESUME to 1 for approximately 10 ms. According to the USB 2.0 Specification [3], the resume signaling must be present for at least 1 ms and no more than 15 ms. It is, however, recommended to keep the resume signaling for approximately 10 ms. Notice that support for remote wakeup must be declared in the USB descriptor, and that the USB host must grant the device the privilege to perform remote wakeup (through a SET_FEATURE request).

## 21.12 USB Registers

This section describes all USB registers used for control and status for the USB. The USB registers reside in XDATA memory space in the region 0x6200–0x622B. These registers can be divided into three groups: The common USB registers, the indexed endpoint registers, and the endpoint FIFO registers. The indexed endpoint registers represent the currently selected endpoint. The USBINDEX register is used to select the endpoint.

The registers return to their reset values and the FIFOs are cleared when the chip enters PM2 or PM3.

Bit	Name	Reset	R/W	Description
7	UPDATE	0	R	This bit is set when the USBADDR register is written and cleared when the address becomes effective.
6:0	USBADDR[6:0]	000 000 0	R/W	Device address

#### USBADDR (0x6200) - Function Address

Bit	Name	Reset	R/W	Description
7	ISO_WAIT_SOF	0	R/W	When this bit is set to 1, the USB controller sends zero-length data packets from the time INPKT_RDY is asserted and until the first SOF token has been received. This only applies to isochronous endpoints.
6:4	-	000	R0	Reserved
3	RST	0	R	During reset signaling, this bit is set to 1.
2	RESUME	0	R/W	Drives resume signaling for remote wakeup. According to the USB Specification, the duration of driving resume must be at least 1 ms and no more than 15 ms. It is recommended to keep this bit set for approximately 10 ms.
1	SUSPEND	0	R	Suspend mode entered. This bit is only used when SUSPEND_EN = 1. Reading the USBCIF register or asserting RESUME clears this bit.
0	SUSPEND_EN	0	R/W	Suspend enable. When this bit is set to 1, suspend mode is entered when the USB has been idle for 3 ms.

#### USBPOW (0x6201) - Power/Control Register

#### USBIIF (0x6202) - IN Endpoints and EP0 Interrupt Flags

Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5	INEP5IF	0	R, H0	Interrupt flag for IN endpoint 5. Cleared by hardware when read
4	INEP4IF	0	R, H0	Interrupt flag for IN endpoint 4. Cleared by hardware when read
3	INEP3IF	0	R, H0	Interrupt flag for IN endpoint 3. Cleared by hardware when read
2	INEP2IF	0	R, H0	Interrupt flag for IN endpoint 2. Cleared by hardware when read
1	INEP1IF	0	R, H0	Interrupt flag for IN endpoint 1. Cleared by hardware when read
0	EPOIF	0	R, H0	Interrupt flag for endpoint 0. Cleared by hardware when read

#### USBOIF (0x6204) - OUT-Endpoint Interrupt Flags

Bit	Name	Reset	R/W	Description
7:6	-	-	R0	Reserved
5	OUTEP51F	0	R, H0	Interrupt flag for OUT endpoint 5. Cleared by hardware when read
4	OUTEP4IF	0	R, H0	Interrupt flag for OUT endpoint 4. Cleared by hardware when read
3	OUTEP3IF	0	R, H0	Interrupt flag for OUT endpoint 3. Cleared by hardware when read
2	OUTEP21F	0	R, H0	Interrupt flag for OUT endpoint 2. Cleared by hardware when read
1	OUTEP1IF	0	R, H0	Interrupt flag for OUT endpoint 1. Cleared by hardware when read
0	-	-	R0	Reserved

USB	USBCIF (0x6206) – Common USB Interrupt Flags					
Bit	Name	Reset	R/W	Description		
7:4	-	-	R0	Reserved		
3	SOFIF	0	R, H0	Start-of-frame interrupt flag. Cleared by hardware when read		
2	RSTIF	0	R, H0	Reset interrupt flag. Cleared by hardware when read		
1	RESUMEIF	0	R, H0	Resume interrupt flag. Cleared by hardware when read		
0	SUSPENDIF	0	R, H0	Suspend interrupt flag. Cleared by hardware when read		

#### USBIIE (0x6207) – IN Endpoints and EP0 Interrupt-Enable Mask

Bit	Name	Reset	R/W	Description
7:6		00	R/W	Reserved. Always write 00
5	INEP5IE	1	R/W	IN endpoint-5 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
4	INEP4IE	1	R/W	IN endpoint- 4 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
3	INEP3IE	1	R/W	IN endpoint-3 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
2	INEP2IE	1	R/W	IN endpoint-2 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
1	INEP1IE	1	R/W	IN endpoint-1 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
0	EPOIE	1	R/W	Endpoint-0 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled

## USBOIE (0x6209) – Out Endpoints Interrupt Enable Mask

Bit	Name	Reset	R/W	Description
7:6		00	R/W	Reserved. Always write 00
5	OUTEP5IE	1	R/W	OUT endpoint 5 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
4	OUTEP4IE	1	R/W	OUT endpoint 4 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
3	OUTEP3IE	1	R/W	OUT endpoint 3 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
2	OUTEP2IE	1	R/W	OUT endpoint 2 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
1	OUTEP1IE	1	R/W	OUT endpoint 1 interrupt enable
				0: Interrupt disbled
				1: Interrupt enabled
0	_	1	R0	Reserved



#### USB Registers

USE	USBCIE (0x620B) – Common USB Interrupt-Enable Mask					
Bit	Name	Reset	R/W	Description		
7:4		-	R0	Reserved		
3	SOFIE	0	R/W	Start-of-frame interrupt enable		
				0: Interrupt disbled		
				1: Interrupt enabled		
2	RSTIE	1	R/W	Reset interrupt enable		
				0: Interrupt disbled		
				1: Interrupt enabled		
1	RESUMEIE	1	R/W	Resume interrupt enable		
				0: Interrupt disbled		
				1: Interrupt enabled		
0	SUSPENDIE	0	R/W	Suspend interrupt enable		
				0: Interrupt disbled		
				1: Interrupt enabled		

## USBFRML (0x620C) – Current Frame Number (Low Byte)

Bit	Name	Reset	R/W	Description
7:0	FRAME[7:0]	0x00	R	Low byte of 11-bit frame number

#### USBFRMH (0x620D) – Current Frame Number (High Byte)

Bit	Name	Reset	R/W	Description
7:3	-	-	R0	Reserved
2:0	FRAME[10:8]	000	R	3 MSBs of 11-bit frame number

#### USBINDEX (0x620E) - Current-Endpoint Index Register

Bit	Name	Reset	R/W	Description
7:4	-	-	R0	Reserved
3:0	USBINDEX[3:0]	0000	R/W	Endpoint selected. Must be set to a value in the range 0-5

#### USBCTRL (0x620F) – USB Control Register

Bit	Name	Reset	R/W	Description
7	PLL_LOCKED	0	R	PLL locked status
6:3	-	-	R0	Reserved
2	-	0	R/W	Reserved. Always write 0
1	PLL_EN	0	R/W	48-MHz USB PLL enable. When this bit is set, the 48-MHz PLL is started. However, the USB must not be accessed before the PLL has locked, i.e. ,PLL_LOCKED is 1. This bit can only be set when USB_EN = 1.
				Note: The PLL must be disabled before exiting active mode and re-enabled after entering active mode.
0	USB_EN	0	R/W	USB enable. The USB controller is reset when writing 0 to this bit.

#### USBMAXI (0x6210) – Max. Packet Size for IN Endpoint{1–5}

Bit	Name	Reset	R/W	Description
7:0	USBMAXI[7:0]	0x00		Maximum packet size, in units of 8 bytes, for IN endpoint selected by USBINDEX register. The value of this register should correspond to the <i>wMaxPacketSize</i> field in the standard endpoint descriptor for the endpoint. This register must not be set to a value greater than the available FIFO memory for the endpoint.

USB	CS0 (0x6211) – EP0	Control a	nd Status	s (USBINDEX = 0)
Bit	Name	Reset	R/W	Description
7	CLR_SETUP_END	0	R/W H0	Set this bit to 1 to de-assert the SETUP_END bit of this register. This bit is cleared automatically.
6	CLR_OUTPKT_RDY	0	R/W H0	Set this bit to 1 to de-assert the OUTPKT_RDY bit of this register. This bit is cleared automatically.
5	SEND_STALL	0	R/W H0	Set this bit to 1 to terminate the current transaction. The USB controller sends the STALL handshake and this bit is de-asserted.
4	SETUP_END	0	R	This bit is set if the control transfer ends due to a premature end-of-control transfer. The FIFO is flushed and an interrupt request (EP0) is generated if the interrupt is enabled. Setting CLR_SETUP_END = 1 de-asserts this bit.
3	DATA_END	0	R/W H0	This bit is used to signal the end of a data transfer and must be asserted in the following three situations: 1: When the last data packet has been loaded and USECS0.INPKT_RDY is set to 1
				<ol> <li>When the last data packet has been unloaded and USBCS0. INPRI_RD1 is set to 1</li> <li>When the last data packet has been unloaded and USBCS0. CLR_OUTPKT_RD1 is set to 1</li> </ol>
				3: When USBCS0.INPKT_RDY has been asserted without having loaded the FIFO (for sending a zero-length data packet).
				The USB controller clears this bit automatically.
2	SENT_STALL	0	R/W H1	This bit is set when a STALL handshake has been sent. An interrupt request (EP0) is generated if the interrupt is enabled. This bit must be cleared from firmware.
1	INPKT_RDY	0	R/W H0	Set this bit when a data packet has been loaded into the EP0 FIFO to notify the USB controller that a new data packet is ready to be transferred. When the data packet has been sent, this bit is cleared, and an interrupt request (EP0) is generated if the interrupt is enabled.
0	OUTPKT_RDY	0	R	Data packet received. This bit is set when an incoming data packet has been placed in the OUT FIFO. An interrupt request (EP0) is generated if the interrupt is enabled. Set CLR_OUTPKT_RDY = 1 to de-assert this bit.

#### USBCSIL (0x6211) - IN EP{1-5} Control and Status, Low

000						
Bit	Name	Reset	R/W	Description		
7	-	-	R0	Reserved		
6	CLR_DATA_TOG	0	R/W H0	Setting this bit resets the data toggle to 0. Thus, setting this bit forces the next data packet to be a DATA0 packet. This bit is automatically cleared.		
5	SENT_STALL	0	R/W	This bit is set when a STALL handshake has been sent. The FIFO is flushed and the INPKT_RDY bit in this register is de-asserted. An interrupt request (IN EP{1–5}) is generated if the interrupt is enabled. This bit must be cleared from firmware.		
4	SEND_STALL	0	R/W	Set this bit to 1 to make the USB controller reply with a STALL handshake when receiving IN tokens. Firmware must clear this bit to end the STALL condition. It is not possible to stall an isochronous endpoint; thus, this bit only has an effect if the IN endpoint is configured as bulk/interrupt.		
3	FLUSH_PACKET	0	R/W H0	Set to 1 to flush next packet that is ready to transfer from the IN FIFO. The INPKT_RDY bit in this register is cleared. If there are two packets in the IN FIFO due to double buffering, this bit must be set twice to completely flush the IN FIFO. This bit is automatically cleared.		
2	UNDERRUN	0	R/W	In isochronous mode, this bit is set if an IN token is received when INPKT_RDY = 0, and a zero-length data packet is transmitted in response to the IN token. In bulk/interrupt mode, this bit is set when a NAK is returned in response to an IN token. Firmware should clear this bit.		
1	PKT_PRESENT	0	R	This bit is 1 when there is at least one packet in the IN FIFO.		
0	INPKT_RDY	0	R/W H0	Set this bit when a data packet has been loaded into the IN FIFO to notify the USB controller that a new data packet is ready to be transferred. When the data packet has been sent, this bit is cleared, and an interrupt request (IN EP{1–5}) is generated if the interrupt is enabled.		



#### USB Registers

USE	3CSIH (0x6212) – IN E	P{1–5} Co	ntrol an	JSBCSIH (0x6212) – IN EP{1–5} Control and Status, High							
Bit	Name	Reset	R/W	Description							
7	AUTOSET	0	R/W	When this bit is 1, the USBCSIL.INPKT_RDY bit is automatically asserted when a data packet of maximum size (specified by USBMAXI) has been loaded into the IN FIFO.							
6	ISO	0	R/W	Selects IN endpoint type							
				0: Bulk/interrupt							
				1: Isochronous							
5:4		10	R/W	Reserved. Always write 10							
3	FORCE_DATA_TOG	0	R/W	Setting this bit forces the IN endpoint data toggle to switch and the data packet to be flushed from the IN FIFO, even though an ACK was received. This feature can be useful when reporting rate feedback for isochronous endpoints.							
2:1		_	R0	Reserved							
0	IN_DBL_BUF	0	R/W	Double buffering enable (IN FIFO)							
				0: Double buffering disabled							
I				1: Double buffering enabled							

#### USBMAXO (0x6213) – Max. Packet Size for OUT EP{1-5}

Bit	Name	Reset	R/W	Description
7:0	USBMAXO[7:0]	0x00	R/W	Maximum packet size, in units of 8 bytes, for OUT endpoint selected by USBINDEX register. The value of this register should correspond to the wMaxPacketSize field in the standard endpoint descriptor for the endpoint. This register must not be set to a value greater than the available FIFO memory for the endpoint.

#### USBCSOL (0x6214) – OUT EP{1–5} Control and Status, Low

Bit	Name	Reset	R/W	Description
7	CLR_DATA_TOG	0	R/W H0	Setting this bit resets the data toggle to 0. Thus, setting this bit forces the next data packet to be a DATA0 packet. This bit is automatically cleared.
6	SENT_STALL	0	R/W	This bit is set when a STALL handshake has been sent. An interrupt request (OUT EP{1–5}) is generated if the interrupt is enabled. This bit must be cleared from firmware.
5	SEND_STALL	0	R/W	Set this bit to 1 to make the USB controller reply with a STALL handshake when receiving OUT tokens. Firmware must clear this bit to end the STALL condition. It is not possible to stall an isochronous endpoint; thus, this bit only has an effect if the IN endpoint is configured as bulk/interrupt.
4	FLUSH_PACKET	0	R/W H0	Set to 1 to flush the next packet that is to be read from the OUT FIFO. The OUTPKT_RDY bit in this register is cleared. If there are two packets in the OUT FIFO due to double buffering, this bit must be set twice to completely flush the OUT FIFO. This bit is automatically cleared after a write to 1.
3	DATA_ERROR	0	R	This bit is set if there is a CRC or bit-stuff error in the packet received. Cleared when OUTPKT_RDY is cleared. This bit is only valid if the OUT endpoint is isochronous.
2	OVERRUN	0	R/W	This bit is set when an OUT packet cannot be loaded into the OUT FIFO. Firmware should clear this bit. This bit is only valid in isochronous mode.
1	FIFO_FULL	0	R	This bit is asserted when no more packets can be loaded into the OUT FIFO because it is full.
0	OUTPKT_RDY	0	R/W	This bit is set when a packet has been received and is ready to be read from the OUT FIFO. An interrupt request (OUT EP{1–5}) is generated if the interrupt is enabled. This bit should be cleared when the packet has been unloaded from the FIFO.



#### USBCSOH (0x6215) – OUT EP{1–5} Control and Status, High

Bit	Name	Reset	R/W	Description
7	AUTOCLEAR	0	R/W	When this bit is set to 1, the USBCSOL.OUTPKT_RDY bit is automatically cleared when a data packet of maximum size (specified by USBMAXO) has been unloaded to the OUT FIFO.
6	ISO	0	R/W	Selects OUT endpoint type
				0 Bulk/interrupt
				1 Isochronous
5:4		00	R/W	Reserved. Always write 00
3:1		-	R0	Reserved
0	OUT_DBL_BUF	0	R/W	Double buffering enable (OUT FIFO)
				0 Double buffering disabled
				1 Double buffering enabled

#### USBCNT0 (0x6216) – Number of Received Bytes in EP0 FIFO (USBINDEX = 0)

Bit	Name	Reset	R/W	Description
7:6	-	-	R0	Reserved
5:0	USBCNT0[5:0]	00 0000	R	Number of received bytes into EP 0 FIFO. Only valid when OUTPKT_RDY is asserted

#### USBCNTL (0x6216) - Number of Bytes in EP{1-5} OUT FIFO, Low

Bit	Name	Reset	R/W	Description
7:0	USBCNT[7:0]	0x00	R	8 lsbs of number of received bytes in OUT FIFO selected by USBINDEX register. Only valid when USBCSOL.OUTPKT_RDY is asserted.

#### USBCNTH (0x6217) – Number of Bytes in EP{1–5} OUT FIFO, High

Bit	Name	Reset	R/W	Description
7:3	-	-	R0	Reserved
2:0	USBCNT[10:8]	000	R	3 msbs of number of received bytes in OUT FIFO selected by USBINDEX register. Only valid when USBCSOL.OUTPKT_RDY is set

#### USBF0 (0x6220) - Endpoint-0 FIFO

Bit	Name	Reset	R/W	Description
7:0	USBF0[7:0]	0x00	R/W	Endpoint 0 FIFO. Reading this register unloads one byte from the EP0 FIFO. Writing to this register loads one byte into the EP0 FIFO. <i>Note: The FIFO memory for EP0 is used for both incoming and outgoing data packets.</i>

#### USBF1 (0x6222) - Endpoint-1 FIFO

Bit	Name	Reset	R/W	Description
7:0	USBF1[7:0]	0x00		Endpoint 1 FIFO register. Reading this register unloads one byte from the EP1 OUT FIFO. Writing to this register loads one byte into the EP1 IN FIFO.

#### USBF2 (0x6224) - Endpoint-2 FIFO

Bit	Name	Reset	R/W	Description
7:0	USBF2[7:0]	0x00	R/W	Endpoint 2 FIFO register. Reading this register unloads one byte from the EP2 OUT FIFO. Writing to this register loads one byte into the EP2 IN FIFO.

#### USBF3 (0x6226) - Endpoint-3 FIFO

Bit	Name	Reset	R/W	Description
7:0	USBF3[7:0]	0x00	R/W	Endpoint 3 FIFO register. Reading this register unloads one byte from the EP3 OUT FIFO. Writing to this register loads one byte into the EP3 IN FIFO.

#### USBF4 (0x6228) - Endpoint-4 FIFO

Bit	Name	Reset	R/W	Description
7:0	USBF4[7:0]	0x00	R/W	Endpoint 4 FIFO register. Reading this register unloads one byte from the EP4 OUT FIFO. Writing to this register loads one byte into the EP4 IN FIFO.



## USB Registers

USBF	USBF5 (0x622A) – Endpoint-5 FIFO						
Bit	Name	Reset	R/W	Description			
7:0	USBF5[7:0]	0x00	R/W	Endpoint 5 FIFO register. Reading this register unloads one byte from the EP5 OUT FIFO. Writing to this register loads one byte into the EP5 IN FIFO.			



# Timer 2 (MAC Timer)

Timer 2 is mainly used to provide timing for 802.15.4 CSMA-CA algorithms and for general timekeeping in the 802.15.4 MAC layer on CC253x devices, for timekeeping in the BLE link layer on CC2540/41, and for general radio timekeeping when running the radio in proprietary mode on CC2541. Timer 2 must not be used by the application on the CC2540/41 when the BLE stack is running. When Timer 2 is used together with the Sleep Timer, the timing function is provided even when the system enters low-power modes PM1 and PM2. The timer runs at a speed according to the system clock. If Timer 2 is to be used with the Sleep Timer, the system clock source must be the 32-MHz crystal whenever Timer 2 is running, and an external 32-kHz XOSC should be used for accurate results.

The main features of Timer 2 are the following:

- 16-bit timer up-counter providing, for example, a symbol/frame period of 16 μs/320 μs
- · Adjustable period with accuracy of 31.25 ns
- 2 × 16-bit timer compare function
- 24-bit overflow count
- 2 × 24-bit overflow compare function
- Start-of-frame-delimiter capture function
- Timer start/stop synchronous with 32-kHz clock and timekeeping maintained by Sleep Timer.
- · Interrupts generated on compare and overflow
- DMA trigger capability
- · Possible to adjust timer value while counting by introducing delay

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Page

	Timer Operation	
22.2	Interrupts	209
22.3	Event Outputs (DMA Trigger and Radio Events)	210
22.4	Timer Start/Stop Synchronization	210
22.5	Timer 2 Registers	212



#### 22.1 Timer Operation

This section describes the operation of the timer.

#### 22.1.1 General

After a reset, the timer is in the timer IDLE mode, where it is stopped. The timer starts running when T2CTRL.RUN is set to 1. The timer then enters the timer RUN mode. Either the entry is immediate, or it is performed synchronously with the 32-kHz clock. See Section 22.4 for a description of the synchronous start and stop mode.

Once the timer is running in RUN mode, it can be stopped by writing a 0 to T2CTRL.RUN. The timer then enters the timer IDLE mode. The stopping of the timer is performed either immediately or synchronously with the 32-kHz clock.

## 22.1.2 Up Counter

Timer 2 contains a 16-bit timer, which increments on each clock cycle. The counter value can be read from registers T2M1:T2M0 with register T2MSEL.T2MSEL set to 000. Note that the register content in T2M1 is latched when T2M0 is read, meaning that T2M0 must always be read first.

When the timer is idle, the counter can be modified by writing to registers T2M1:T2M0 with register T2MSEL.T2MSEL set to 000. T2M0 must be written first.

## 22.1.3 Timer Overflow

At the same time as the timer counts to a value that is equal to the set timer period, a timer overflow occurs. When the timer overflow occurs, the timer is set to 0x0000. If the overflow interrupt mask bit T2IRQM.TIMER2_PERM is 1, an interrupt request is generated. The interrupt flag bit T2IRQF.TIMER2_PERF is set to 1, regardless of the interrupt mask value.

## 22.1.4 Timer Delta Increment

The timer period may be adjusted once during a timer period by writing a timer delta value. When the timer is running and a timer delta value is written to multiplexed registers T2M1:T2M0 with T2MSEL.T2MSEL set to 000, the 16-bit timer halts at its current value and a delta counter starts counting. The T2M0 register must be written before T2M1. The delta counter starts counting from the delta value written, down to zero. Once the delta counter reaches zero, the 16-bit timer starts counting again.

The delta counter decrements at the same rate as the timer. When the delta counter has reached zero, it does not start counting again until a delta value is written once again. In this way, a timer period may be increased by the delta value in order to make adjustments to the timer overflow events over time.

#### 22.1.5 Timer Compare

A timer compare occurs at the same time as the timer counts to a value that is equal to one of the 16-bit compare values set. When a timer compare occurs, the interrupt flag T2IRQF.TIMER2_COMPARE1F or T2IRQF.TIMER2_COMPARE2F is set to 1, depending of which compare value is reached. An interrupt request is also generated if the corresponding interrupt mask in T2IRQM.TIMER2_COMPARE1M or T2IRQM.TIMER2_COMPARE2M is set to 1.

## 22.1.6 Overflow Count

At each timer overflow, the 24-bit overflow counter is incremented by 1. The overflow counter value is read through registers T2MOVF2:T2MOVF1:T2MOVF0 with register T2MSEL.T2MOVEFSEL set to 000. The registers are latched as in the following description.

If one wants a unique timestamp, where both timer and overflow counter are latched at the same time, do the following: Read T2M0 with T2MSEL.T2MSEL set to 000 and T2CTRL.LATCH_MODE set to 1. This returns the low byte of the timer value, and also latches the high byte of the timer and the entire overflow counter, so the rest of the timestamp is ready to be read.

If one wants to read just the overflow counter without reading timer first, read T2MOVF0 with T2MSEL.T2MOVFSEL set to 000 and T2CTRL.LATCH_MODE set to 0. This returns the low byte of the overflow counter, and latches the two most-significant bytes of the overflow counter so the values are ready to be read.

## 22.1.7 Overflow-Count Update

The overflow count value can be updated by writing to registers T2MOVF2:T2MOVF1:T2MOVF0 with T2MSEL.T2MOVFSEL set to 000. Always write the least-significant byte first, and always write all three bytes. The write takes effect once the high byte is written.

## 22.1.8 Overflow-Count Overflow

At the same time as the overflow counter counts to a value that is equal to the overflow period setting, an overflow period event occurs. When the period event occurs, the overflow counter is set to 0x00 0000. If the overflow interrupt mask bit T2IRQM.TIMER2_OVF_PERM is 1, an interrupt request is generated. The interrupt flag bit T2IRQF.TIMER2_OVF_PERF is set to 1, regardless of the interrupt mask value.

## 22.1.9 Overflow-Count Compare

Two compare values may be set for the overflow counter. The compare values are set by writing to T2MOVF2:T2MOVF1:T2MOVF0 with register T2MSEL.T2MOVFSEL set to 011 or 100. At the same time as the overflow counter counts to a value equal to one of the overflow count compare values, an overflow count compare event occurs. If the corresponding overflow compare interrupt mask bit T2IRQM.TIMER2_OVF_COMPARE1M or T2IRQM.TIMER2_OVF_COMPARE2M is 1, an interrupt request is generated. The interrupt flags bit T2IRQF.TIMER2_OVF_COMPARE1F and T2IRQF.TIMER2_OVF_COMPARE2F are set to 1, regardless of the interrupt mask value.

## 22.1.10 Capture Input

Timer 2 has a timer capture function, which captures the time when the start-of-frame delimiter (SFD) status in the radio goes high.

When the capture event occurs, the current timer value is captured in the capture register. The capture value can be read from registers T2M1:T2M0 if register T2MSEL.T2MSEL is set to 001. The value of the overflow count is also captured at the time of the capture event and can be read from registers T2MOVF2:T2MOVF1:T2MOVF0 if T2MSEL.T2MOVFSEL is set to 001.

## 22.1.11 Long Compare (CC2541 Only)

In the CC2541, two compare values may be set for the combination of the 16-bit timer and the overflow counter. The compare values are a combination of either timer compare 1 and overflow compare 1, or timer compare 2 and overflow compare 2. These combinations are known as the long compare 1 and long compare 2 values, respectively. At the same time as the combination of the 16-bit timer and the 24-bit overflow counter counts to a value equal to one of the long compare values, a long compare event occurs. If the corresponding overflow compare interrupt mask bit T2IRQM.TIMER2_LONG_COMPARE1M or T2IRQM.TIMER2_LONG_COMPARE2M is 1, an interrupt request is generated. The corresponding interrupt flag bit T2IRQF.TIMER2_LONG_COMPARE1F or T2IRQF.TIMER2_LONG_COMPARE2F is set to 1, regardless of the interrupt mask value.

## 22.2 Interrupts

The timer has six (eight on CC2541) individually maskable interrupt sources. These are the following:

- Timer overflow
- Timer compare 1
- Timer compare 2
- Overflow-count overflow
- Overflow-count compare 1
- Overflow-count compare 2



- Long compare 1 (CC2541 only)
- Long compare 2 (CC2541 only)

The interrupt flags are given in the T2IRQF registers. The interrupt flag bits are set only by hardware and can be cleared only by writing to the SFR register.

Each interrupt source can be masked by its corresponding mask bit in the T2IRQM register. An interrupt is generated when the corresponding mask bit is set; otherwise, the interrupt is not generated. The interrupt flag bit is set, however, regardless of the state of the interrupt mask bit.

## 22.3 Event Outputs (DMA Trigger and Radio Events)

Timer 2 has two event outputs, T2_EVENT1 and T2_EVENT2. These can be used as DMA triggers, as inputs to the radio, for conditions in conditional instructions in the CSP on CC253x, for use by the BLE stack on CC2540/41, or for timing TX or RX in CC2541 when running the radio in proprietary mode. The event outputs can be configured individually to any of the following events:

- Timer overflow
- Timer compare 1
- Timer compare 2
- Overflow-count overflow
- Overflow-count compare 1
- Overflow-count compare 2
- Long compare 1 (CC2541 only)
- Long compare 2 (CC2541 only)

The DMA triggers are configured using T2EVTCFG.TIMER2_EVENT1_CFG and T2EVTCFG.TIMER2_EVENT2_CFG.

## 22.4 Timer Start/Stop Synchronization

This section describes the synchronized timer start and stop.

## 22.4.1 General

The timer can be started and stopped synchronously with the 32-kHz clock rising edge. Note that this event is derived from a 32-kHz clock signal, but is synchronous with the 32-MHz system clock and thus has a period approximately equal to that of the 32-kHz clock period. Syncronous starting and stopping must not be attempted unless both the 32-kHz clock and 32-MHz XOSC are running and stable.

At the time of a synchronous start, the timer is reloaded with new calculated values for the timer and overflow count such that it appears that the timer has not been stopped.

#### 22.4.2 Timer Synchronous Stop

After the timer has started running, i.e., entered timer RUN mode, it is stopped synchronously by writing 0 to T2CTRL.RUN when T2CTRL.SYNC is 1. After T2CTRL.RUN has been set to 0, the timer continues running until the 32-kHz clock rising edge is sampled as 1. When this occurs, the timer is stopped, the current Sleep Timer value is stored, and T2CTRL.STATE goes from 1 to 0.

## 22.4.3 Timer Synchronous Start

When the timer is in the IDLE mode, it is started synchronously by writing 1 to T2CTRL.RUN when T2CTRL.SYNC is 1. After T2CTRL.RUN has been set to 1, the timer remains in the IDLE mode until the 32-kHz clock rising edge is detected. When this occurs, the timer first calculates new values for the 16-bit timer value and for the 24-bit timer overflow count, based on the current and stored Sleep Timer values



and the current 16-bit timer values. The new Timer 2 and overflow count values are loaded into the timer, and the timer enters the RUN mode. T2CTRL.STATE = 1 indicates that the module is running. This synchronous start process takes 86 clock cycles from the time when the 32-kHz clock rising edge is sampled high. The synchronous start-and-stop function requires that the system clock frequency is selected to be 32 MHz. If the 16-MHz clock is selected, an offset is added to the new calculated value.

If a synchronous start is done without a previous synchronuous stop, the timer is loaded with unpredictable values. To avoid this, do the first start of the timer asynchronously, then enable synchronous mode for subsequent stops and starts.

The method for calculating the new Timer 2 value and overflow-count value is given as follows. Because the Timer 2 and Sleep Timer clocks are asynchronous with a noninteger clock ratio, there is an error of maximum ±1 in the calculated timer value compared to the ideal timer value, not taking clock inaccuracies into account.



Calculation of New Timer Value and Overflow Count Value
N _c = Current Sleep Timer value
N _{ST} = Stored Sleep Timer value
$K_{ck} = Clock ratio = 976.5625^{(1)}$
stw = Sleep Timer width = 24
P _T = Timer 2 period
P _{OVF} = Overflow period
O _{ST} = Stored overflow-count value
O _{TICK} = Overflow tics while sleeping
t _{ST} = Stored timer value
$T_{OH} = Overhead = 86$
$N_t = N_c - N_{ST}$
$N_t \le 0 \rightarrow N_d = 2^{stw} + N_t; N_t \ge 0 \rightarrow N_d = N_t$
$C = N_d \times K_{ck} + T_{ST} + T_{OH}$ (rounded to nearest integer value)
$T = C \mod P_T$
Timer2Value = T
$O_{TICK} = \frac{(C-T)}{P_{T}}$
$O = (O_{TICK} + O_{ST}) \mod P_{OVF}$
Timer2OverflowCount = O
⁽¹⁾ Clock ratio of Timer 2 clock frequency (32 MHz) and Sleep Timer clock frequency (32 kHz)

For a given Timer 2 period value,  $P_T$ , there is a maximum duration between Timer 2 synchronous stop and start for which the timer value is correctly updated after starting. The maximum value is given in terms of the number of Sleep Timer clock periods, i.e., 32-kHz clock periods,  $t_{ST(max)}$ .

$$t_{ST(max)} \leq \frac{(2^{24} - 1) \times P_T + T_{OH}}{K_{ck}}$$

## 22.5 Timer 2 Registers

The SFR registers associated with Timer 2 are listed in this section. These registers are the following:

- T2MSEL Timer 2 multiplexed register control
- T2M1 Timer 2 multiplexed count high
- T2M0 Timer 2 multiplexed count low
- T2MOVF2 Timer 2 multiplexed overflow count 2
- T2MOVF1 Timer 2 multiplexed overflow count 1
- T2MOVF0 Timer 2 multiplexed overflow count 0
- T2IRQF Timer 2 interrupt flags
- T2IRQM Timer 2 interrupt masks
- T2EVTCFG Timer 2 event output configuration
- T2CTRL Timer 2 configuration

Timer 2 has several multiplexed registers. This is to be able to fit all the registers into the limited SFR address space. The internal registers listed in Table 22-1 can be accessed indirectly through T2M0, T2M1, T2MOVF0, T2MOVF1, and T2MOVF2.

Register Name	Reset	R/W	Function
t2tim[15:0]	0x0000	R/W	Holds the 16-bit upcounter
t2_cap[15:0]	0x0000	R	Holds the last captured value of the upcounter
t2_per[15:0]	0x0000	R/W	Holds the period of the upcounter
t2_cmp1[15:0]	0x0000	R/W	Holds compare value 1 for the upcounter
t2_cmp2[15:0]	0x0000	R/W	Holds compare value 2 for the upcounter
t2ovf[23:0]	0x00 0000	R/W	Holds the 24-bit overflow counter
t2ovf_cap[23:0]	0x00 0000	R	Holds the last captured value of the overflow counter
t2ovf_per[23:0]	0x00 0000	R/W	Holds the period of the overflow counter
t2ovf_cmp1[23:0]	0x00 0000	R/W	Holds compare value 1 for the overflow counter
t2ovf_cmp2[23:0]	0x00 0000	R/W	Holds compare value 2 for the overflow counter

## Table 22-1. Internal Registers

The registers listed in the remainder of this section are directly accessible in the SFR address space.

T2MSEL (0xC3) – Timer 2 Multiplex Select

Bit No.	Name	Reset	R/W	Function
7:0	-	0	R0	Reserved. Read as 0
6:4	T2MOVFSEL	0	R/W	The value of this register selects the internal registers that are modified or read when accessing T2MOVF0, T2MOVF1, and T2MOVF2.
				000: t2ovf (overflow counter)
				001: t2ovf_cap (overflow capture)
				010: t2ovf_per (overflow period)
				011: t2ovf_cmp1 (overflow compare 1)
				100: t2ovf_cmp2 (overflow compare 2)
				101 to 111: Reserved
3	-	0	R0	Reserved. Read as 0
2:0	T2MSEL	0	R/W	The value of this register selects the internal registers that are modified or read when accessing $\tt T2M0$ and $\tt T2M1$ .
				000: <b>t2tim</b> (timer count value)
				001: <b>t2_cap</b> (timer capture)
				010: t2_per (timer period)
				011: t2_cmp1 (timer compare 1)
				100: t2_cmp2 (timer compare 2)
				101 to 111: Reserved

#### T2M0 (0xA2) - Timer 2 Multiplexed Register 0

Bit No.	Name	Reset	R/W	Function
7:0	Т2М0	0	R/W	Indirectly returns/modifies bits [7:0] of an internal register depending on the T2MSEL.T2MSEL value.
				When reading the T2M0 register with T2MSEL.T2MSEL set to 000 and T2CTRL.LATCH_MODE set to 0, the timer (t2tim) value is latched.
				When reading the T2M0 register with T2MSEL.T2MSEL set to 000 and T2CTRL.LATCH_MODE set to 1, the timer ( <b>t2tim</b> ) and overflow counter ( <b>t2ovf</b> ) values are latched.



## Timer 2 Registers

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## T2M1 (0xA3) – Timer 2 Multiplexed Register 1

Bit No.	Name	Reset	R/W	Function
7:0	T2M1	0	R/W	Indirectly returns/modifies bits [15:8] of an internal register, depending on T2MSEL.T2MSEL value.
				When reading the T2M0 register with T2MSEL.T2MSEL set to 000, the timer (t2tim) value is latched.
				Reading this register with T2MSEL.T2MSEL set to 000 returns the latched value of <b>t2tim</b> [15:8].

#### T2MOVF0 (0xA4) - Timer 2 Multiplexed Overflow Register 0

Bit No.	Name	Reset	R/W	Function
7:0	T2MOVF0	0	R/W	Indirectly returns/modifies bits [7:0] of an internal register, depending on the T2MSEL.T2MOVFSEL value.
				When reading the T2MOVF0 register with T2MSEL.T2MOVFSEL set to 000 and T2CTRL.LATCH_MODE set to 0, the overflow counter value ( <b>t2ovf</b> ) is latched.
				When reading the T2M0 register with T2MSEL.T2MOVFSEL set to 000 and T2CTRL.LATCH_MODE set to 1, the overflow counter value ( <b>t2ovf</b> ) is latched.

#### T2MOVF1 (0xA5 – Timer 2 Multiplexed Overflow Register 2

Bit No.	Name	Reset	R/W	Function
7:0	T2MOVF1	0	R/W	Indirectly returns/modifies bits [15:8] of an internal register, depending on the T2MSEL.T2MSEL value.
				Reading this register with T2MSEL.T2MOVFSEL set to 000 returns the latched value of t2ovf[15:8].

#### T2MOVF2 (0xA6) - Timer 2 Multiplexed Overflow Register 2

Bit No.	Name	Reset	R/W	Function
7:0	T2MOVF2	0	R/W	Indirectly returns/modifies bits [23:16] of an internal register, depending on the T2MSEL.T2MOVFSEL value.
				Reading this register with T2MSEL.T2MOVFSEL set to 000 returns the latched value of t2ovf[23:16].

#### T2IRQF (0xA1) – Timer 2 Interrupt Flags (CC253x/CC2540)

Bit No.	Name	Reset	R/W	Function
7:6	-	0	R0	Reserved. Read as 0
5	TIMER2_OVF_COMPARE2F	0	R/W0	Set when the Timer 2 overflow counter counts to the value set at t2ovf_cmp2
4	TIMER2_OVF_COMPARE1F	0	R/W0	Set when the Timer 2 overflow counter counts to the value set at Timer 2 t2ovf_cmp1
3	TIMER2_OVF_PERF	0	R/W0	Set when the Timer 2 overflow counter would have counted to a value equal to <b>t2ovf_per</b> , but instead wraps to 0
2	TIMER2_COMPARE2F	0	R/W0	Set when the Timer 2 counter counts to the value set at t2_cmp2
1	TIMER2_COMPARE1F	0	R/W0	Set when the Timer 2 counter counts to the value set at t2_cmp1
0	TIMER2_PERF	0	R/W0	Set when the Timer 2 counter would have counted to a value equal to <b>t2_per</b> , but instead wraps to 0.

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Bit No.	Name	Reset	R/W	Function
7	TIMER2_LONG_COMPARE2F	0	R/W0	Set when the Timer 2 overflow counter is equal to <b>t2ovf_cmp2</b> and the timer counts to the value set at <b>t2_cmp2</b>
6	TIMER2_LONG_COMPARE1F	0	R/W0	Set when the Timer 2 overflow counter is equal to <b>t2ovf_cmp1</b> and the timer counts to the value set at <b>t2_cmp1</b>
5	TIMER2_OVF_COMPARE2F	0	R/W0	Set when the Timer 2 overflow counter counts to the value set at t2ovf_cmp2
4	TIMER2_OVF_COMPARE1F	0	R/W0	Set when the Timer 2 overflow counter counts to the value set at Timer 2 t2ovf_cmp1
3	TIMER2_OVF_PERF	0	R/W0	Set when the Timer 2 overflow counter would have counted to a value equal to <b>t2ovf_per</b> , but instead wraps to 0
2	TIMER2_COMPARE2F	0	R/W0	Set when the Timer 2 counter counts to the value set at t2_cmp2
1	TIMER2_COMPARE1F	0	R/W0	Set when the Timer 2 counter counts to the value set at t2_cmp1
0	TIMER2_PERF	0	R/W0	Set when the Timer 2 counter would have counted to a value equal to <b>t2_per</b> , but instead wraps to 0.

## T2IRQM (0xA7) – Timer 2 Interrupt Mask (CC253x/CC2540)

Bit No.	Name	Reset	R/W	Function
7:6	-	0	R0	Reserved. Read as 0
5	TIMER2_OVF_COMPARE2M	0	R/W	Enables the TIMER2_OVF_COMPARE2 interrupt
4	TIMER2_OVF_COMPARE1M	0	R/W	Enables the TIMER2_OVF_COMPARE1 interrupt
3	TIMER2_OVF_PERM	0	R/W	Enables the TIMER2_OVF_PER interrupt
2	TIMER2_COMPARE2M	0	R/W	Enables the TIMER2_COMPARE2 interrupt
1	TIMER2_COMPARE1M	0	R/W	Enables the TIMER2_COMPARE1 interrupt
0	TIMER2_PERM	0	R/W	Enables the TIMER2_PER interrupt

## T2IRQM (0xA7) – Timer 2 Interrupt Mask (CC2541 Only)

Bit No.	Name	Reset	R/W	Function
7	TIMER2_LONG_COMARE2M	0	R/W	Enables the TIMER2_LONG_COMPARE2F interrupt
6	TIMER2_LONG_COMARE1M	0	R/W	Enables the TIMER2_LONG_COMPARE1F interrupt
5	TIMER2_OVF_COMPARE2M	0	R/W	Enables the TIMER2_OVF_COMPARE2 interrupt
4	TIMER2_OVF_COMPARE1M	0	R/W	Enables the TIMER2_OVF_COMPARE1 interrupt
3	TIMER2_OVF_PERM	0	R/W	Enables the TIMER2_OVF_PER interrupt
2	TIMER2_COMPARE2M	0	R/W	Enables the TIMER2_COMPARE2 interrupt
1	TIMER2_COMPARE1M	0	R/W	Enables the TIMER2_COMPARE1 interrupt
0	TIMER2_PERM	0	R/W	Enables the TIMER2_PER interrupt



## Timer 2 Registers

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T2CTRL (0x94) -	- Timer 2	Control	Register
-----------------	-----------	---------	----------

Bit No.	Name	Reset	R/W	Function
7:4	_	0	R0	Reserved. Read as 0
3	LATCH_MODE	0	R/W	<ul> <li>Reading T2M0 with T2MSEL.T2MSEL = 000 latches the high byte of the timer, making it ready to be read from T2M1. Reading T2MOVF0 with T2MSEL.T2MOVFSEL</li> <li>= 000 latches the two most-significant bytes of the overflow counter, making it possible to read these from T2MOVF1 and T2MOVF2.</li> </ul>
				1: Reading T2M0 with T2MSEL.T2MSEL = 000 latches the high byte of the timer and the entire overflow counter at once, making it possible to read the values from T2M1, T2MOVF0, T2MOVF1, and T2MOVF2.
2	STATE	0	R	State of Timer 2
				0: Timer idle
				1: Timer running
1	SYNC	1	R/W	0: Starting and stopping of timer is immediate, i.e., synchronous with clk_rf_32m.
				1: Starting and stopping of timer happens at the first positive edge of the 32-kHz clock.
				Read Section 22.4 for more details regarding timer start and stop.
0	RUN	0	R/W	Write 1 to start timer, write 0 to stop timer. When read, it returns the last written value.

## T2EVTCFG (0x9C) – Timer 2 Event Configuration (CC253x/CC2540)

Bit No.	Name	Reset	R/W	Function
7	_	0	R0	Reserved. Read as 0
6:4	TIMER2_EVENT2_CFG	0	R/W	Selects the event that triggers a T2_EVENT2 pulse
				000: t2_per_event
				001: t2_cmp1_event
				010: t2_cmp2_event
				011: t2ovf_per_event
				100: t2ovf_cmp1_event
				101: t2ovf_cmp2_event
				110: Reserved
				111: No event
3	-	0	R0	Reserved. Read as 0
2:0	TIMER2_EVENT1_CFG	0	R/W	Selects the event that triggers a T2_EVENT1 pulse
				000: t2_per_event
				001: t2_cmp1_event
				010: t2_cmp2_event
				011: t2ovf_per_event
				100: t2ovf_cmp1_event
				101: t2ovf_cmp2_event
				110: Reserved
				111: No event



Bit No.	Name	Reset	R/W	Function
7:4	TIMER2_EVENT2_CFG	0	R/W	Selects the event that triggers a T2_EVENT2 pulse
				0000: t2_per_event
				0001: t2_cmp1_event
				0010: t2_cmp2_event
				0011: t2ovf_per_event
				0100: t2ovf_cmp1_event
				0101: t2ovf_cmp2_event
				0110: Reserved
				0111: No event
				1000: t2ovf_long_cmp1_event
				1001: t2ovf_long_cmp2_event
				1010–1110: Reserved
				1111: No event
3:0	TIMER2_EVENT1_CFG	0	R/W	Selects the event that triggers a T2_EVENT1 pulse
				0000: t2_per_event
				0001: t2_cmp1_event
				0010: t2_cmp2_event
				0011: t2ovf_per_event
				0100: t2ovf_cmp1_event
				0101: t2ovf_cmp2_event
				0110: Reserved
				0111: No event
				1000: t2ovf_long_cmp1_event
				1001: t2ovf_long_cmp2_event
				1010–1110: Reserved
				1111: No event



# CC253x Radio

The **RF Core** controls the analog radio modules. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events.

Торіс		Page
23.1	RF Core	220
23.2	FIFO Access	224
23.3	DMA	224
23.4	Метогу Мар	224
23.5	Frequency and Channel Programming	226
23.6	IEEE 802.15.4-2006 Modulation Format	226
23.7	IEEE 802.15.4-2006 Frame Format	228
23.8	Transmit Mode	229
23.9	Receive Mode	233
23.10	RXFIFO Access	243
23.11	Radio Control State Machine	245
23.12	Random-Number Generation	247
23.13	Packet Sniffing and Radio Test Output Signals	248
23.14	Command Strobe/CSMA-CA Processor	249
23.15	Registers	266



#### _____

RF Core

# 23.1 RF Core

The **RF Core** controls the analog radio modules. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events.

The **FSM** submodule controls the RF transceiver state, the transmitter and receiver FIFOs, and most of the dynamically controlled analog signals, such as power up/down of analog modules. The FSM is used to provide the correct sequencing of events (such as performing an FS calibration before enabling the receiver/transmitter). Also, it provides step-by-step processing of incoming frames from the demodulator: reading the frame length, counting the number of bytes received, checking the FCS, and finally, optionally handling automatic transmission of ACK frames after successful frame reception. It performs similar tasks in TX, including performing an optional CCA before transmission and automatically going to RX after the end of transmission to receive an ACK frame. Finally, the FSM controls the transfer of data between modulator/demodulator and the TXFIFO/RXFIFO in RAM.

The **modulator** transforms raw data into I/Q signals to the transmitter DAC. This is done in compliance with the IEEE 802.15.4 standard.

The **demodulator** is responsible for retrieving the over-the-air data from the received signal.

The amplitude information from the demodulator is used by the **automatic gain control** (AGC). The AGC adjusts the gain of the analog LNA so that the signal level within the receiver is approximately constant.

The **frame filtering and source matching** supports the FSM in the RF Core by performing all operations needed in order to do frame filtering and source address matching, as defined by IEEE 802.15.4.

The frequency synthesizer (FS) generates the carrier wave for the RF signal.

The **command strobe processor (CSP)** processes all commands issued by the CPU. It also has a short program memory of 24 bytes, making it possible to automate CSMA-CA algorithms.

The **radio RAM** holds a FIFO for transmit data (TXFIFO) and a FIFO for receive data (RXFIFO). Both FIFOs are 128 bytes long. In addition, the RAM holds parameters for frame filtering and source matching, and for which 128 bytes are reserved.

**Timer 2 (MAC Timer)** is used for timing of radio events and to capture time stamps of incoming packets. This timer keeps counting even in power modes PM1 and PM2.

### 23.1.1 Interrupts

The radio is associated with two **interrupt** vectors on the CPU. These are the RFERR interrupt (interrupt 0) and the RF interrupt (interrupt 12) with the following functions.

- RFERR: Error situations in the radio are signaled using this interrupt.
- RF: Interrupts coming from normal operation are signaled using this interrupt.

The RF interrupt vector combines the interrupts in RFIF. Note that these RF interrupts are rising-edge triggered. Thus, an interrupt is generated when, e.g., the SFD status flag goes from 0 to 1. The RFIF interrupt flags are described in Section 23.1.2.

### 23.1.2 Interrupt Registers

Two of the main interrupt control SFR registers are used to enable the RF and RFERR interrupts. These are the following:

- **RFERR**: IEN0.RFERRIE
- **RF**: IEN2.RFIE

Two main interrupt flag SFR registers hold the RF and RFERR interrupt flags. These are the following:

- RFERR: TCON.RFERRIF
- RF: S1CON.RFIF

The two interrupts generated by the RF Core are a combination of several sources within the RF Core. Each of the individual sources has its own enable and interrupt flags in the RF Core. Flags can be found in RFIRQF0, RFIRQF1, and RFIERRF. Interrupt masks can be found in RFIRQM0, RFIRQM1, and RFIERRF.

The interrupt-enable bits in the mask registers are used to enable individual interrupt sources for the two RF interrupts. Note that masking an interrupt source does not affect the updating of the status in the flag registers.

Due to the use of individual interrupt masks in the RF Core, the interrupts coming from RF Core have two-layered masking, and care must be taken when processing these interrupts. The procedure is described as follows.

To clear an interrupt from the RF Core, one must clear two flags, both the flag set in RF Core and the one set in S1CON or TCON (depending on which interrupt is triggered). If a flag is cleared in the RF Core and there are other unmasked flags standing, another interrupt is generated.

Bit	Name	Reset	R/W	Description
7	RXMASKZERO	0	R/W0	The RXENABLE register has gone from a nonzero state to an all-zero state.
				0: No interrupt pending
				1: Interrupt pending
6	RXPKTDONE	0	R/W0	A complete frame has been received.
				0: No interrupt pending
				1: Interrupt pending
5	FRAME_ACCEPTED	0	R/W0	Frame has passed frame filtering.
				0: No interrupt pending
				1: Interrupt pending
4	SRC_MATCH_FOUND	0	R/W0	Source match found
				0: No interrupt pending
				1: Interrupt pending
3	SRC_MATCH_DONE	0	R/W0	Source matching complete
				0: No interrupt pending
				1: Interrupt pending
2	FIFOP	0	R/W0	The number of bytes in the RXFIFO is above the threshold. Also raised when a complete frame has been received, and when a packet has been read out completely and there are more complete packets available.
				0: No interrupt pending
				1: Interrupt pending
1	SFD	0	R/W0	SFD has been received or transmitted.
				0: No interrupt pending
				1: Interrupt pending
0	ACT_UNUSED	0	R/W0	Reserved
				0: No interrupt pending
				1: Interrupt pending

#### RFIRQF0 (0xE9) RF Interrupt Flags

# RF Core

### RFIRQF1 (0x91) RF Interrupt Flags

Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved. Read as 0
5	CSP_WAIT	0	R/W0	Execution continued after a wait instruction in CSP.
				0: No interrupt pending
				1: Interrupt pending
4	CSP_STOP	0	R/W0	CSP has stopped program execution.
				0: No interrupt pending
				1: Interrupt pending
3	CSP_MANINT	0	R/W0	Manual interrupt generated from CSP.
				0: No interrupt pending
				1: Interrupt pending
2	RFIDLE	0	R/W0	Radio state machine has entered the idle state.
				0: No interrupt pending
				1: Interrupt pending
1	TXDONE	0	R/W0	A complete frame has been transmitted.
				0: No interrupt pending
				1: Interrupt pending
0	TXACKDONE	0	R/W0	An acknowledgement frame has been completely transmitted.
				0: No interrupt pending
				1: Interrupt pending

### RFERRF (0xBF) – RF Error Interrupt Flags

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	STROBEERR	0	R/W0	A command strobe was issued at a time it could not be processed. Triggered if trying to disable radio when already disabled, or when trying to do a SACK, SACKPEND, or SNACK command when not in active RX.
				0: No interrupt pending
				1: Interrupt pending
5	TXUNDERF	0	R/W0	TXFIFO underflowed
				0: No interrupt pending
				1: Interrupt pending
4	TXOVERF	0	R/W0	TXFIFO overflowed
				0: No interrupt pending
				1: Interrupt pending
3	RXUNDERF	0	R/W0	RXFIFO underflowed
				0: No interrupt pending
				1: Interrupt pending
2	RXOVERF	0	R/W0	RXFIFO overflowed
				0: No interrupt pending
				1: Interrupt pending
1	RXABO	0	R/W0	Reception of a frame was aborted.
				0: No interrupt pending
				1: Interrupt pending
0	NLOCK	0	R/W0	Frequency synthesizer failed to achieve lock after timeout, or lock is lost during reception. Receiver must be restarted to clear this error situation.
				0: No interrupt pending
				1: Interrupt pending



Bit	Name	Reset	R/W	Description
7	RXMASKZERO	0	R/W	The RXENABLE register has gone from a nonzero state to an all-zero state.
				0: Interrupt disabled
				1: Interrupt enabled
6	RXPKTDONE	0	R/W	A complete frame has been received.
				0: Interrupt disabled
				1: Interrupt enabled
5	FRAME_ACCEPTED	0	R/W	Frame has passed frame filtering.
				0: Interrupt disabled
				1: Interrupt enabled
4	SRC_MATCH_FOUND	0	R/W	Source match found
				0: Interrupt disabled
				1: Interrupt enabled
3	SRC_MATCH_DONE	0	R/W	Source matching complete
				0: Interrupt disabled
				1: Interrupt enabled
2	FIFOP	0	R/W	The number of bytes in the RXFIFO is above the threshold. Also raised when a complete frame has been received, or when a complete packet has been read out and there are more complete packets available.
				0: Interrupt disabled
				1: Interrupt enabled
1	SFD	0	R/W	SFD has been received or transmitted.
				0: Interrupt disabled
				1: Interrupt enabled
0	ACT_UNUSED	0	R/W	Reserved
				0: Interrupt disabled
				1: Interrupt enabled

#### RFIRQM1 (0x61A4 ) – RF Interrupt Masks

Bit	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved. Read as 0
5	CSP_WAIT	0	R/W	Execution continued after a wait instruction in CSP.
				0: Interrupt disabled
				1: Interrupt enabled
4	CSP_STOP	0	R/W	CSP has stopped program execution.
				0: Interrupt disabled
				1: Interrupt enabled
3	CSP_MANINT	0	R/W	Manual interrupt generated from CSP.
				0: Interrupt disabled
				1: Interrupt enabled
2	RFIDLE	0	R/W	Radio state machine has entered the idle state.
				0: Interrupt disabled
				1: Interrupt enabled
1	TXDONE	0	R/W	A complete frame has been transmitted.
				0: Interrupt disabled
				1: Interrupt enabled
0	TXACKDONE	0	R/W	An acknowledgement frame has been completely transmitted.
				0: Interrupt disabled
				1: Interrupt enabled

#### FIFO Access

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Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
7:6	STROBEERR	0	R/W	A command strobe was issued at a time it could not be processed. Triggered if trying to disable radio when already disabled, or when trying to do a SACK, SACKPEND, or SNACK command when not in active RX.
				0: Interrupt disabled
				1: Interrupt enabled
5	TXUNDERF	0	R/W	TXFIFO underflowed
				0: Interrupt disabled
				1: Interrupt enabled
4	TXOVERF	0	R/W	TXFIFO overflowed
				0: Interrupt disabled
				1: Interrupt enabled
3	RXUNDERF	0	R/W	RXFIFO underflowed
				0: Interrupt disabled
				1: Interrupt enabled
2	RXOVERF	0	R/W	RXFIFO overflowed
				0: Interrupt disabled
				1: Interrupt enabled
1	RXABO	0	R/W	Reception of a frame was aborted.
				0: Interrupt disabled
				1: Interrupt enabled
0	NLOCK	0	R/W	Frequency synthesizer failed to achieve lock after timeout, or lock is lost during reception. Receiver must be restarted to clear this error situation.
				0: Interrupt disabled
				1: Interrupt enabled

# 23.2 FIFO Access

The TXFIFO and RXFIFO may be accessed though the SFR register RFD (0xD9). Data is written to the TXFIFO when writing to the RFD register. Data is read from the RXFIFO when the RFD register is read.

The XREG registers RXFIFOCNT and TXFIFOCNT provide information on the amount of data in the FIFOs. The FIFO contents can be cleared by issuing SFLUSHRX and SFLUSHTX.

#### RFD (0xD9) – RF Data

Bit	Name	Reset	R/W	Description
7:0	RFD[7:0]	0x00		Data written to the register is written to the TXFIFO. When reading this register, data from the RXFIFO is read.

# 23.3 DMA

It is possible to use direct memory access (DMA) to move data between memory and the radio. The DMA controller is described in Chapter 8. See this section for a detailed description on how to set up and use DMA transfers.

To support the DMA controller, there is one DMA trigger associated with the radio, the RADIO DMA trigger (DMA trigger 19). The RADIO DMA trigger is activated by two events. The first event to cause a RADIO DMA trigger is when the first data is present in the RXFIFO, i.e. when the RXFIFO goes from the empty state to a nonempty state. The second event that causes a RADIO DMA trigger is when data is read from the RXFIFO (through RFD) and there is still more data available in the RXFIFO.

# 23.4 Memory Map

The RF Core contains 384 bytes of physical RAM located at addresses 0x6000 to 0x0617F. The configuration and status registers of the RF Core are located at addresses from 0x6180 to 0x61EF. Configuration registers, RXFIFO, and TXFIFO are all preserved during sleep modes.



# 23.4.1 RXFIFO

The RXFIFO memory area is located at addresses 0x6000 to 0x607F and is thus 128 bytes. Although this memory area is intended for the RXFIFO, it is not protected in any way, so it is still accessible in the XREG memory space. Normally, only the designated instructions should be used to manipulate the contents of the RXFIFO. The RXFIFO can contain more than one frame at a time.

# 23.4.2 TXFIFO

The TXFIFO memory area is located at addresses 0x6080 to 0x60FF and is thus 128 bytes. Although this memory area is intended for the TXFIFO, it is not protected in any way, so it is still accessible in the XREG memory space. Normally, only the designated instructions should be used to manipulate the contents of the TXFIFO. The TXFIFO can only contain one frame at a time.

# 23.4.3 Frame-Filtering and Source-Matching Memory Map

The frame-filtering and source-address-matching functions use a 128 byte block of the RF Core RAM to store local-address information and source-matching configuration and results; this is located in the area 0x6100 to 0x617F. This memory space is described in Table 23-1. Values that do not fill an entire byte/word are in the least-significant part of the byte/word. Note that the values in these registers are unknown after reset. However, the values are retained during power modes.

ADDRESS	<b>REGISTER/VARIABLE</b>	ENDIAN	DESCRIPTION					
	RESERVED							
0x6176–0x617F	Temporary storage		Memory space used for temporary storage of variables					
			RESS INFORMATION					
0x6174–0x6175	SHORT_ADDR	LE	The short address used during destination address filtering					
0x6172–0x6173	PAN_ID	LE	The PAN ID used during destination address filtering					
0x616A–0x6171	EXT_ADD	LE	The IEEE extended address used during destination address filtering					
	SO		SS MATCHING CONTROL					
0x6169	SRCSHORTPENDEN2		8 MSBs of the 24-bit mask that enables/disables automatic pending for each of the 24 short addresses					
0x6168	SRCSHORTPENDEN1							
0x6167	SRCSHORTPENDEN0		8 LSBs of the 24-bit mask that enables/disables automatic pending for each of the 24 short addresses					
0x6166	SRCEXTPENDEN2		8 MSBs of the 24-bit mask that enables/disables automatic pending for each of the 12 extended addresses. Entry n is mapped to SRCEXTPENDEN[2n]. All SRCEXTPENDEN[2n + 1] bits are don't care.					
0x6165	SRCEXTPENDEN1							
0x6164	SRCEXTPENDEN0		8 LSBs of the 24-bit mask that enables/disables automatic pending for each of the 12 extended addresses. Entry n is mapped to SRCEXTPENDEN[2n]. All SRCEXTPENDEN[2n + 1] bits are don't care.					
	sc	URCE ADDR	ESS MATCHING RESULT					
0x6163	SRCRESINDEX		The bit index of the least-significant 1 in SRCRESMASK, or 0x3F when there is no source match. On a match, bit 5 is 0 when the match is on a short address and 1 when it is on an extended address. On a match, bit 6 is 1 when the conditions for automatic pending bit in acknowledgment have been met (see the description of SRCMATCH.AUTOPEND). The bit gives no indication of whether or not the acknowledgment actually is transmitted, and does not take the PENDING_OR register bit and the SACK/SACKPEND/SNACK strobes into account.					
0x6162	SRCRESMASK2		24-bit mask that indicates source address match for each individual entry in the source address table					
0x6161	SRCRESMASK1		Short address matching. When there is a match on entry panid_n + short_n, bit n is set in SRCRESMASK.					

ADDRESS	REGISTER/VARIABLE		REGISTER/VARIABLE ENDIAN		IAN	DESCRIPTION		
0x6160	SRCRESMA	SK0			Extended address matching. When there is a match on entry ext_n, bits 2n and 2n + 1 are set in SRCRESMASK.			
			SO	URCE	ADDRESS TABLE			
0x615E-0x615F	short_23		LE					
0x615C-0x615D	panid_23	ovt 11	LE	IF	Two individual short-address entries (combination of 16-bit PAN ID			
0x615A-0x615B	short_22	ext_11	LE		and 16-bit short address) or one extended address entry			
0x6158–0x6159	panid_22		LE					
0x610E-0x610F	short_03		LE					
0x610C-0x610D	panid_03	out 01	LE	- LE	Two individual short address entries (combination of 16-bit PAN ID			
0x610A-0x610B	short_02	ext_01	LE		and 16-bit short address) or one extended address entry			
0x6108–0x6109	panid_02		LE					
0x6106–0x6107	short_01		LE					
0x6104–0x6105	panid_01	aut 00	LE	. –	Two individual short address entries (combination of 16-bit PAN ID			
0x6102–0x6103	short_00	ext_00	LE	LE	and 16-bit short address) or one extended address entry			
0x6100–0x6101	panid_00	1	LE					

#### Table 23-1. Frame Filtering and Source Matching Memory Map (continued)

# 23.5 Frequency and Channel Programming

The carrier frequency is set by programming the 7-bit frequency word located in FREQCTRL.FREQ[6:0]. Changes take effect after the next recalibration. Carrier frequencies in the range from 2394 MHz to 2507 MHz are supported. The carrier frequency  $f_c$ , in MHz, is given by  $f_c = (2394 + FREQCTRL.FREQ[6:0])$  MHz, and is programmable in 1-MHz steps.

IEEE 802.15.4-2006 specifies 16 channels within the 2.4-GHz band. They are numbered 11 through 26 and are 5 MHz apart. The RF frequency of channel k is given by Equation 4.

 $f_c = 2405 + 5(k - 11) \text{ [MHz] } k \in [11, 26]$ 

(4)

For operation in channel k, the FREQCTRL.FREQ register should therefore be set to FREQCTRL.FREQ = 11 + 5 (k - 11).

# 23.6 IEEE 802.15.4-2006 Modulation Format

This section is meant as an introduction to the 2.4-GHz direct-sequence spread-spectrum (DSSS) RF modulation format defined in IEEE 802.15.4-2006. For a complete description, see the standard document [1].

The modulation and spreading functions are illustrated at the block level in Figure 23-1. Each byte is divided into two symbols, 4 bits each. The least-significant symbol is transmitted first. For multibyte fields, the least-significant byte is transmitted first, except for security-related fields, where the most-significant byte is transmitted first.

Each symbol is mapped to one out of 16 pseudorandom sequences, 32 chips each. The symbol-to-chip mapping is shown in Table 23-2. The chip sequence is then transmitted at 2 Mchips/s, with the least-significant chip ( $C_0$ ) transmitted first for each symbol. The transmitted bit stream and the chip sequences are observable on GPIO pins P1[0:5]. See Chapter 7 for details on how to configure the GPIO to do this.



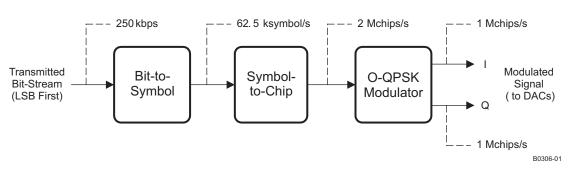
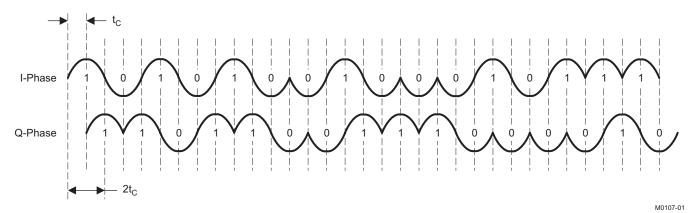
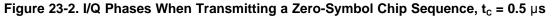


Figure 23-1. Modulation

Symbol	Chip Sequence (C0, C1, C2, , C31)
0	11011001110000110101001000101110
1	111011011001110000110101000010
2	00101110110110011100001101010010
3	00100010111011011001110000110101
4	01010010001011101101100111000011
5	00110101001000101110110110011100
6	11000011010100100010111011011001
7	10011100001101010001001011101101
8	10001100100101100000011101111011
9	10111000110010010110000001110111
10	01111011100011001001011000000111
11	0111011110111000110010010100000
12	00000111011110111000110010010110
13	0110000001110111011100011001001
14	10010110000001110111101110001100
15	1100100101100000011101110111000

The modulation format is offset – quadrature phase shift keying (O-QPSK) with half-sine chip shaping. This is equivalent to MSK modulation. Each chip is shaped as a half-sine, transmitted alternately in the I and Q channels with one-half chip-period offset. This is illustrated for the zero-symbol in Figure 23-2.



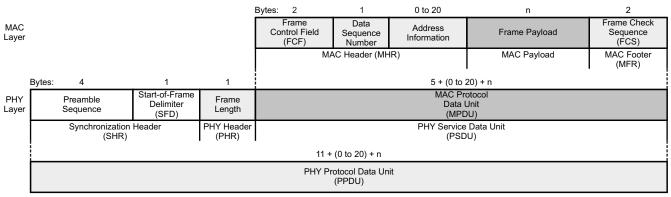




### 23.7 IEEE 802.15.4-2006 Frame Format

This section gives a brief summary of the IEEE 802.15.4 frame format [1]. The radio has built-in support for processing of parts of the frame. This is described in the following sections.

Figure 23-3 shows a schematic view of the IEEE 802.15.4 frame format. Similar figures describing specific frame formats (data frames, beacon frames, acknowledgment frames, and MAC command frames) are included in the standard document [1].



M0108-01

### Figure 23-3. Schematic View of the IEEE 802.15.4 Frame Format [1]

### 23.7.1 PHY Layer

#### Synchronization Header

The synchronization header (SHR) consists of the preamble sequence followed by the start-of-frame delimiter (SFD). In the IEEE 802.15.4 specification [1], the preamble sequence is defined to be 4 bytes of 0x00. The SFD is one byte with value 0xA7.

#### **PHY Header**

The PHY header consists only of the frame-length field. The frame-length field defines the number of bytes in the MPDU. Note that the value of the frame-length field does not include the frame-length field itself. It does, however, include the frame-check sequence (FCS), even if this is inserted automatically by the hardware.

The frame-length field is 7 bits long and has a maximum value of 127. The most-significant bit in the frame-length field is reserved, and should always be set to zero.

#### PHY Service Data Unit

The PHY service data unit (PSDU) contains the MAC protocol data unit (MPDU). It is the MAC layer's function to generate/interpret the MPDU, and the radio has built-in support for processing of some of the MPDU subfields.

### 23.7.2 MAC Layer

The FCF, data sequence number, and address information follow the frame-length field as shown in Figure 23-3. Together with the MAC data payload and frame check sequence, they form the MPDU. The format of the FCF is shown in Figure 23-4. For full details, see the IEEE 802.15.4 specification [1].

Bits: 0–2	3	4	5	6	7–9	10–11	12–13	14–15
Frame type	Security enabled	Frame pending	Acknowledge request	Intra PAN	Reserved	Destination addressing mode	Reserved	Source addressing mode

### Frame-Check Sequence

A 2-byte frame-check sequence (FCS) follows the last MAC payload byte as shown in Figure 23-3. The FCS is calculated over the MPDU, i.e., the frame-length field is not part of the FCS.

The FCS polynomial defined in [1] is

 $G(s) = x^{16} + x^{12} + x^5 + 1$ 

The radio supports automatic calculation/verification of the FCS. See Section 23.8.10 for details.

# 23.8 Transmit Mode

This section describes how to control the transmitter, how to control the integrated frame processing, and how to use the TXFIFO.

# 23.8.1 TX Control

The radio has many built-in features for frame processing and status reporting. Note that the radio provides features that make it easy to have precise control of the timing of outgoing frames. This is very important in an IEEE 802.15.4/ ZigBee® system, because there are strict timing requirements to such systems.

Frame transmission is started by the following actions:

- The STXON command strobe
  - The SAMPLED_CCA signal is not updated.
- The STXONCCA command strobe, provided that the CCA signal is high.
  - Aborts ongoing transmission/reception and forces a TX calibration followed by transmission.
  - The SAMPLED_CCA signal is updated.

Clear channel assessment is described in detail in Section 23.8.12.

Frame transmission is aborted by the following command actions:

- The SRXON command strobe
  - Aborts ongoing transmission and forces an RX calibration
- The SRFOFF command strobe
  - Aborts ongoing transmission/reception and forces the FSM to the IDLE state.
- The STXON command strobe
  - Aborts ongoing transmission and forces an RX calibration

To enable the receiver after transmission with STXON, the FRMCTRL1.SET_RXENMASK_ON_TX bit should be set. This sets bit 6 in RXENABLE when STXON is executed. When transmitting with STXONCCA, the receiver is on before the transmission and is turned back on afterwards (unless the RXENABLE registers have been cleared in the meantime).

# 23.8.2 TX State Timing

Transmission of preamble begins 192 µs after the STXON or STXONCCA command strobe. This is referred to as *TX turnaround time* in [1]. There is an equal delay when returning to receive mode.

When returning to idle or receive mode, there is a 2-µs delay while the modulator ramps down the signals to the DACs. The down ramping happens automatically after the complete MPDU (as defined by the length byte) has been transmitted or if TX underflow occurs. This affects:

- The SFD signal, which is stretched by 2 μs.
- The radio FSM transition to the IDLE state, which is delayed by 2 µs.

# 23.8.3 TXFIFO Access

The TXFIFO can hold 128 bytes and only one frame at a time. The frame can be buffered before or after the TX command strobe is executed, as long as it does not generate TX underflow (see the error conditions listed in Section 23.8.5).



Transmit Mode

Figure 23-5 illustrates what must be written to the TXFIFO (marked blue). Additional bytes are ignored, unless TX overflow occurs (see the error conditions listed in Section 23.8.5).

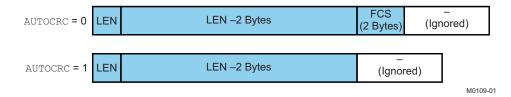


Figure 23-5. Frame Data Written to the TXFIFO

There are two ways to write to the TXFIFO.

- Write to the RFD register.
- Frame buffering always begins at the start of the TXFIFO memory. By enabling the
  FRMCTRL1.IGNORE_TX_UNDERF bit, it is possible to write directly into the RAM area in the radio
  memory, which holds the TXFIFO. Note that it is recommended to use the RFD register for writing data
  to the TXFIFO.

The number of bytes in the TXFIFO is stored in the TXFIFOCNT register.

The TXFIFO can be emptied manually with the SFLUSHTX command strobe. TX underflow occurs if the FIFO is emptied during transmission.

### 23.8.4 Retransmission

In order to support simple retransmission of frames, the radio does not delete the TXFIFO contents as they are transmitted. After a frame has been successfully transmitted, the FIFO contents are left unchanged. To retransmit the same frame, simply restart TX by issuing an STXON or STXONCCA command strobe. Note that a retransmission of a packet is only possible if the packet has been completely transmitted; i.e., a packet cannot be aborted and then be retransmitted.

If a different frame is to be transmitted, issue an ISFLUSHTX strobe and then write the new frame to the TXFIFO.

# 23.8.5 Error Conditions

There are two error conditions associated with the TXFIFO:

- Overflow happens when the TXFIFO is full and another byte write is attempted.
- Underflow happens when the TXFIFO is empty and the radio attempts to fetch another byte for transmission.

TX overflow is indicated by the TX_OVERFLOW interrupt flag being set. When this error occurs, the writing is aborted, i.e., the data byte that caused the overflow is lost. The error condition must be cleared with the SFLUSHTX strobe.

TX underflow is indicated by the TX_UNDERFLOW interrupt flag being set. When this error occurs, the ongoing transmission is aborted. The error condition must be cleared with the SFLUSHTX strobe.

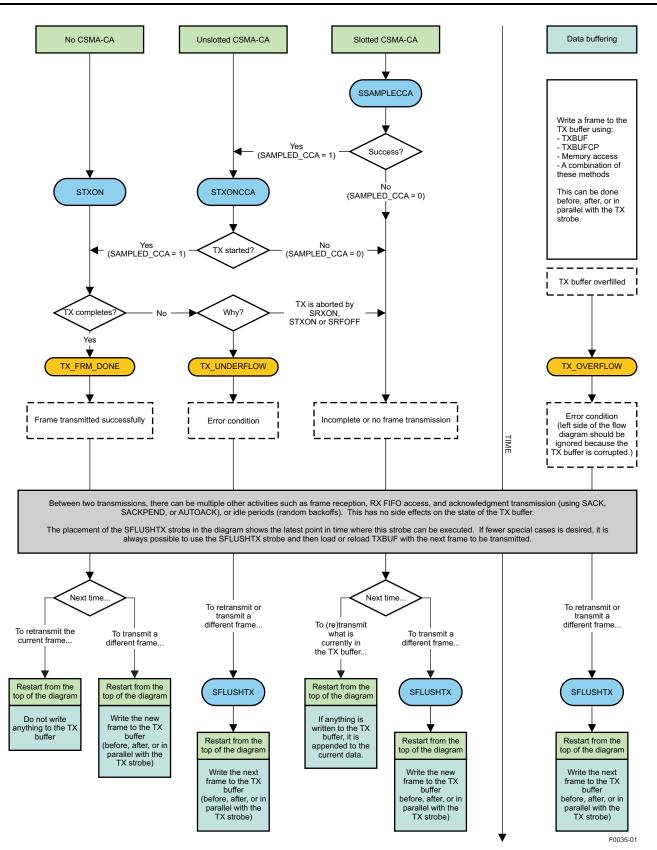
The TX_UNDERFLOW exception can be disabled by setting the FRMCTRL1.IGNORE_TX_UNDERF bit. In this case, the radio continues transmitting the bytes that happen to be in the TXFIFO memory, until the number of bytes given by the first byte (i.e., the length byte) has been transmitted.

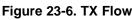
# 23.8.6 TX Flow Diagram

Figure 23-6 summarizes the previous sections in a flow diagram:









# 23.8.7 Transmitted Frame Processing

The radio performs the following frame generation tasks for TX frames:

	_	_	Received Frame		
Preamble	SFD	LEN	MHR	MAC Payload	FCS
(	1)	(2)			(3)
					M0110-0

- (1) Generation and automatic transmission of the PHY layer synchronization header, which consists of the preamble and the SFD
- (2) Transmission of the number of bytes specified by the frame-length field
- (3) Calculation of and automatic transmission of the FCS (can be disabled)

### Figure 23-7. Single Transmitted Frame

The recommended usage is to write the frame-length field followed by the MAC header and MAC payload to the TXFIFO and let the radio handle the rest. Note that the frame-length field must include the two FCS bytes, even though the radio adds these automatically.

# 23.8.8 Synchronization Header

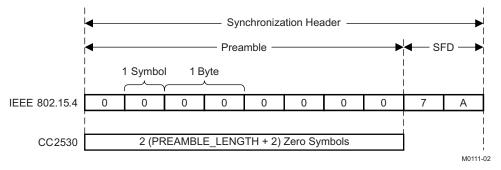


Figure 23-8. Transmitted Synchronization Header

The radio has programmable preamble length. The default value is compliant with [1], and changing the value makes the system noncompliant to IEEE 802.15.4.

The preamble sequence length is set by MDMCTRL0.PREAMBLE_LENGTH. Figure 23-8 shows how the synchronization header relates to the IEEE 802.15.4 specification.

When the required number of preamble bytes has been transmitted, the radio automatically transmits the 1-byte SFD. The SFD is fixed, and it is not possible to change this value from software.

# 23.8.9 Frame-Length Field

When the SFD has been transmitted, the modulator starts to read data from the TXFIFO. It expects to find the frame-length field followed by the MAC header and MAC payload. The frame-length field is used to determine how many bytes are to be transmitted.

Note that the minimum frame length is 3 bytes when AUTOCRC = 1 and 1 byte when AUTOCRC = 0.

# 23.8.10 Frame Check Sequence

When the FRMCTRL0.AUTOCRC control bit is set, the FCS field is automatically generated and appended to the transmitted frame at the position defined by the frame-length field. The FCS is not written to the TXFIFO, but stored in a separate 16-bit register. It is recommended always to have AUTOCRC enabled, except possibly for debug purposes. If FRMCTRL0.AUTOCRC = 0, then the modulator expects to find the FCS in the TXFIFO, so software must generate the FCS and write it to the TXFIFO along with the rest of the MPDU.

The hardware implementation of the FCS calculator is shown in Figure 23-9. See [1] for further details.



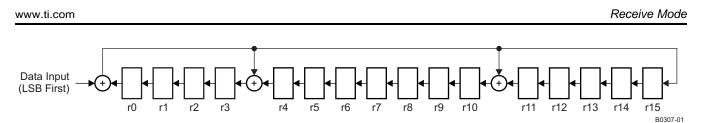


Figure 23-9. FCS Hardware Implementation

### 23.8.11 Interrupts

The SFD interrupt is raised when the SFD field of the frame has been transmitted. At the end of the frame, the TX_FRM_DONE interrupt is raised when the complete frame has been successfully transmitted.

Note that there is a second SFD signal available on GPIO (through radio observation mux) that should not be confused with the SFD interrupt.

### 23.8.12 Clear-Channel Assessment

The clear-channel assessment (CCA) status signal indicates whether the channel is available for transmission or not. The CCA function is used to implement the CSMA-CA functionality specified in the IEEE 802.15.4 specification [1]. The CCA signal is valid when the receiver has been enabled for at least eight symbol periods. The RSSI_VALID status signal can be used to verify this.

The CCA is based on the RSSI value and a programmable threshold. The exact behavior is configurable in the CCACTRL0 and CCACTRL1 registers.

There are two variations of the CCA signal, one that is updated at every new RSSI sample and one that is only updated on SSAMPLECCA/ISAMPLECCA and STXONCCA/ISTXONCCA command strobes. They are both available in the FSMSTAT1 register.

Note that the CCA signal is updated four clock cycles (system clock) after the RSSI_VALID signal has been set.

# 23.8.13 Output Power Programming

The RF output power is controlled by the 7-bit value in the TXPOWER register. The device data sheet (Appendix C) shows typical output power and current consumption for recommended settings when the center frequency is set to 2.440 GHz. Note that the recommended settings are only a small subset of all the possible register settings.

# 23.8.14 Tips and Tricks

- Note that there is no requirement to have the complete frame in the TXFIFO before starting a transmission. Bytes may be added to the TXFIFO during transmission.
- It is possible to transmit non-IEEE 802.15.4 compliant frames by setting MDMTEST1.MODULATION_MODE = 1.

# 23.9 Receive Mode

This section describes how to control the receiver, control the integrated RX frame processing, and how to use the RXFIFO.

# 23.9.1 RX Control

The receiver is turned on and off with the SRXON and SRFOFF command strobes, and with the RXENABLE registers. The command strobes provide a *hard* on/off mechanism, whereas RXENABLE manipulation provides a *soft* on/off mechanism.

The receiver is turned on by the following actions:

• The SRXON strobe:



Receive Mode

- Sets RXENABLE [7]
- Aborts ongoing transmission/reception by forcing a transition to RX calibration.
- The STXON strobe, when FRMCTRL1.SET_RXENMASK_ON_TX is enabled:
  - Sets RXENABLE[6]
  - The receiver is enabled after transmission completes.
- Setting RXENABLE != 0x00 by writing to RXENMASKOR:
  - Does not abort ongoing transmission/reception.

The receiver is turned off by the following actions:

- The SRFOFF strobe:
  - Clears RXENABLE [7:0]
  - Aborts ongoing transmission/reception by forcing the transition to IDLE mode.
- Setting RXENABLE = 0x00 by writing to RXENMASKAND
  - Does not abort ongoing transmission/reception. Once the ongoing transmission/reception is finished, the radio returns to the IDLE state.

There are several ways to manipulate the RXENABLE registers:

- The SRXMASKBITSET and SRXMASKBITCLR strobes (affecting RXENABLE[5])
- The SRXON, SRFOFF and STXON strobes, including the FRMCTRL1.SET_RXMASK_ON_TX setting

### 23.9.2 RX State Timing

The receiver is ready 192 µs after RX has been enabled by one of the methods described in Section 23.9.1. This is referred to as *RX turnaround time* in [1].

When returning to receive mode after frame reception, there is by default an interval of 192 µs where SFD detection is disabled. This interval can be disabled by clearing FSMCTRL.RX2RX_TIME_OFF.

### 23.9.3 Received Frame Processing

The radio integrates critical portions of the RX requirements in IEEE 802.15.4-2003 and -2006 in hardware. This reduces the CPU interruption rate, simplifies the software that handles frame reception, and provides the results with minimum latency.

During reception of a single frame, the following frame-processing steps are performed:

Received Frame								Transm	nitted Ac	knowle	dgment Frame	9
Preamble	SFD	LEN	MHR		MAC Payload	FCS		Preamble	SFD	LEN	MHR	FCS
(	1)		(2)	(3)		(4)				(5)		

M0110-02

- Detection and removal of the received PHY synchronization header (preamble and SFD), and reception of the number of bytes specified by the frame-length field.
- (2) Frame filtering as specified by [1], section 7.5.6.2, third filtering level.
- (3) Matching of the source address against a table containing up to 24 short addresses or 12 extended IEEE addresses. The source address table is stored in the radio RAM.
- (4) Automatic FCS checking, and attaching this result and other status values (RSSI, correlation and source match result) to received frames.
- (5) Automatic acknowledgment transmission with correct timing, and correct setting of the frame pending bit, based on the results from source address matching and FCS checking.

#### Figure 23-10. Single Received Frame and Transmitted Acknowledgement Frame



# 23.9.4 Synchronization Header and Frame-Length Fields

Frame reception starts with detection of a start-of-frame delimiter (SFD), followed by the length byte, which determines when the reception is complete. The SFD signal, which can be output on GPIO, can be used to capture the start of received frames:

Received Frame	Preamble	SFD	LEN	MPDU (LEN[6:0] Bytes)	
					_
SFD (Accepted Frame)					L
SFD (Rejected Frame)					
				<b>↑</b>	
				Frame Rejected	
					T0319-01

Figure 23-11. SFD Signal Timing

Preamble and SFD are not written to the RXFIFO.

The radio uses a correlator to detect the SFD. The correlation threshold value in MDMCTRL1.CORR_THR determines how closely the received SFD must match an *ideal* SFD. The threshold must be adjusted with care:

- · If set too high, the radio misses many actual SFDs, effectively reducing the receiver sensitivity.
- If set too low, the radio detects many false SFDs. Although this does not reduce the receiver sensitivity, the effect is similar, because false frames might overlap with the SFDs of actual frames. It also increases the risk of receiving false frames with correct FCS.

In addition to SFD detection, it is also possible to require a number of valid preamble symbols (also above the correlation threshold) prior to SFD detection. See the register descriptions of MDMCTRL0 and MDMCTRL1 for available options and recommended settings.

# 23.9.5 Frame Filtering

The frame filtering function rejects nonintended frames as specified by [1], section 7.5.6.2, third filtering level. In addition, it provides filtering on:

- The eight different frame types (see the FRMFILT1 register)
- The reserved bits in the frame control field (FCF)

The function is controlled by:

- The FRMFILT0 and FRMFILT1 registers
- The PAN_ID, SHORT_ADDR and EXT_ADDR values in RAM

### Filtering Algorithm

The FRMFILTO.FRM_FILTER_EN bit controls whether frame filtering is applied or not. When disabled, the radio accepts all received frames. When enabled (which is the default setting), the radio only accepts frames that fulfill all of the following requirements:

- The length byte must be equal to or higher than the minimum frame length, which is derived from the source- and destination-address mode and PAN ID compression subfields of the FCF.
- The reserved FCF bits [9:7] ANDed together with FRMFILT0.FCF_RESERVED_BITMASK must equal 000b.
- The value of the frame version subfield of the FCF cannot be higher than FRMFILT0.MAX_FRAME_VERSION.
- The source and destination address modes cannot be reserved values (1).
- Destination address:

235



- If a destination PAN ID is included in the frame, it must match PAN_ID or must be the broadcast PAN identifier (0xFFFF).
- If a short destination address is included in the frame, it must match either SHORT_ADDR or the broadcast address (0xFFFF).
- If an extended destination address is included in the frame, it must match EXT_ADDR.
- Frame type:
  - Beacon frames (0) are only accepted when:
    - FRMFILT1.ACCEPT_FT0_BEACON = 1
    - Length byte  $\geq$  9
    - The destination address mode is 0 (no destination address).
    - The source address mode is 2 or 3 (i.e., a source address is included).
    - The source PAN ID matches PAN_ID, or PAN_ID equals 0xFFFF.
  - Data (1) frames are only accepted when:
    - FRMFILT1.ACCEPT_FT1_DATA = 1
    - Length byte  $\geq$  9
    - A destination address and/or source address is included in the frame. If no destination address is included in the frame, the FRMFILT0.PAN_COORDINATOR bit must be set, and the source PAN ID must equal PAN_ID.
  - Acknowledgment (2) frames are only accepted when:
    - FRMFILT1.ACCEPT_FT2_ACK = 1
    - Length byte = 5
  - MAC command (3) frames are only accepted when:
    - FRMFILT1.ACCEPT_FT3_MAC_CMD = 1
    - Length byte  $\geq$  9
    - A destination address and/or source address is included in the frame. If no destination address is included in the frame, the FRMFILT0.PAN_COORDINATOR bit must be set, and the source PAN ID must equal PAN_ID for the frame to be accepted.
  - Reserved frame types (4, 5, 6, and 7) are only accepted when
    - FRMFILT1.ACCEPT_FT4T07_RESERVED = 1 (default is 0)
    - Length byte  $\geq$  9

The following operations are performed before the filtering begins, with no effect on the frame data stored in the RXFIFO:

- Bit 7 of the length byte is masked out (don't care).
- If FRMFILT1.MODIFY_FT_FILTER is not equal to zero, the MSB of the frame type subfield of the FCF is either inverted or forced to 0 or 1.

If a frame is rejected, the radio only starts searching for a new frame after the rejected frame has been completely received (as defined by the frame-length field) to avoid detecting false SFDs within the frame. Note that a rejected frame can generate RX overflow if it occurs before the frame is rejected.

### Interrupts

When frame filtering is enabled and the filtering algorithm accepts a received frame, an RX_FRM_ACCEPTED interrupt is generated. It is not generated if frame filtering is disabled or RX_OVERFLOW or RX_FRM_ABORTED is generated before the filtering result is known.

Figure 23-12 illustrates the three different scenarios (not including the overflow and abort-error conditions).



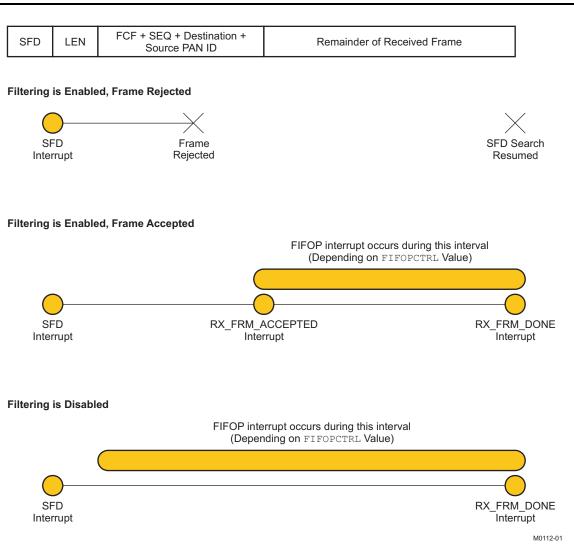


Figure 23-12. Filtering Scenarios (Exceptions Generated During Reception)

The FSMSTAT1.SFD register bit goes high when a start-of-frame delimiter is completely received and remains high until either the last byte in MPDU is received or the received frame has failed to pass address recognition and been rejected.

### **Tips and Tricks**

The following register settings must be configured correctly:

- FRMFILT0.PAN_COORDINATOR must be set if the device is a PAN coordinator, and cleared if not.
- FRMFILTO.MAX_FRAME_VERSION must correspond to the supported version(s) of the IEEE 802.15.4 standard.
- The local address information must be loaded into RAM.

To avoid completely the receiving of frames during energy-detection scanning, set FRMCTRL0.RX_MODE = 11b and then (re)start RX. This disables symbol search and thereby prevents SFD detection.

To resume normal RX mode, set FRMCTRL0.RX_MODE = 00b and (re)start RX.

During operation in a busy IEEE 802.15.4 environment, the radio receives large numbers of nonintended acknowledgment frames. To block reception of these frames effectively, use the FRMFILT1.ACCEPT_FT2_ACK bit to control when acknowledgment frames should be received:

- Set FRMFILT1.ACCEPT_FT2_ACK after successfully starting a transmission with acknowledgment request, and clear the bit again after the acknowledgment frame has been received or the time-out has been reached.
- Keep the bit cleared otherwise.

It is not necessary to turn off the receiver while changing the values of the FRMFILT0/1 registers and the local address information stored in RAM. However, if the changes take place between reception of the SFD byte and the source PAN ID (i.e., between the SFD and RX_FRM_ACCEPTED exceptions), the modified values must be considered as don't care for that particular frame (the radio uses either the old or the new value).

Note that it is possible to make the radio ignore all IEEE 802.15.4 incoming frames by setting MDMTEST1.MODULATION_MODE = 1.

### 23.9.6 Source Address Matching

The radio supports matching of the source address in received frames against a table stored in the on-chip memory. The table is 96 bytes long, and hence it can contain up to:

- 24 short addresses (2 byte pan id + 2 byte short address).
- 12 IEEE extended addresses (8 bytes each).

Source address matching is only performed when frame filtering is also enabled and the received frame has been accepted. The function is controlled by:

- The SRCMATCH, SRCSHORTEN0, SRCSHORTEN1, SRCSHORTEN2, SRCEXTEN0, SRCEXTEN1, and SRCEXTEN2 registers
- The source address table in RAM

#### **Applications**

Automatic acknowledgment transmission with correct setting of the frame-pending bit: When using indirect frame transmission, the devices send data requests to poll frames stored on the coordinator. To indicate whether it actually has a frame stored for the device, the coordinator must set or clear the frame-pending bit in the returned acknowledgment frame. On most 8- and 16-bit MCUs, however, there is not enough time to determine this, and so the coordinator ends up setting the pending bit regardless of whether there are pending frames for the device (as required by IEEE 802.15.4 [1]). This is wasteful in terms of power consumption, because the polling device must keep its receiver enabled for a considerable period of time, even if there are no frames for it. By loading the destination addresses in the indirect frame queue into the source address table and enabling the AUTOPEND function, the radio sets the pending bit in outgoing acknowledgment frames automatically. This way, the operation is no longer timing-critical, as the effort done by the microcontroller is when adding or removing frames in the indirect frame queue and updating the source address table accordingly.

Security material look-up: To reduce the time needed to process secured frames, the source address table can be set up so the entries match the table of security keys on the CPU. A second level of masking on the table entries allows this application to be combined with automatic setting of the pending bit in acknowledgment frames.

*Other applications:* The two previous applications are the main targets for the source-address matching function. However, for proprietary protocols that only rely on the basic IEEE 802.15.4 frame format, there are several other useful applications. For instance, it is possible to create firewall functionality where only a specified set of nodes is to be acknowledged.

### The Source Address Table

The source address table begins at address 0x6100 in RAM. The space is shared between short and extended addresses, and the SRCSHORTEN0/1/2 and SRCEXTEN0/1/2 registers are used to control which entries are enabled. All values in the table are little-endian (as in the received frames).

- A short address entry starts with the 16-bit PAN ID followed by the 16-bit short address. These entries are stored at address 0x6100 + (4 × n), where n is a number between 0 and 23.
- An extended address entry consists only of the 64-bit IEEE extended address. These entries are stored at address 0x6100 + (8 × n), where n is a number between 0 and 11.

### Address Enable Registers



Software is responsible for allocating table entries and for making sure that active short and extended address entries do not overlap. There are separate enable bits for short and extended addresses:

- Short address entries are enabled in the SRCSHORTEN0, SRCSHORTEN1, and SRCSHORTEN2 registers. Register bit **n** corresponds to short address entry **n**.
- Extended address entries are enabled in the SRCEXTEN0, SRCEXTEN1, and SRCEXTEN2 registers. In this case, register bit 2n corresponds to extended address entry n. This mapping is convenient when creating a combined bit vector (of short and extended enable bits) to find unused entries. Moreover, when read, register bit 2n + 1 always has the same value as register bit 2n, because an extended address occupies the same memory as two short-address entries.

#### **Matching Algorithm**

The SRCMATCH.SRC_MATCH_EN bit controls whether source address matching is enabled or not. When enabled (which is the default setting) and a frame passes the filtering algorithm, the radio applies one of the algorithms outlined in Figure 23-15, depending on which type of source address is present.

The result is reported in two different forms:

- A 24-bit vector called SRCRESMASK contains a 1 for each enabled short entry with a match, or two 1s for each enabled extended entry with a match (the bit mapping is the same as for the address-enable registers on read access).
- A 7-bit value called SRCRESINDEX:
  - When no source address is present in the received frame, or there is no match on the received source address:
    - Bits 6:0: 011 1111
  - If there is a match on the received source address:
    - Bits 4:0: The index of the first entry (i.e., the one with the lowest index number) with a match, 0–23 for short addresses or 0–11 for extended addresses.
    - Bit 5: 0 if the match is on a short address, 1 if the match is on an extended address
    - Bit 6: The result of the AUTOPEND function

Short Source Address (Mode 2)	Extended Source Address (Mode 3)
The received source PAN ID is called srcPanid. The received short address is called srcShort.	The received extended address is called srcExt.
<pre>SRCRESMASK = 0x000000; SRCRESINDEX = 0x3F; for (n = 0; n &lt; 24; n++) { bitVector = 0x000001 &lt;&lt; n; if (SRCSHORTEN &amp; bitVector) { if ((panid[n] == srcPanid) &amp;&amp; (short[n] == srcShort)) { SRCRESMASK  = bitVector; if (SRCRESINDEX == 0x3F) { SRCRESINDEX == 0x3F) { SRCRESINDEX = n; } } } }</pre>	<pre>SRCRESMASK = 0x000000; SRCRESINDEX = 0x3F; for (n = 0; n &lt; 12; n++) { bitVector = 0x000003 &lt;&lt; (2*n); if (SRCEXTEN &amp; bitVector) { if (ext[n] == srxExt) { SRCRESMASK  = bitVector; if (SRCRESINDEX == 0x3F) { SRCRESINDEX == 0x3F) { SRCRESINDEX = n   0x20; }</pre>
}	}

### Figure 23-13. Matching Algorithm for Short and Extended Addresses

SRCRESMASK and SRCRESINDEX are written to RF Core memory as soon as the result is available.

SRCRESINDEX is also appended to received frames if the FRMCTRL0.AUTOCRC and FRMCTRL0.APPEND_DATA_MODE bits have been set. The value then replaces the 7-bit correlation value of the 16-bit status word.

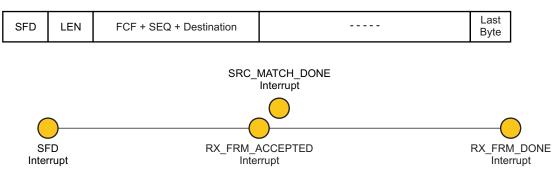
#### Interrupts



When source address matching is enabled and the matching algorithm completes, the *SRC_MATCH_DONE* interrupt flag is set, regardless of the result. If a match is found, the *SRC_MATCH_FOUND* flag is also set immediately before *SRC_MATCH_DONE*.

Figure 23-14 illustrates the timing of the setting of flags:

#### When There Is No Source Address:



#### When There Is a Source Address:

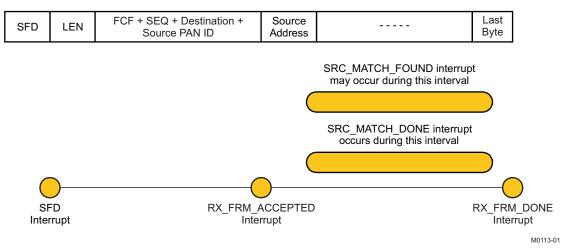


Figure 23-14. Interrupts Generated by Source Address Matching

#### **Tips and Tricks**

The source address table can be modified safely during frame reception. If one address replaces
another while the receiver is active, the corresponding enable bit should be turned off during the
modification. This prevents the RF Core from using a combination of old and new values, because it
only considers entries that are enabled throughout the whole source matching process.

The following measures can be taken to avoid the next received frame overwriting the results from source address matching:

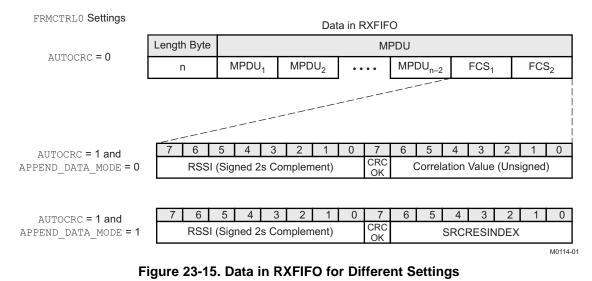
- Use the appended SRCRESINDEX result instead of the value written to RAM (this is the recommended approach).
- Read the results from RAM before RX_FRM_ACCEPTED occurs in the next received frame. For the shortest frame type, this happens after the sequence number, so the total available time (absolute worst-case with a small safety margin) becomes:

16  $\mu$ s (required preamble) + 32  $\mu$ s (SFD) + 128  $\mu$ s (4 bytes) = 176  $\mu$ s

• To increase the available time, clear the FSMCTRL.RX2RX_TIME_OFF bit. This adds another 192 μs, for a total of 368 μs. This also reduces the risk of RX overflow.

# 23.9.7 Frame-Check Sequence

In receive mode, the FCS is verified by hardware if FRMCTRL0.AUTOCRC is enabled. The user is normally only interested in the correctness of the FCS, not the FCS sequence itself. The FCS sequence itself is therefore not written to the RXFIFO during receive. Instead, when FRMCTRL0.AUTOCRC is set, the two FCS bytes are replaced by other more-useful values. The values that are substituted for the FCS sequence are configurable in the FRMCTRL0 register.



Field Descriptions:

- The RSSI value is measured over the first eight symbols following the SFD.
- The *CRC_OK* bit indicates whether the FCS is correct (1) or incorrect (0). When incorrect, software is responsible for discarding the frame.
- The correlation value is the average correlation value over the first eight symbols following the SFD.
- SRCRESINDEX is the same value that is written to RAM after completion of source address matching.

Calculation of the LQI value used by IEEE 802.15.4 is described in Section 23.10.4.

# 23.9.8 Acknowledgement Transmission

The radio includes hardware support for acknowledgment transmission after successful frame reception (i.e., the FCS of the received frame must be correct). Figure 23-16 shows the format of the acknowledgment frame.

Bytes:	4	1	1	2	1	2
	Preamble Sequence	Start-of-Frame Delimiter (SFD)	Frame Length	Frame Control Field (FCF)	Data Sequence Number	Frame Check Sequence (FCS)
Synchronization Header (SHR)			PHY Header (PHR)	MAC Header	(MHR)	MAC Footer (MFR)

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# Figure 23-16. Acknowledge Frame Format

There are three variable fields in the generated acknowledgment frame:

- The pending bit, which may be controlled with command strobes and the AUTOPEND feature (in the FCF field)
- The data sequence number (DSN), which is taken automatically from the last received frame
- The FCS, which is given implicitly



#### Receive Mode

There are three different sources for setting the pending bit in an ACK frame (i.e., the SACKPEND strobe, the PENDING_OR register bit, and the AUTOPEND feature). The pending bit is set if one or more of these sources are set.

#### **Transmission Timing**

Acknowledgment frames can only be transmitted immediately after frame reception. The transmission timing is controlled by the FSMCTRL.SLOTTED_ACK bit.

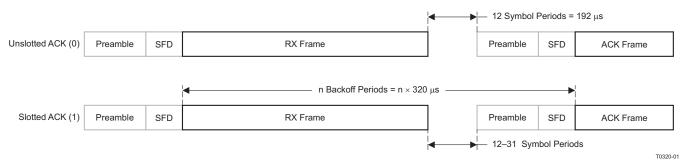


Figure 23-17. Acknowledgment Timing

The IEEE 802.15.4 requires unslotted mode in nonbeacon-enabled PANs, and slotted mode for beacon-enabled PANs.

#### Manual Control

The SACK, SACKPEND, and SNACK command strobes can only be issued during frame reception. If the strobes are issued at any other time, they have no effect but generating a STROBE_ERROR interrupt.

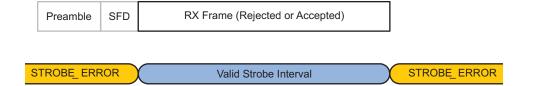


Figure 23-18. Command Strobe Timing

The command strobes may be issued several times during reception; however, only the last strobe has an effect:

- No strobe / SNACK / incorrect FCS: No acknowledgment transmission
- SACK: Acknowledgment transmission with the frame pending-bit cleared
- · SACKPEND: Acknowledgment transmission with the frame pending-bit set

### Automatic Control (AUTOACK)

When FRMFILTO.FRM_FILTER_EN and FRMCTRLO.AUTOACK are both enabled, the radio determines automatically whether or not to transmit acknowledgment frames:

- The RX frame must be accepted by frame filtering (indicated by the RX_FRM_ACCEPTED exception).
- The acknowledgment request bit must be set in the RX frame.
- The RX frame must not be a beacon or an acknowledgment frame.
- The FCS of the RX frame must be correct.

Automatic acknowledgments can be overridden by the SACK, SACKPEND, and SNACK command strobes. For instance, if the microcontroller is low on memory resources and cannot store a received frame, the SNACK strobe can be issued during reception and prevent acknowledging the discarded frame.

By default, the AUTOACK feature never sets the frame-pending bit in the acknowledgment frames. Apart from manual override with command strobes, there are two options:

- Automatic control, using the AUTOPEND feature
- Manual control, using the FRMCTRL1.PENDING_OR bit

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### Automatic Setting of the Frame Pending Field (AUTOPEND)

When the SRCMATCH. AUTOPEND bit is set, the result from source address matching determines the value of the frame-pending field. On reception of a frame, the frame-pending field in the (possibly) returned acknowledgment is set, given that:

- FRMFILT0.FRAME_FILTER_EN is set.
- SRCMATCH.SRC_MATCH_EN is set.
- SRCMATCH.AUTOPEND is set.
- The received frame matches the current SRCMATCH.PEND_DATAREQ_ONLY setting.
- The received source address matches at least one source-match table entry, which is enabled in both SRCSHORTEN and SRCSHORTPENDEN, or SRCEXTEN and SRCEXTPENDEN.

If the source-matching table runs full, the FRMCTRL1.PENDING_OR bit may be used to override the AUTOPEND feature and temporarily acknowledge all frames with the frame-pending field set.

# 23.10 RXFIFO Access

The RXFIFO can hold one or more received frames, provided that the total number of bytes is 128 or less. There are two ways to determine the number of bytes in the RXFIFO:

- Reading the **RXFIFOCNT** register
- Using the FIFOP and FIFO signals in combination with the FIFOPCTRL.FIFOPTHR setting

The RXFIFO is accessed through the RFD register.

The data in the RXFIFO can also be accessed by accessing the radio RAM directly. The FIFO pointers are readable in RXFIRST_PTR, RXLAST_PTR, and RXP1_PTR. This can be useful if one wants to access quickly a certain byte within a frame without having to read out the entire frame first. Note that when using this direct accessing, no FIFO pointers are updated.

The ISFLUSHRX command strobe resets the RXFIFO, resetting all FIFO pointers and clearing all counters, status signals, and sticky error conditions. However, if the receiver is actively receiving a frame when the FIFO is flushed, the RFERRF.ABO flag is asserted.

The SFLUSHRX command strobe resets the RXFIFO, removing all received frames and clearing all counters, status signals, and sticky-error conditions.

# 23.10.1 Using the FIFO and FIFOP

The FIFO and FIFOP signals are useful when reading out received frames in small portions while the frame is received:

- FSMSTAT1.FIFO goes high when one or more bytes are in the RXFIFO, but low when RX overflow has occurred.
- The FSMSTAT1.FIFOP signal goes high when:
  - The number of valid bytes in the RXFIFO exceeds the FIFOP threshold value programmed into FIFOPCTRL. When frame filtering is enabled, the bytes in the frame header are not considered valid until the frame has been accepted.
  - The last byte of a new frame is received, even if the FIFOP threshold is not exceeded. If so, FIFOP goes back to low at the next RXFIFO read access.



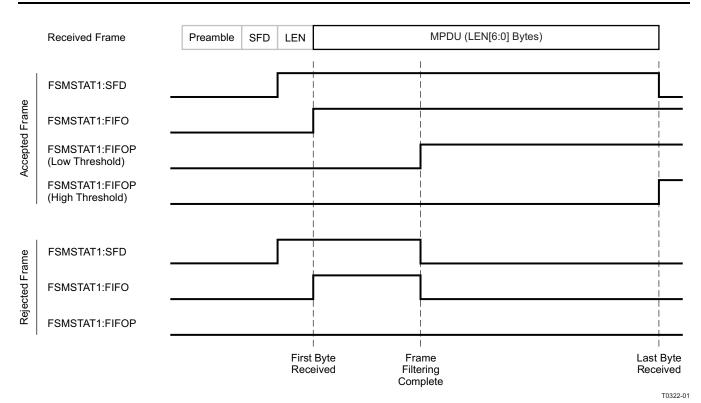


Figure 23-19. Behavior of FIFO and FIFOP Signals

When using the FIFOP as an interrupt source for the microcontroller, the FIFOP threshold should be adjusted by the interrupt service routine to prepare for the next interrupt. When preparing for the last interrupt for a frame, the threshold should match the number of bytes remaining.

# 23.10.2 Error Conditions

There are two error conditions associated with the RXFIFO:

- Overflow, in which case the RXFIFO is full when another byte is received
- Underflow, in which case software attempts to read a byte from an empty RXFIFO

RX overflow is indicated by the RFERRF.RXOVERF flag being set and by the signal values FSMSTAT1.FIFO = 0 and FSMSTAT1.FIFOP = 1. When the error occurs, frame reception is halted. The frames currently stored in the RXFIFO may be read out before the condition is cleared with the ISFLUSHRX strobe. Note that rejected frames can generate RX overflow if the condition occurs before the frame is rejected.

RX underflow is indicated by the RFERRF.RXUNDERF flag being set. RX underflow is a serious error condition that should not occur in error-free software, and the RXUNDERF event should only be used for debugging or in a *watchdog* function. Note that the RXUNDERF error is not generated when the read operation occurs simultaneously with the reception of a new byte.

# 23.10.3 RSSI

The radio has a built-in received signal-strength indication (RSSI), which calculates an 8-bit signed digital value that can be read from a register or automatically appended to received frames. The RSSI value is the result of averaging the received power over eight symbol periods (128  $\mu$ s) as specified by IEEE 802.15.4 [1].

The RSSI value is a 2s-complement signed number on a logarithmic scale with 1-dB steps.



The status bit RSSI_VALID should be checked before reading the RSSI value register. RSSI_VALID indicates that the RSSI value in the register is in fact valid, which means that the receiver has been enabled for at least eight symbol periods.

To find the actual signal power P at the RF pins with reasonable accuracy, an offset must be added to the RSSI value.

P = RSSI - OFFSET [dBm]

For example, with an offset of 73 dB, reading an RSSI value of -10 from the RSSI register means that the RF input power is approximately -83 dBm. For the correct offset value to use, see the data sheet (Appendix C).

There are two ways the radio can update the RSSI register after it has first become valid. If FRMCTRL0.ENERGY_SCAN = 0 (default), the RSSI register contains the latest value available, but if this bit is set to 1, a peak search is performed, and the RSSI register contains the largest value since the energy scan was enabled.

# 23.10.4 Link Quality Indication

The link quality indication (LQI) is a measurement of the strength and/or quality of the received frame as defined by the IEEE 802.15.4 standard [1]. The LQI value is required by the IEEE 802.15.4 standard [1] to be limited to the range 0 through 255, with at least eight unique values. The radio does not provide an LQI value directly, but reports several measurements that can be used by the microcontroller to calculate an LQI value.

The RSSI value can be used by the MAC software to calculate the LQI value. This approach has the disadvantage that, e.g., a narrowband interferer inside the channel bandwidth can increase the RSSI and thus the LQI value, although the true link quality actually decreases. The radio therefore also provides an average correlation value for each incoming frame, based on the first eight symbols following the SFD. This unsigned 7-bit value can be looked on as a measurement of the *chip error rate*, although the radio does not do chip decision.

As described in Section 23.9.7, the average correlation value for the first eight symbols is appended to each received frame, together with the RSSI and CRC OK/not OK, when FRMCTRL0.AUTOCRC is set. A correlation value of ~110 indicates a maximum-quality frame, whereas a value of ~50 is typically the lowest-quality frame detectable by the radio.

Software must convert the correlation value to the range 0–255 as defined by [1], for instance by calculating:

LQI = (CORR - a)b

limited to the range 0–255, where a and b are found empirically based on PER measurements as a function of the correlation value.

A combination of RSSI and correlation values may also be used to generate the LQI value.

# 23.11 Radio Control State Machine

The FSM module is responsible for maintaining the TXFIFO and RXFIFO pointers, control of analog *dynamic* signals such as power up/power down, control of the data flow within the RF Core, generation of automatic acknowledgement frames, and control of all analog RF calibration.



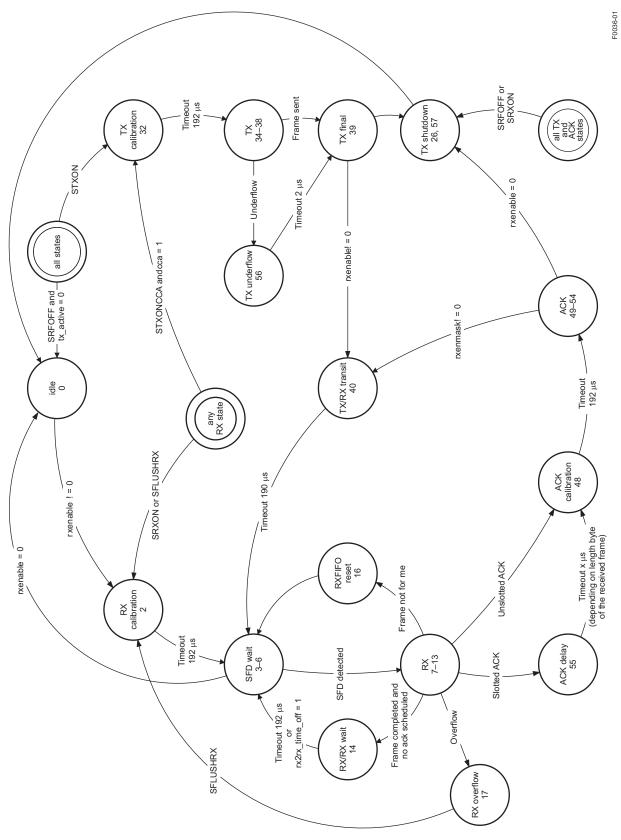


Figure 23-20. Main FSM

Table 23-3 shows the mapping from FSM state to the number which can be read from the FSMSTAT0 register. Note that although it is possible to read the state of the FSM, this information should not be used to control the program flow in the application software. The states may change very quickly (every 32-MHz clock cycle), and an 8-MHz SPI is not able to capture all the activities.

State Name	State Number, Decimal	Number, Hex	TX_ACTIV E	RX_ACTIVE
Idle	0	0x00	0	0
RX calibration	2	0x02	0	1
SFD wait	3–6	0x03–0x06	0	1
RX	7–13	0x07–0x0D	0	1
RX/RX wait	14	0x0E	0	1
RXFIFO reset	16	0x10	0	1
RX overflow	17	0x11	0	0
TX calibration	32	0x20	1	0
ТХ	34–38	0x22–0x26	1	0
TX final	39	0x27	1	0
TX/RX transit	40	0x28	1	0
ACK calibration	48	0x30	1	0
ACK	49–54	0x31–0x36	1	0
ACK delay	55	0x37	1	0
TX underflow	56	0x38	1	0
TX shutdown	26, 57	0x1A, 0x39	1	0

Table 23-3. FSM State Mapping

# 23.12 Random-Number Generation

The RF Core can generate random bits. The chip should be in RX when generation of random bits is required. One must also make sure that the chip has been in RX long enough for the transients to have died out. A convenient way to do this is to wait for the RSSI-valid signal to go high.

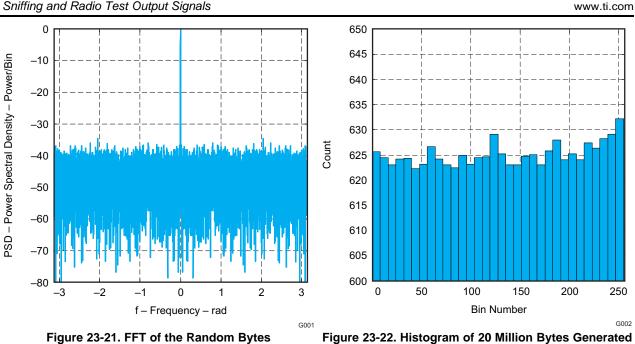
Single random bits from either the I or Q channel can be read from the register RFRND.

Randomness tests show good results for this module. However, a slight dc component exists. In a simple test where the RFRND.IRND register was read a number of times and the data grouped into bytes, about 20 million bytes were read. When interpreted as unsigned integers between 0 and 255, the mean value was 127.6518, which indicates that there is a dc component.

The FFT of the first 2¹⁴ bytes is shown in Figure 23-21. Note that the dc component is clearly visible. A histogram (32 bins) of the 20 million values is shown in Figure 23-22.







With the RANDOM Instruction

For the first 20 million individual bits, the probability of a one is P(1) = 0.500602 and P(0) = 1 - P(1) =0.499398.

Note that to fully qualify the random generator as *true random*, much more elaborate tests are required. There are software packages available on the internet that may be useful in this respect [8], [9].

# 23.13 Packet Sniffing and Radio Test Output Signals

Packet sniffing is a nonintrusive way of observing data that is either being transmitted or received. The packet sniffer outputs a clock and a data signal, which should be sampled on the rising edges of the clock. The two packet sniffer signals are observable as GPIO outputs. For accurate time stamping, the SFD signal should also be output.

Because the radio has a data rate of 250 kbps, the packet sniffer clock frequency is 250 kHz. The data is output serially, with the MSB of each byte first, which is opposite of the actual RF transmission, but more convenient when processing the data. It is possible to use a SPI slave to receive the data stream.

When sniffing frames in TX mode, the data that is read from the TXFIFO by the modulator is the same data that is output by the packet sniffer. However, if automatic CRC generation is enabled, the packet sniffer does NOT output these 2 bytes. Instead, it replaces the CRC bytes with 0x8080. This value can never occur as the last two bytes of a received frame (when automatic CRC checking is enabled), and thus it provides a way for the receiver of the sniffed data to separate frames that were transmitted and frames that were received.

When sniffing frames in RX mode, the data that is written to the RXFIFO by the demodulator is the same data that is output by the packet sniffer. In other words, the last two bytes are either the received CRC value or the CRC OK/RSSI/correlation/SRCRESINDEX value that may automatically replace the CRC value, depending on configuration settings.

To set up the packet sniffer signals or some of the other RF Core observation outputs (in total maximum 3; rfc obs siq0, rfc obs siq1, and rfc obs siq2), the user must perform the following steps:

Step1: Determine which signal (rfc_obs_sig) to output on which GPIO pin (P1[0:5]). This is done using the OBSSELx control registers (OBSSEL0-OBSSEL5) that control the observation output to pins P1[0:5] (overriding the standard GPIO behavior for those pins).

Step2: Set the RFC_OBS_CTRL control registers (RFC_OBS_CTRL0-RFC_OBS_CTRL2) to select the correct signals (rfc_obs_sig); e.g., for packet sniffing one needs the rfc_sniff_data for the packet sniffer data signal and rfc sniff clk for the corresponding clock signal.

**Step3**: For packet sniffing, the packet sniffer module must be enabled in the MDMTEST1 register.

# 23.14 Command Strobe/CSMA-CA Processor

The command strobe/CSMA-CA processor (CSP) provides the control interface between the CPU and the radio.

The CSP interfaces with the CPU through the SFR register RFST and the XREG registers CSPX, CSPY, CSPZ, CSPT, CSPSTAT, CSPCTRL, and CSPPROG<n> (where n is in the range 0 to 23). The CSP produces interrupt requests to the CPU. In addition, the CSP interfaces with the MAC Timer by observing MAC Timer events.

The CSP allows the CPU to issue command strobes to the radio, thus controlling the operation of the radio.

The CSP has two modes of operation, which are described as follows.

- Immediate command strobe execution
- Program execution

Immediate command strobes are written as Immediate Command Strobe instructions to the CSP, which are issued instantly to the radio module. The Immediate Command Strobe instructions are also used to control the CSP. The Immediate Command Strobe instructions are described in Section 23.14.8.

Program execution mode means that the CSP executes a sequence of instructions, comprising a short user-defined program, from a program memory or instruction memory. The available instructions are from a set of 20 instructions. The instruction set is defined in Section 23.14.8. The required program is first loaded into the CSP by the CPU, and then the CPU instructs the CSP to start executing the program.

The program execution mode, together with the MAC Timer, allows the CSP to automate CSMA-CA algorithms and thus act as a coprocessor for the CPU.

The operation of the CSP is described in detail in the following sections. The command strobes and other instructions supported by the CSP are given in Section 23.14.9.

Bit	Name	Reset	R/W	Description
7:0	<pre>INSTR[7:0]</pre>	0xD0	R/W	Data written to this register is written to the CSP instruction memory. Reading this register returns the CSP instruction currently being executed.

#### RFST (0xE1) - RF CSMA-CA/Strobe Processor

# 23.14.1 Instruction Memory

The CSP executes single-byte program instructions which are read from a 24-byte instruction memory. Writes to the instruction memory are sequential, written through SFR register RFST. An instruction write pointer is maintained within the CSP to hold the location within the instruction memory where the next instruction written to RFST is to be stored. For debugging purposes, the program currently loaded into the CSP can be read from the XREG registers CSPPROG<n>. Following a reset, the write pointer is reset to location 0. During each RFST register write, the write pointer is incremented by 1 until the end of memory is reached, at which time the write pointer stops incrementing. The first instruction written to RFST is stored in location 0, the location where program execution starts. Thus, a complete 24-instruction program is written to the instruction memory by writing each instruction in the desired order to the RFST register.

The write pointer can be reset to 0 by writing the immediate command strobe instruction ISSTOP. In addition, the write pointer is reset to 0 when the command strobe SSTOP is executed in a program.

Following a reset, the instruction memory is filled with SNOP (No Operation) instructions (opcode value 0xC0). The immediate strobe ISCLEAR clears the instruction memory, filling it with SNOP instructions.

While the CSP is executing a program, there must be no attempts to write instructions to the instruction memory by writing to RFST. Failure to observe this rule can lead to incorrect program execution and corrupt instruction memory contents. However, Immediate Command Strobe instructions may be written to RFST (see Section 23.14.3).



#### 23.14.2 Data Registers

The CSP has three data registers, CSPT, CSPX, CSPY, and CSPZ, which are read/write accessible for the CPU as XREG registers. These registers are read or modified by some instructions, thus allowing the CPU to set parameters to be used by a CSP program, or allowing the CPU to read CSP program status.

The CSPT data register is not modified by any instruction. The CSPT data register is used to set a MAC Timer overflow-compare value. Once program execution has started on the CSP, the content of this register is decremented by 1 each time the MAC Timer overflows. When CSPT reaches zero, program execution is halted and the interrupt IRQ_CSP_STOP is asserted. The CSPT register is not decremented if the CPU writes 0xFF to this register.

### 23.14.3 Program Execution

After the instruction memory has been filled, program execution is started by writing the immediate command strobe instruction ISSTART to the RFST register. Program execution continues until the instruction at the last location has been executed, the CSPT data register content is zero, an SSTOP instruction has been executed, an immediate ISSTOP instruction is written to RFST, or a SKIP instruction returns a location beyond the last location in the instruction memory. The CSP runs at the set system clock frequency, which must be set to 32 MHz for correct radio operation.

Immediate command strobe instructions may be written to RFST while a program is being executed. In this case, the immediate instruction is executed before the instruction in the instruction memory, which is executed once the immediate instruction has been completed.

During program execution, reading RFST returns the current instruction being executed. An exception to this is the execution of immediate command strobes, during which RFST returns 0xD0.

### 23.14.4 Interrupt Requests

The CSP has three interrupt flags which can produce the RF interrupt vector. These are the following:

- IRQ_CSP_STOP: asserted when the processor has executed the last instruction in memory or when the processor stops due to an SSTOP or ISSTOP instruction or the CSPT register being equal to zero
- IRQ_CSP_WT: asserted when the processor continues executing the next instruction after a WAIT W
  or WAITX instruction
- IRQ_CSP_INT: asserted when the processor executes an INT instruction

### 23.14.5 Random Number Instruction

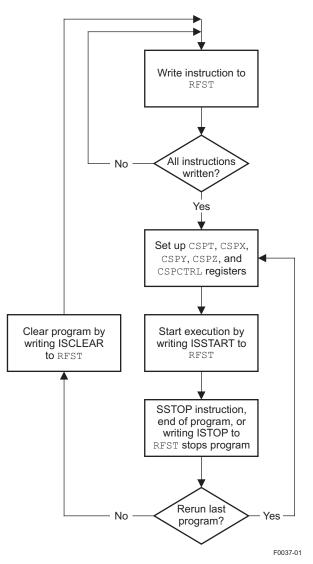
There is a delay in the update of the random number used by the RANDXY instruction. Therefore, if the instruction RANDXY, which uses this value, is issued immediately after a previous RANDXY instruction, the random value read may be the same in both cases.

# 23.14.6 Running CSP Programs

The basic flow for loading and running a program on the CSP is shown in Figure 23-23. When program execution stops at the end of the program, the current program remains in program memory so that the same program can be run again by starting execution once again with the ISSTART command. To clear the program contents, use the ISCLEAR instruction.

**NOTE:** If the CSPT register compare function is not used, this register must be set to 0xFF before the program execution is started.







# 23.14.7 Registers

#### CSPROG<N> (N Ranging From 0 to 23) (0x61C0 + N) - CSP Program

Bit	Name	Reset	R/W	Description
7:0	CSP_INSTR	0xD0	R	Byte N of the CSP program memory

### CSPCTRL (0x61E0) – CSP Control Bit

Bit	Name	Reset	R/W	Description
7:1	—	0000 000	R0	Reserved. Read as 0
0	MCU_CTRL	0	R/W	CSP MCU control input



### Command Strobe/CSMA-CA Processor

CSPSTAT (0x61E1) – CSP Status Register					
Bit	Name	Reset	R/W	Description	
7:6	—	00	R0	Reserved. Read as 0	
5	CSP_RUNNING	0	R	1: CSP is running. 0: CSP is idle.	
4:0	CSP_PC	0 0000	R	CSP program counter	

#### CSPX (0x61E2) – CSP X Register

Bit	Name	Reset	R/W	Description
7:0	CSPX	0x00	R/W	CSP X data register. Used by CSP instructions WAITX, RANDXY, INCX, DECX, and conditional instructions

#### CSPY (0x61E3) - CSP Y Register

Bit	Name	Reset	R/W	Description
7:0	CSPY	0x00	R/W	CSP Y data register. Used by CSP instructions RANDXY, INCY, DECY, and conditional instructions

#### CSPZ (0x61E4) - CSP Z Register

Bit	Name	Reset	R/W	Description
7:0	CSPZ	0x00	R/W	CSP Z data register. Used by CSP instructions INCZ, DECZ, and conditional instructions

#### CSPT (0x61E5) - CSP T Register

Bit	Name	Reset	R/W	Description
7:0	CSPT	0xFF	R/W	CSP T data register. Content is decremented each time the MAC Timer overflows while the CSP program is running. The SCP program stops when decremented to 0. Setting CSPT = $0xFF$ prevents the register from being decremented.

### 23.14.8 Instruction Set Summary

This section gives an overview of the instruction set. This is intended as a summary and definition of instruction opcodes. See Section 23.14.9 for a description of each instruction. Each instruction consists of one byte, which is written to the RFST register to be stored in the instruction memory.

The Immediate Strobe instructions (ISxxx) are not used in a program. When these instructions are written to the RFST register, they are executed immediately. If the CSP is already executing a program, the current instruction is delayed until the immediate strobe instruction has completed.

For undefined opcodes, the behavior of the CSP is defined as a no-operation strobe command (SNOP).

### Command Strobe/CSMA-CA Processor

Mnemonic	7	6	5	4	3	2	1	0	Description
SKIP <c>, <s></s></c>	0	S2	S1	S0	N	C2	C1	C0	Skip S instructions on condition C. When condition (C XOR N) is true, skip the next S instructions, else execute the next instruction. If $S = 0$ , re-execute the conditional jump (i.e., busy loop until condition is false). Skipping past the last instruction in the command buffer results in an implicit STOP command. The conditions are:
									C = 0 CCA true
									C = 1 Synchronization word received and still receiving packet or synchronization word transmitted and still transmitting packet (SFD found, not yet frame end)
									C = 2 MCU control bit is 1.
									C = 3 Reserved
									C = 4 Register $X = 0$
									C = 5 Register $Y = 0$
									C = 6 Register $Z = 0$
									C = 7 RSSI_VALID = 1
WAIT <w></w>	1	0	0	W4	W3	W2	W1	W0	Wait for MAC Timer to overflow W times. Waits until the MAC Timer has overflowed W times (W = 0 waits 32 times), then continues execution. Generates an IRQ_CSP_WAIT interrupt request when execution continues.
RPT <c></c>	1	0	1	0	N	C2	C1	C0	Repeat loop while condition C. If condition C is true, go to the instruction following the last LABEL instruction (address in loop-start register); if the condition is false or no LABEL instruction has been executed, go to the next instruction.
									Note condition C is as defined for SKIP, defined previously in this table. It is not possible to have a RPT instruction placed at index 23 of the command buffer.
WEVENT1	1	0	1	1	1	0	0	0	Wait for mact_event1 to go high, and then continue execution.
WEVENT2	1	0	1	1	1	0	0	1	Wait for mact_event2 to go high, and then continue execution.
INT	1	0	1	1	1	0	1	0	Generate an IRQ_CSP_MANINT. Issues an IRQ_CSP_MANINT interrupt request.
LABEL	1	0	1	1	1	0	1	1	Set the next instruction as the start of a repeat loop. Enters the address of the next instruction into the loop-start register.
WAITX	1	0	1	1	1	1	0	0	Wait for MAC Timer to overflow [X] times, where [X] is the value of register X. Each time a MAC Timer overflow is detected, X is decremented. Execution continues as soon as $X = 0$ . (If $X = 0$ when instruction is run, no wait is performed and execution continues directly). An IRQ_CSP_WAIT interrupt request is generated when execution continues.
RANDXY	1	0	1	1	1	1	0	1	Load the [Y] LSBs of register X with random value.
SETCMP1	1	0	1	1	1	1	1	0	Set the output csp_mact_setcmp1 high. This sets the compare value of the MAC Timer to the current timer value.
INCX	1	1	0	0	0	0	0	0	Increment register X.
INCY	1	1	0	0	0	0	0	1	Increment register Y.
INCZ	1	1	0	0	0	0	1	0	Increment register Z.
DECX	1	1	0	0	0	0	1	1	Decrement register X.
DECY	1	1	0	0	0	1	0	0	Decrement register Y.
DECZ	1	1	0	0	0	1	0	1	Decrement register Z.
INCMAXY <m></m>	1	1	0	0	1	M2	M1	M0	Register $Y \le min(Y + 1, M)$ . Increment Y, but not beyond M.
Sxxx	1	1	0	1	S3	S2	S1	S0	Execute command strobe S. Send command strobe S to FFCTRL. Up to 32 command strobes are supported. In addition to the regular command strobes, two additional command strobes that only apply to the command strobe processor are supported: SNOP: Do nothing.
									SSTOP: Do nothing. SSTOP: Stops the command strobe processor execution and invalidates any set label. An IRQ_CSP_STOP interrupt request is issued.

Table 23-4. Instruction Set Summary



Mnemonic	7	6	5	4	3	2	1	0	Description
ISxxx	1	1	1	0	S3	S2	S1	SO	Execute command strobe S immediately. Send command strobe S to FFCTRL immediately, bypassing the instructions in the command buffer. If the current buffer instruction is a strobe, it is delayed. In addition to the regular command strobes, two additional command strobes that only apply to the command strobe processor are supported:
									ISSTART: The command strobe processor starts execution at the first instruction in the command buffer. Do not issue an ISSTART instruction if the CSP is already running.
									ISSTOP: Stops the command strobe processor execution and invalidates any set label. An IRQ_CSP_STOP interrupt request is issued.
ISCLEAR	1	1	1	1	1	1	1	1	Clear the CSP program. Reset PC.

 Table 23-4. Instruction Set Summary (continued)

### 23.14.9 Instruction Set Definition

There are 20 basic instruction types. Furthermore, the command-strobe and immediate-strobe instructions can each be divided into 16 subinstructions, giving an effective number of 42 different instructions. The following subsections describe each instruction in detail.

Note: the following definitions are used in this section

PC = CSP program counter X = RF register CSPX Y = RF register CSPY Z = RF register CSPZ T = RF register CSPT

### 23.14.9.1 DECZ

Function:Decrement ZDescription:The Z register is decremented by 1. An original value of 0x00 underflows to 0xFF.Operation:Z = Z - 1

#### Opcode: 0xC5

7	6	5	4	3	2	1	0
1	1	0	0	0	1	0	1

### 23.14.9.2 DECY

Function:Decrement YDescription:The Y register is decremented by 1. An original value of 0x00 underflows to 0xFF.Operation:Y = Y - 1

#### Opcode: 0xC4

7	6	5	4	3	2	1	0
1	1	0	0	0	1	0	0

#### 23.14.9.3 DECX

TEXAS INSTRUMENTS

Command Strobe/CSMA-CA Processor

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Function:	Decrement X
Description:	The X register is decremented by 1. An original value of 0x00 underflows to 0xFF.
Operation:	X = X - 1

Opcode: 0xC3

7	6	5	4	3	2	1	0
1	1	0	0	0	0	1	1

### 23.14.9.4 INCZ

Function:Increment ZDescription:The X register is incremented by 1. An original value of 0xFF overflows to 0x00.Operation:Z = Z + 1

Opcode: 0xC2

	-	_	_	-	-	-	-
7	6	5	4	3	2	1	0
1	1	0	0	0	0	1	0

### 23.14.9.5 INCY

Function:	Increment Y
Description:	The Y register is incremented by 1. An original value of 0xFF overflows to 0x00.
Operation:	Y = Y + 1

Opcode: 0xC1

7	6	5	4	3	2	1	0
1	1	0	0	0	0	0	1

### 23.14.9.6 INCX

Function:	Increment X
Description:	The X register is incremented by 1. An original value of 0xFF overflows to 0x00.
Operation:	X = X + 1

Opcode: 0xC0

7	6	5	4	3	2	1	0
1	1	0	0	0	0	0	0

### 23.14.9.7 INCMAXY

Function:	Increment Y not greater than M.
Description:	The Y register is incremented by 1 if the result is less than M; otherwise, Y register is loaded with value M.
Operation:	Y = min(Y + 1, M)



Command Strobe/CSMA-CA Processor

	Opcode: 0xC8   M (M = 0–7)							
Ī	7	6	5	4	3	2	1	0
	1	1	0	0	1		М	

#### 23.14.9.8 RANDXY

Function:	Load random value into X.
Description:	The [Y] LSBs of the X register are loaded with a random value. Note that if a second RANDXY instruction is issued immediately (within 13 clock cycles) after the first, the same random value is used in both cases. If Y equals zero or is greater than 7, then an 8-bit random value is loaded into X.
Operation:	$X[(Y - 1):0]: = RNG_DOUT[(Y - 1):0], X[7:Y]: = 0$

#### Opcode: 0xBD

7	6	5	4	3	2	1	0
1	0	1	1	1	1	0	1

### 23.14.9.9 INT

Function:	Interrupt
Description:	The interrupt IRQ_CSP_INT is asserted when this instruction is executed.
Operation:	IRQ_CSP_INT = 1

7	6	5	4	3	2	1	0
1	0	1	1	1	0	1	0

### 23.14.9.10 WAITX

Function:	Wait for X MAC Timer overflows
Description:	Wait for MAC Timer to overflow [X] times, where [X] is the value of register X. Each time a MAC Timer overflow is detected, the value in register X is decremented. Program execution continues as soon as $X = 0$ . (If $X = 0$ when instruction is run, no wait is performed and execution continues directly.) An IRQ_CSP_WAIT interrupt request is generated when execution continues. <b>Note:</b> The difference compared to WAIT W is that W is a fixed value, whereas X is a register value (which could potentially be changed, such that the number of overflows actually does not correspond to the value of X at the time WAITX instruction is run).
Operation:	X = X - 1 when MAC Timer overflow = true PC = PC while X > 0

FС	= =	WIII		U	
PC	= PC	+ 1	when	X =	0

#### Opcode: 0xBC

7	6	5	4	3	2	1	0
1	0	1	1	1	1	0	0

### 23.14.9.11 SETCMP1



#### Command Strobe/CSMA-CA Processor

Function:Set the compare value of the MAC Timer to the current timer value.Description:Set the compare value of the MAC Timer to the current timer value.Operation:Csp_mact_setcmp1 = 1

#### Opcode: 0xBE

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7	6	5	4	3	2	1	0
1	0	1	1	1	1	1	0

### 23.14.9.12 WAIT W

Function:	Wait for W MAC Timer overflows
Description:	Wait until MAC Timer overflows a number of times equal to the value of W. If $W = 0$ , the instruction waits for 32 overflows. Program execution continues with the next instruction, and the interrupt flag IRQ_CSP_WT is asserted when the wait condition is true.
Operation:	PC = PC while number of MAC Timer overflows < W PC = PC + 1 when number of MAC Timer overflows = W

#### Opcode: 0x80 | W (W = 0-31)

7	6	5	4	3	2	1	0
1	0	0			W		

### 23.14.9.13 WEVENT1

Function:	Wait until MAC Timer event 1
Description:	Wait until next MAC Timer event. Program execution continues with the next instruction when the wait condition is true.
Operation:	PC = PC while MAC Timer compare = false PC = PC + 1 when MAC Timer compare = true

#### Opcode: 0xB8

7	6	5	4	3	2	1	0
1	0	1	1	1	0	0	0

### 23.14.9.14 WEVENT2

Function:	Wait until MAC Timer event 2
Description:	Wait until next MAC Timer event. Program execution continues with the next instruction when the wait condition is true.
Operation:	PC = PC while MAC Timer compare = false PC = PC + 1 when MAC Timer compare = true

#### Opcode: 0xB9

7	6	5	4	3	2	1	0
1	0	1	1	1	0	0	1



### 23.14.9.15 LABEL

Function:	Set loop label
Description:	Sets next instruction as start of loop. If the current instruction is the last instruction in the instruction memory, then the current PC is set as start of loop. If several label instructions are executed, the last label executed is the active label. Earlier labels are removed, which means that only one level of loops is supported.
Operation:	LABEL: $= PC + 1$

#### Opcode: 0xBB

7	6	5	4	3	2	1	0
1	0	1	1	1	0	1	1

### 23.14.9.16 RPT C

### Function: Conditional repeat

**Description:** If condition C is true, then jump to the instruction defined by the last LABEL instruction, i.e., jump to start of loop. If the condition is false or if a LABEL instruction has not been executed, then execution continues from next instruction. The condition C may be negated by setting N = 1 and is described in the following table.

1

Condition Code C	Description	Function
000	CCA is true	CCA = 1
001	Synchronization word received and still receiving packet or synchronization word transmitted and still transmitting packet	SFD = 1
010	CPU control true	CSPCTRL.CPU_CTRL =
011	Reserved	
100	Register X = 0	X = 0
101	Register $Y = 0$	Y = 0
110	Register Z = 0	Z = 0
111	RSSI is valid	RSSI_VALID = 1

**Operation:** PC = LABEL when (C XOR N) = true PC = PC + 1 when (C XOR N) = false or LABEL = not set

#### Opcode: 0xA0 | N | C (N = 0, 8; C = 0–7)

7	6	5	4	3	2	1	0
1	0	1	0	Ν		С	

### 23.14.9.17 SKIP S, C

#### Function: Conditional skip instruction

**Description:** Skip S instructions on condition C (where condition C may be negated; N = 1). When condition (C xor N) is true, skip the next S instructions, else execute the next instruction. If S = 0, re-execute the conditional jump (i.e., busy loop until condition is false). Skipping past the last instruction in the command buffer results in an implicit STOP command.

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Condition Code C	Description	Function
000	CCA is true	CCA = 1
001	Synchronization word received and still receiving packet or synchronization word transmitted and still transmitting packet	SFD = 1
010	CPU control true	$CSPCTRL.CPU_CTRL = 1$
011	Reserved	
100	Register X = 0	X = 0
101	Register Y = 0	Y = 0
110	Register Z = 0	Z = 0
111	RSSI is valid	RSSI_VALID = 1

**Operation:** PC = PC + S + 1 when (C XOR N) = true PC = PC + 1 when (C XOR N) = false

#### Opcode: 0x00 | S | N | C

7	6	5	4	3	2	1	0
0		S		Ν		С	

### 23.14.9.18 STOP

Function:	Stop program execution
Description:	The SSTOP instruction stops the CSP program execution.
Operation:	Stop execution

### Opcode: 0xD2

-							
7	6	5	4	3	2	1	0
1	1	0	1	0	0	1	0

### 23.14.9.19 SNOP

Function:	No operation
Description:	Operation continues at the next instruction.
Operation:	PC = PC + 1

#### Opcode: 0xD0

7	6	5	4	3	2	1	0
1	1	0	1	0	0	0	0

#### 23.14.9.20 SRXON

Function:	Enable and calibrate frequency synthesizer for RX
Description:	The SRXON instruction asserts the output FFCTL_SRXON_STRB to enable and calibrate the frequency synthesizer for RX. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	SRXON



Command Strobe/CSMA-CA Processor

7	6	5	4	3	2	1	0
1	1	0	1	0	0	1	1

### 23.14.9.21 STXON

Function:	Enable TX after calibration
Description:	The STXON instruction enables TX after calibration. The instruction waits for the radio to acknowledge the command before executing the next instruction. Sets a bit in RXENABLE if SET_RXENMASK_ON_TX is set
•	

Operation: STXON

#### Opcode: 0xD9

7	6	5	4	3	2	1	0
1	1	0	1	1	0	0	1

### 23.14.9.22 STXONCCA

Function:	Enable calibration and TX if CCA indicates a clear channel
Description:	The STXONCCA instruction enables TX after calibration if CCA indicates a clear channel.
Operation:	STXONCCA

#### Opcode: 0xDA

7	6	5	4	3	2	1	0
1	1	0	1	1	0	1	0

### 23.14.9.23 SSAMPLECCA

Function:	Sample the current CCA value to SAMPLED_CCA
Description:	The current CCA value is written to ${\tt SAMPLED_CCA}$ in XREG.
Operation:	SSAMPLECCA

#### Opcode: 0xDB

7	6	5	4	3	2	1	0
1	1	0	1	1	0	1	1

### 23.14.9.24 SRFOFF

Function:	Disable RX/TX and frequency synthesizer.
Description:	The SRFOFF instruction disables RX/TX and the frequency synthesizer.
Operation:	SRFOFF

Opcode: 0xDF

7	6	5	4	3	2	1	0
1	1	0	1	1	1	1	1

### 23.14.9.25 SFLUSHRX

Function:	Flush RXFIFO buffer and reset demodulator
Description:	The SFLUSHRX instruction flushes the RXFIFO buffer and resets the demodulator. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	SFLUSHRX

Opcode: 0xDD

7	6	5	4	3	2	1	0
1	1	0	1	1	1	0	1

### 23.14.9.26 SFLUSHTX

Function:	Flush TXFIFO buffer
Description:	The SFLUSHTX instruction flushes the TXFIFO buffer. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	SFLUSHTX

#### Opcode: 0xDE

7	6	5	4	3	2	1	0
1	1	0	1	1	1	1	0

### 23.14.9.27 SACK

Function:	Send acknowledge frame with pending field cleared
Description:	The SACK instruction sends an acknowledge frame. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	SACK

#### Opcode: 0xD6

7	6	5	4	3	2	1	0
1	1	0	1	0	1	1	0

### 23.14.9.28 SACKPEND

Function:	Send acknowledge frame with the pending field set
Description:	The SACKPEND instruction sends an acknowledge frame with the pending field set. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	SACKPEND

#### Opcode: 0xD7

7	6	5	4	3	2	1	0
1	1	0	1	0	1	1	1



### 23.14.9.29 SNACK

Function:	Abort sending of acknowledge frame
Description:	The SACKPEND instruction aborts sending acknowldedge to the frame currently being received.
Operation:	SNACK

#### Opcode: 0xD8

7	6	5	4	3	2	1	0
1	1	0	1	1	0	0	0

### 23.14.9.30 SRXMASKBITSET

Function:	Set bit in RXENABLE
Description:	The SRXMASKBITSET instruction sets bit 5 in the RXENABLE register.
Operation:	SRXMASKBITSET

Opcode: 0xD4

7	6	5	4	3	2	1	0
1	1	0	1	0	1	0	0

### 23.14.9.31 SRXMASKBITCLR

Function:	Clear bit in RXENABLE
Description:	The SRXMASKBITCLR instruction clears bit 5 in the $\ensuremath{\mathtt{RXENABLE}}$ register.
Operation:	SRXMASKBITCLR

#### Opcode: 0xD5

7	6	5	4	3	2	1	0
1	1	0	1	0	1	0	1

#### 23.14.9.32 ISSTOP

Function:	Stop program execution
Description:	The ISSTOP instruction stops the CSP program execution and the IRQ_CSP_STOP interrupt flag is asserted.
Operation:	Stop execution

#### Opcode: 0xE2

7	6	5	4	3	2	1	0
1	1	1	0	0	0	1	0

### 23.14.9.33 ISSTART



Function:	Start program execution
Description:	The ISSTART instruction starts the CSP program execution from first instruction written to instruction memory. Do not issue an ISSTART instruction if CSP is already running.
Operation:	PC := 0, start execution

#### Opcode: 0xE1

7	6	5	4	3	2	1	0
1	1	1	0	0	0	0	1

#### 23.14.9.34 ISRXON

Function:	Enable and calibrate frequency synthesizer for RX
Description:	The ISRXON instruction immediately enables and calibrates the frequency synthesizer for RX.
Operation:	SRXON

Opcode: 0xE3

7	6	5	4	3	2	1	0
1	1	1	0	0	0	1	1

#### 23.14.9.35 ISRXMASKBITSET

 Function:
 Set bit in RXENABLE

 Description:
 The ISRXMASKBITSET instruction immediately sets bit 5 in the RXENABLE register.

 Operation:
 SRXMASKBITSET

Opcode: 0xE4

7	6	5	4	3	2	1	0
1	1	1	0	0	1	0	0

### 23.14.9.36 ISRXMASKBITCLR

Function:	Clear bit in RXENABLE
Description:	The ISRXMASKBITCLR instruction immediately clears bit 5 in the RXENABLE register.
Operation:	SRXMASKBITCLR

#### Opcode: 0xE5

-								
7	6	5	4	3	2	1	0	
1	1	1	0	0	1	0	1	

### 23.14.9.37 ISTXON

Function:	Enable TX after calibration
Description:	The ISTXON instruction immediately enables TX after calibration. The instruction waits for the radio to acknowledge the command before executing the next instruction.
Operation:	STXON_STRB



Command Strobe/CSMA-CA Processor

0	ncode	: 0xE9
	peoue	. 0.1.3

7	6	5	4	3	2	1	0
1	1	1	0	1	0	0	1

### 23.14.9.38 ISTXONCCA

Function:	Enable calibration and TX if CCA indicates a clear channel
Description:	The ISTXONCCA instruction immediately enables TX after calibration if CCA indicates a clear channel.
Operation:	STXONCCA

#### Opcode: 0xEA

7	6	5	4	3	2	1	0
1	1	1	0	1	0	1	0

### 23.14.9.39 ISSAMPLECCA

Function:	Sample the current CCA value to SAMPLED_CCA
Description:	The current CCA value is immediately written to ${\scriptstyle\tt SAMPLED_CCA}$ in XREG.
Operation:	SSAMPLECCA

#### Opcode: 0xEB

7	6	5	4	3	2	1	0
1	1	1	0	1	0	1	1

### 23.14.9.40 ISRFOFF

Function:Disable RX/TX and frequency synthesizer.Description:The ISRFOFF instruction immediately disables RX/TX and the frequency synthesizer.Operation:FFCTL_SRFOFF_STRB = 1

#### Opcode: 0xEF

7	6	5	4	3	2	1	0
1	1	1	0	1	1	1	1

### 23.14.9.41 ISFLUSHRX

Function:	Flush RXFIFO buffer and reset demodulator
•	The ISFLUSHRX instruction immediately flushes the RXFIFO buffer and resets the demodulator.
Operation:	SFLUSHRX

#### Opcode: 0xED

7	6	5	4	3	2	1	0
1	1	1	0	1	1	0	1



Command Strobe/CSMA-CA Processor

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### 23.14.9.42 ISFLUSHTX

Function:	Flush TXFIFO buffer
Description:	The ISFLUSHTX instruction immediately flushes the TXFIFO buffer.
Operation:	SFLUSHTX

Opcode: 0xEE

•							
7	6	5	4	3	2	1	0
1	1	1	0	1	1	1	0

### 23.14.9.43 ISACK

Function:	Send acknowledge frame with the pending field cleared
Description:	The ISACK instruction immediately sends an acknowledge frame.
Operation:	SACK

Opcode: 0xE6

7	6	5	4	3	2	1	0
1	1	1	0	0	1	1	0

### 23.14.9.44 ISACKPEND

Function:	Send acknowledge frame with the pending field set
Description:	The ISACKPEND instruction immediately sends an acknowledge frame with the pending field set. The instruction waits for the radio to receive and interpret the command before executing the next instruction.
Operation:	SACKPEND

#### Opcode: 0xE7

7	6	5	4	3	2	1	0
1	1	1	0	0	1	1	1

#### 23.14.9.45 ISNACK

Function:	Abort sending of acknowledge frame
Description:	The ISNACK instruction immediately prevents sending of an acknowledge frame to the currently received frame.
Operation:	SNACK

#### Opcode: 0xE8

7	6	5	4	3	2	1	0
1	1	1	0	1	0	0	0

### 23.14.9.46 ISCLEAR



Function:	Clear CSP program memory, reset program counter
Description:	The ISCLEAR clears the program memory, resets the program counter, and aborts any running program. No stop interrupt is generated. The LABEL pointer is cleared. The ISCLEAR instruction must be issued twice to reset the program counter.
Operation:	PC := 0, clear program memory

#### Opcode: 0xFF

7	6	5	4	3	2	1	0
1	1	1	1	1	1	1	1

### 23.15 Registers

### Table 23-5. Register Overview

Address (Hex)	+ 0x000	+ 0x001	+ 0x002	+ 0x003
0x6180	FRMFILT0	FRMFILT1	SRCMATCH	SRCSHORTEN0
0x6184	SRCSHORTEN1	SRCSHORTEN2	SRCEXTEN0	SRCEXTEN1
0x6188	SRCEXTEN2	FRMCTRL0	FRMCTRL1	RXENABLE
0x618C	RXMASKSET	RXMASKCLR	FREQTUNE	FREQCTRL
0x6190	TXPOWER	TXCTRL	FSMSTAT0	FSMSTAT1
0x6194	FIFOPCTRL	FSMCTRL	CCACTRL0	CCACTRL1
0x6198	RSSI	RSSISTAT	RXFIRST	RXFIFOCNT
0x619C	TXFIFOCNT	RXFIRST_PTR	RXLAST_PTR	RXP1_PTR
0x61A0		TXFIRST_PTR	TXLAST_PTR	RFIRQM0
0x61A4	RFIRQM1	RFERRM	MONMUX	RFRND
0x61A8	MDMCTRL0	MDMCTRL1	FREQEST	RXCTRL
0x61AC	FSCTRL		FSCAL1	FSCAL2
0x61B0	FSCAL3	AGCCTRL0	AGCCTRL1	AGCCTRL2
0x61B4	AGCCTRL3	ADCTEST0	ADCTEST1	ADCTEST2
0x61B8	MDMTEST0	MDMTEST1	DACTEST0	DACTEST1
0x61BC	DACTEST2	ATEST	PTEST0	PTEST1
0x61C0	CSPPROG0	CSPPROG1	CSPPROG2	CSPPROG3
0x61C4	CSPPROG4	CSPPROG5	CSPPROG6	CSPPROG7
0x61C8	CSPPROG8	CSPPROG9	CSPPROG10	CSPPROG11
0x61CC	CSPPROG12	CSPPROG13	CSPPROG14	CSPPROG15
0x61D0	CSPPROG16	CSPPROG17	CSPPROG18	CSPPROG19
0x61D4	CSPPROG20	CSPPROG21	CSPPROG22	CSPPROG23
0x61D8				
0x61DC				
0x61E0	CSPCTRL	CSPSTAT	CSPX	CSPY
0x61E4	CSPZ	CSPT		
0x61E8				RFC_OBS_CTRL0
0x61EC	RFC_OBS_CTRL1	RFC_OBS_CTRL2		
0x61F0				
0x61F4				
0x61F8			TXFILTCFG	



### 23.15.1 Register Settings Update

This section contains a summary of the register settings that must be updated from their default value to have optimal performance.

The following settings should be set for both RX and TX. Although not all settings are necessary for both RX and TX, it is recommended for simplicity (allowing one set of settings to be written at the initialization of the code).

#### Table 23-6. Registers That Require Update From Their Default Value

Register Name	New Value (Hex)	Description
AGCCTRL1	0x15	Adjusts AGC target value.
TXFILTCFG	0x09	Sets TX anti-aliasing filter to appropriate bandwidth.
FSCAL1	0x00	Recommended setting for lowest spurious emission

### 23.15.2 Register Access Modes

The *Mode* column in Table 23-7 shows what kind of accesses are allowed for each bit. The *Description* column gives the meaning of the different alternatives.

Table 23-7. Register-Bit Access Mo	odes
------------------------------------	------

Mode	Description
R	Read
W	Write
R0	Read constant zero
R1	Read constant one
W1	Only possible to write one
W0	Only possible to write zero
R*	The value read is not the actual register value, but rather the value seen by the module. This is typically used where a configuration value may be generated automatically (through calibration, dynamic control etc.) or manually overridden with a register value. An example structure is shown for the AGCCTRL2 register in Figure 23-24.

read_data

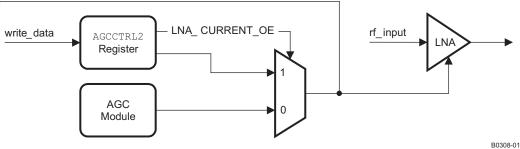


Figure 23-24. Example Hardware Structure for the R* Register Access Mode

## 23.15.3 Register Descriptions

Bit No.	Name	Reset	R/W	Description
7	-	0	R/W	Reserved. Always write 0
6:4	<pre>FCF_RESERVED_MASK[2:0]</pre>	000	R/W	Used for filtering on the reserved part of the frame control field (FCF). FCF_RESERVED_MASK[2:0] is ANDed with FCF[9:7]. If the result is nonzero and frame filtering is enabled, the frame is rejected.
3:2	MAX_FRAME_VERSION[1:0]	11	R/W	Used for filtering on the frame version field of the frame control field (FCF).
				If FCF[13:12] (the frame version subfield) is higher than MAX_FRAME_VERSION[1:0] and frame filtering is enabled, the frame is rejected.
1	PAN_COORDINATOR	0	R/W	Should be set high when the device is a PAN coordinator, to accept frames with no destination address (as specified in section 7.5.6.2 of IEEE 802.15.4(b))
				0: Device is not PAN coordinator.
				1: Device is PAN coordinator.
0	FRAME_FILTER_EN	1	R/W	Enables frame filtering
				When this bit is set, the radio performs frame filtering as specified in section 7.5.6.2 of IEEE 802.15.4(b), third filtering level. FRMFILT0[6:1] and FRMFILT1[7:1], together with the local address information, define the behavior of the filtering algorithm.
				0: Frame filtering off. (FRMFILT0[6:1], FRMFILT1[7:1] and SRCMATCH[2:0] are don't care.)
				1: Frame filtering on.

#### FRMFILT1 (0x6181) - Frame Filtering

Bit No.	Name	Reset	R/W	Description
7	ACCEPT_FT_4T07_RESERVED	0	R/W	Defines whether reserved frames are accepted or not. Reserved frames have frame type = 100, 101, 110, or 111.
				0: Reject 1: Accept
6	ACCEPT_FT_3_MAC_CMD	1	R/W	Defines whether MAC command frames are accepted or not. MAC command frames have frame type = 011.
				0: Reject 1: Accept
5	ACCEPT_FT_2_ACK	1	R/W	Defines whether acknowledgment frames are accepted or not. Acknowledgement frames have frame type = 010.
				0: Reject 1: Accept
4	ACCEPT_FT_1_DATA	1	R/W	Defines whether data frames are accepted or not. Data frames have frame type = 001.
				0: Reject 1: Accept
3	ACCEPT_FT_0_BEACON	1	R/W	Defines whether beacon frames are accepted or not. Beacon frames have frame type = 000.
				0: Reject 1: Accept
2:1	MODIFY_FT_FILTER[1:0]	00	R/W	These bits are used to modify the frame-type field of a received frame before frame-type filtering is performed. The modification does not influence the frame that is written to the RXFIFO.
				00: Leave as it is 01: Invert MSB 10: Set MSB to 0 11: Set MSB to 1
0	_	0	R/W	Reserved. Always write 0

Bit No.	Name	Reset	R/W	Description
7:3	-	0000 0	R/W	Reserved. Always write 0
2	PEND_DATAREQ_ONLY	1	R/W	When this bit is set, the AUTOPEND function also requires that the received frame is a <i>DATA REQUEST</i> MAC command frame.
1	AUTOPEND	1	R/W	Automatic acknowledgment pending flag enable.
				On reception of a frame, the pending bit in the (possibly) returned acknowledgment is set automatically, given that:
				<ul> <li>FRMFILT0.FRAME_FILTER_EN is set.</li> </ul>
				- SRCMATCH.SRC_MATCH_EN is set.
				– SRCMATCH.AUTOPEND is set.
				<ul> <li>The received frame matches the current SRCMATCH.PEND_DATAREQ_ONLY setting.</li> </ul>
				<ul> <li>The received source address matches at least one source-match table entry, which is enabled in both SHORT_ADDR_EN and SHORT_PEND_EN or EXT_ADDR_EN and EXT_PEND_EN.</li> </ul>
				Note: Details for SHORT_PEND_EN and EXT_PEND_EN is found in the memory map description (Section 23.4).
0	SRC_MATCH_EN	1	R/W	Source address matching enable (This bit is don't care if FRMFILT0.FRAME_FILTER_EN = 0.)

#### SRCSHORTEN0 (0x6183) – Short Address Matching

Bit No.	Name	Reset	R/W	Description
7:0	SHORT_ADDR_EN[7:0]	0x00	R/W	The 7:0 part of the 24-bit word SHORT_ADDR_EN that enables/disables source address matching for each of the 24 short address table entries. Optional safety feature: To ensure that an entry in the source-matching table is not used while it is being updated, set the corresponding SHORT_ADDR_EN bit to 0 while updating.

### SRCSHORTEN1 (0x6184) – Short Address Matching

Bit No.	Name	Reset	R/W	Description
7:0	SHORT_ADDR_EN[15:8]	0x00	R/W	The 15:8 part of the 24-bit word SHORT_ADDR_EN. See previous description of SRCSHORTEN0.

#### SRCSHORTEN2 (0x6185) – Short Address Matching

Bit No.	Name	Reset	R/W	Description
7:0	SHORT_ADDR_EN[23:16]	0x00	R/W	The 23:16 part of the 24-bit word SHORT_ADDR_EN. See previous description of SRCSHORTEN0.

#### SRCEXTEN0 (0x6186) - Extended Address Matching

Bit No.	Name	Reset	R/W	Description
7:0	EXT_ADDR_EN[7:0]	0x00	R/W	The 7:0 part of the 24-bit word EXT_ADDR_EN that enables/disables source address matching for each of the 12 extended address table entries.
				Write access: Extended address enable for table entry $n$ (0 to 7) is mapped to EXT_ADDR_EN[2n]. All EXT_ADDR_EN[2n + 1] bits are read-only and don't care when written to.
				Read access: Extended address enable for table entry <i>n</i> (0 to 7) is mapped to both EXT_ADDR_EN[2n] and EXT_ADDR_EN[2n + 1].
				To ensure that an entry in the source matching table is not used while it is being updated, set the corresponding EXT_ADDR_EN bit to 0 while updating.

#### Registers

SRCE	SRCEXTEN1 (0x6187) – Extended Address Matching								
Bit No.	Name	Reset	R/W	Description					
7:0	EXT_ADDR_EN[15:8]	0x00	R/W	The 15:8 part of the 24-bit word EXT_ADDR_EN. See previous description of SRCEXTEN0.					

SRCEX	SRCEXTEN2 (0x6188) – Extended Address Matching								
Bit	Name	Reset	R/W	Description					
No.									
7:0	EXT_ADDR_EN[23:16]	0x00	R/W	The 23:16 part of the 24-bit word EXT_ADDR_EN. See previous description of SRCEXTEN0.					

### FRMCTRL0 (0x6189) – Frame Handling

6         AUTOCRC         1         R/W         In TX           6         AUTOCRC         1         R/W         In TX           1         A CRC-16 (ITU-T) is generated in hardware and appended to transmitted frame. There is no need to write the last 2 bytes of the must be generated manually and written to TXBUF (if not, TX_BUF.)           0         No CRC-16 is appended to the frame. The last 2 bytes of the must be generated manually and written to TXBUF (if not, TX_UNDERFLOW occurs).           In RX         1:         The CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The status word is controllable through APPEND_DATA_MODE.           5         AUTOACK         0         R/W         Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addred filtering, have the acknowledge 12 symbol periods after being rece 0: Autoack isabled           4         ENERGY_SCAN         0         R/W         Defines whether the rasis 1 register contains the most-recent signal strength or the peak signal strength           3:2         Rx_MODE[1:0]         00         R/W         Set RX modes 0: Normal operation, use RXFIFO.           1:         Tx_MODE[1:0]         00         R/W         Set RX modes 0: Normal operation except that symbol search is disabled 1: Reserved 10: RXFIFO looping ignores overflow in RXFIFO; infinite receptor 11: Same as normal operation except that symbol search is disab can be used for RSSI or CCA mea	Bit No.	Name	Reset	R/W	Description
0:         RSSI + The CRC_OK bit and the 7-bit correlation value are appended at the end of each received frame. See Table 23-1 details.           6         AUTOCRC         1         RW         In TX           1:         A CRC-16 (ITU-T) is generated in hardware and appended to the frame. There is no need to write the last 2 bytes to transmitted frame. There is no need to write the last 2 bytes to TABUF.           6         AUTOCRC         1         RW         In TX           1:         A CRC-16 (ITU-T) is generated in hardware and appended to the frame. The last 2 bytes of the must be generated manually and written to TXBUF (ff not, TX, UNDERFLOW occurs).           0:         No CRC-16 is checked in hardware, and replaced in the RXF UNDERFLOW occurs).           1:         The CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The stat word is controliable through ADDE.           5         AUTOACK         0         RW         Defines whether the radio automaking and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are automaking and work and have a valid CI are autoack disabled           4 <td>7</td> <td>APPEND_DATA_MODE</td> <td>0</td> <td>R/W</td> <td>When AUTOCRC = 0: Don't care</td>	7	APPEND_DATA_MODE	0	R/W	When AUTOCRC = 0: Don't care
6         AUTOCRC         1         RW         In TX           6         AUTOCRC         1         RW         In TX           1:         RSSI + The CRC_OK bit and the 7-bit SRCRESINDEX are appended at the end of each received frame. See Table 23-1 details.           6         AUTOCRC         1         RW         In TX           1:         A CRC-16 (ITU-T) is generated in hardware and appended to transmitted frame. There is no need to write the last 2 bytes of the must be generated manually and writen to TXBUF.         0: No CRC-16 is checked in hardware, and replaced in the RXF TXUUPERFLOW occurs).           1:         The CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word is contailisbit through a PENTUD_DATU.     <					When AUTOCRC = 1:
6         AUTOCRC         1         RW         In TX           6         AUTOCRC         1         RW         In TX           1:         A CRC-16 (ITU-T) is generated in hardware and appended to transmitted frame. There is no need to write the last 2 bytes to TXBUF.         0:         No CRC-16 is appended to the frame. There is no need to write the last 2 bytes of the must be generated manually and written to TXBUF (if not, TX_UNDERFLOW occurs).           In RX         1:         The CRC-16 is chacked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The stat word is controllable through APPEND_DATA_WODE.           5         AUTOACK         0         RW         Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addin filtering, have the acknowledge frage st, and have a valid C are automatically acknowledged 12 symbol periods after being rece 0:           4         ENERGY_SCAN         0         RW         Defines whether the RSSI register contains the most-recent signal strength ince the energy scan was enal 0:           3:2         RX_MODE[1:0]         00         RW         Set RX modes           0:         Nort necent signal strength         1:         Peak signal strength           3:2         RX_MODE[1:0]         00         RW         Set RX modes         00:           0:         NortHecent signal strength         1:					
1:       A CRC-16 (ITU-T) is generated in hardware and appended to transmitted frame. There is no need to write the last 2 bytes to TXBUF.         0:       No CRC-16 is appended to the frame. The last 2 bytes of the must be generated manually and written to TXBUF (if not, TX_UNDERLOW occurs).         10:       N CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The stat word is controllable through APPEND_DATA_MODE.         5       AUTOACK       0         6       No CRC-16 is appended to the frame (CRC-16 field) are stored in the RXF ipO. The CRC check (if any) must be done manually.         5       AUTOACK       0         6       RW       Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addref filtering, have the acknowledged 12 symbol periods after being rece 0:         6       AUTOACK       0       R/W         9       Reserved       0       <					appended at the end of each received frame. See Table 23-1 for
4         ENERGY_SCAN         0         RW         Defines whether the rabled at the assignal strength           3:2         RX_MODE[1:0]         00         RW         Set RX modes           3:2         RX_MODE[1:0]         00         RW         Set RX modes           1:0         TX_MODE[1:0]         00         RW         Set RX modes           1:0         TX_MODE[1:0]         00         RW         Set test modes for TX           0:1:0         TX_MODE[1:0]         00         RW         Set test modes for TX           0:1:0         TX_MODE[1:0]         00         RW         Set test modes for TX           0:1:0         TX_MODE[1:0]         00         RW         Set test modes for TX           0:1:0         TX_MODE[1:0]         00         RW         Set test modes for TX	6	AUTOCRC	1	R/W	In TX
must be generated manually and written to TXBUF (if not, TX_UNDERFLOW occurs).           In RX           1: The CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which contains a CRC OK bit. The status word which appendix the status acknowledge manually. Note that this setting does not influence acknowledgement transmiss (including AUTOACK).           5         AUTOACK         0         R/W         Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addred filtering, have the acknowledge request flag set, and have a valid C are automatically acknowledge request flag set, and have a valid C are automatically acknowledge 12 symbol periods after being rece 0: Autoack disabled           4         ENERGY_SCAN         0         R/W         Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength           3:2         RX_MODE[1:0]         00         R/W         Set RX modes 00: Normal operation, use RXFIFO. 01: Reserved 10: RXFIFO looping ignores overflow in RXFIFO; infinite reception 11: Same as normal operation, recept that symbol search is disab Can be used for RSSI or CCA measurements when it is not d to find symbol.           1:0					transmitted frame. There is no need to write the last 2 bytes to
1:         The CRC-16 is checked in hardware, and replaced in the RXF by a 16-bit status word which contains a CRC OK bit. The stat word is controllable through APFEND_DATA_MODE.           0:         The last two bytes of the frame (CRC-16 field) are stored in th RXFIFO. The CRC check (if any) must be done manually. Note that this setting does not influence acknowledgment transmiss (including AUTOACK).           5         AUTOACK         0         R/W         Defines whether the radio automatically transmits acknowledge fran not. When autoack is enabled, all frames that are accepted by addin filtering, have the acknowledge request flag set, and have a valid C are automatically acknowledge 12 symbol periods after being rece 0:           4         ENERGY_SCAN         0         R/W         Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength 1:         Public most-recent signal strength 1:           3:2         RX_MODE[1:0]         00         R/W         Defines overflow in RXFIFO. 01:           1:0         TX_MODE[1:0]         00         R/W         Set RX modes 0:         Normal operation, use RXFIFO. 01:           1:0         TX_MODE[1:0]         00         R/W         Set test modes for TX 0:         Normal operation, transmit TXFIFO 01:           1:0         TX_MODE[1:0]         00         R/W         Set test modes for TX 0:         Normal operation, transmit TXFIFO					must be generated manually and written to TXBUF (if not,
by a 16-bit status word which contains a CRC OK bit. The statword is controllable through APPEND_DATA_MODE.         0: The last two bytes of the frame (CRC-16 field) are stored in th RXFIFO. The CRC check (if any) must be done manually. Note that this setting does not influence acknowledgement transmiss (including AUTOACK).           5         AUTOACK         0         R/W         Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addin filtering, have the acknowledge request flag set, and have a valid C are automatically acknowledge to 2 symbol periods after being rece           4         ENERGY_SCAN         0         R/W         Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength since the energy scan was enal 0: Most-recent signal strength           3:2         RX_MODE[1:0]         00         R/W         Set RX modes           00:         Normal operation, use RXFIFO.         01: Reserved           10:         TX_MODE[1:0]         00         R/W         Set RX modes           00:         Normal operation, use RXFIFO.         01: RSERVED         01: Reserved           10:         TX_MODE[1:0]         00         R/W         Set test modes for TX           00:         NORM         Set test modes for TX         02: Normal operation, transmit TXFIFO           11:0         TX_MODE[1:0]         00         R/W         Set test modes for TX					In RX
RXFIFO. The CRC check (if any) must be done manually. Note that this setting does not influence acknowledgment transmiss (including AUTOACK         AUTOACK       0       R/W       Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addr filtering, have the acknowledge request flag set, and have a valid C are automatically acknowledged 12 symbol periods after being rece 0: Autoack disabled 1: Autoack enabled         4       ENERGY_SCAN       0       R/W       Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength 1: Peak signal strength         3:2       RX_MODE[1:0]       00       R/W       Set RX modes 00: Normal operation, use RXFIFO. 01: Reserved 10: RXFIFO looping ignores overflow in RXFIFO; infinite receptior 11: Same as normal operation except that symbol search is disab Can be used for RSSI or CCA measurements when it is not d to find symbol.         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX 00: Normal operation, transmit TXFIFO 01: Reserved.         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX 00: Normal operation, transmit TXFIFO					by a 16-bit status word which contains a CRC OK bit. The status
including AUTOACK.       including AUTOACK).         5       AUTOACK       0       R/W       Defines whether the radio automatically transmits acknowledge fram not. When autoack is enabled, all frames that are accepted by addra filtering, have the acknowledge request flag set, and have a valid C         4       ENERGY_SCAN       0       R/W       Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength since the energy scan was enal 0: Most-recent signal strength         3:2       RX_MODE[1:0]       00       R/W       Set RX modes         00:       NOW       Set RX modes         01:       Reserved       10: RXFIFO looping ignores overflow in RXFIFO; infinite receptior 11: Same as normal operation except that symbol search is disab C and be used for RSSI or CCA measurements when it is not d to find symbol.         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX 00: Normal operation, transmit TXFIFO ond reads cyclic					
Image: Second					Note that this setting does not influence acknowledgment transmission (including AUTOACK).
1:       Autoack enabled         4       ENERGY_SCAN       0         8:2       RX_MODE[1:0]       00         8:3       RX_MODE[1:0]       00         8:4       RX_MODE[1:0]       00         8:5       RX_MODE[1:0]       00         8:6       R/W       Set RX modes         00:       Normal operation, use RXFIFO.         01:       Reserved         10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         11:       Same as normal operation except that symbol search is disab         Can be used for RSSI or CCA measurements when it is not due to find symbol.         1:0       TX_MODE[1:0]       00         8:W       Set test modes for TX         00:       Normal operation, transmit TXFIFO         01:       Reserved. Should not be used         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclid	5	AUTOACK	0	R/W	Defines whether the radio automatically transmits acknowledge frames or not. When autoack is enabled, all frames that are accepted by address filtering, have the acknowledge request flag set, and have a valid CRC are automatically acknowledged 12 symbol periods after being received.
4       ENERGY_SCAN       0       R/W       Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength since the energy scan was enal 0: Most-recent signal strength         3:2       RX_MODE[1:0]       00       R/W       Set RX modes 00: Normal operation, use RXFIFO. 01: Reserved 10: RXFIFO looping ignores overflow in RXFIFO; infinite receptior 11: Same as normal operation except that symbol search is disab Can be used for RSSI or CCA measurements when it is not due to find symbol.         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX 00: Normal operation, transmit TXFIFO and reads cyclic 10: RXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic 10: TXFIFO looping ignores underflow in TXFIFO looping ignores underflow in TXFIFO looping ignores unde					0: Autoack disabled
3:2       RX_MODE[1:0]       00       R/W       Set RX modes       00: Normal operation, use RXFIFO.         01:       Reserved       10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX         00:       TX_MODE[1:0]       00       R/W       Set reserved         1:0       TX_MODE[1:0]       00       R/W       Set reserved for RSSI or CCA measurements when it is not do to find symbol.         1:0       TX_MODE[1:0]       00       R/W       Set reserved for RSSI or TX         0:       Normal operation, transmit TXFIFO       01: Reserved.       02: Normal operation, transmit TXFIFO         1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX         0:       Normal operation, transmit TXFIFO       01: Reserved.       10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic					1: Autoack enabled
3:2       RX_MODE[1:0]       00       R/W       Set RX modes         00:       Normal operation, use RXFIFO.       01:       Reserved         10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         11:       Same as normal operation except that symbol search is disab         Can be used for RSSI or CCA measurements when it is not de to find symbol.         11:0       TX_MODE[1:0]         00       R/W         Set test modes for TX         00:       Normal operation, transmit TXFIFO         01:       Reserved. Should not be used         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic	4	ENERGY_SCAN	0	R/W	Defines whether the RSSI register contains the most-recent signal strength or the peak signal strength since the energy scan was enabled.
3:2       RX_MODE[1:0]       00       R/W       Set RX modes         00:       Normal operation, use RXFIFO.       01:       Reserved         10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         11:       Same as normal operation except that symbol search is disab         Can be used for RSSI or CCA measurements when it is not de to find symbol.         1:0       TX_MODE[1:0]         00       R/W         Set test modes for TX         00:       Normal operation, transmit TXFIFO         01:       Reserved. Should not be used         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic					0: Most-recent signal strength
00:       Normal operation, use RXFIFO.         01:       Reserved         10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         11:       Same as normal operation except that symbol search is disab Can be used for RSSI or CCA measurements when it is not du to find symbol.         11:0       TX_MODE[1:0]         00       R/W         Set test modes for TX 00:         01:       Reserved. Should not be used 10:         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclid					1: Peak signal strength
1:0       TX_MODE[1:0]       00       R/W       Set test modes for TX         00:       Normal operation, transmit TXFIFO       00:       Normal operation, transmit TXFIFO         10:       TX_MODE[1:0]       00       R/W       Set test modes for TX         01:       Reserved.       Should not be used       10:         10:       TX_MODE[1:0]       00       R/W	3:2	RX_MODE[1:0]	00	R/W	Set RX modes
10:       RXFIFO looping ignores overflow in RXFIFO; infinite reception         11:       Same as normal operation except that symbol search is disab Can be used for RSSI or CCA measurements when it is not du to find symbol.         11:0       TX_MODE[1:0]         00       R/W         Set test modes for TX 00:         00:       Normal operation, transmit TXFIFO 01:         01:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic					
11:       Same as normal operation except that symbol search is disab Can be used for RSSI or CCA measurements when it is not do to find symbol.         1:0       TX_MODE[1:0]         00       R/W         Set test modes for TX 00:         01:       Reserved. Should not be used 10:         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic					
I:0       TX_MODE[1:0]       00       R/W       Set test modes for TX         00:       Normal operation, transmit TXFIFO       01:       Reserved. Should not be used         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic					
00:       Normal operation, transmit TXFIFO         01:       Reserved. Should not be used         10:       TXFIFO looping ignores underflow in TXFIFO and reads cyclic					Can be used for RSSI or CCA measurements when it is not desired
01: Reserved. Should not be used 10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic	1:0	TX_MODE[1:0]	00	R/W	
10: TXFIFO looping ignores underflow in TXFIFO and reads cyclic					00: Normal operation, transmit TXFIFO
					01: Reserved. Should not be used
					1 0 0
11: Send pseudorandom data from CRC, infinite transmission.					11: Send pseudorandom data from CRC, infinite transmission.

### FRMCTRL1 (0x618A) – Frame Handling

Bit No.	Name	Reset	R/W	Description
7:3	-	0000 0	R0	Read as zero
2			R/W	Defines whether the pending data bit in outgoing acknowledgment frames is always set to 1 or controlled by the main FSM and the address filtering.
				0: Pending data bit is controlled by main FSM and address filtering.
				1: Pending data bit is always 1.
1	IGNORE_TX_UNDERF	0	R/W	Defines whether TX underflow should be ignored or not.
				0: Normal TX operation. TX underflow is detected and TX is aborted if underflow occurs.
				1: Ignore TX underflow. Transmit the number of bytes given by the frame-length field.
0	SET_RXENMASK_ON_TX	1	R/W	Defines whether STXON sets bit 6 in the RXENABLE register or leaves it unchanged.
				0: Does not affect RXENABLE.
				1: Sets bit 6 in RXENABLE. Used for backwards compatibility with the CC2420.

### RXENABLE (0x618B) – RX Enabling

Bit No.	Name	Reset	R/W	Description
7:0	RXENMASK[7:0]	0x00	R	RXENABLE enables the receiver. A nonzero value in this register causes the main FSM to enable the receiver when in idle, after transmission, and after acknowledgement transmission.
				The following strobes can modify RXENMASK:
				SRXON: Sets bit 7 in RXENMASK
				STXON: Sets bit 6 in RXENMASK if SET_RXENMASK_ON_TX = 1
				SRFOFF: Clears all bits in RXENMASK
				SRXMASKBITSET: Sets bit 5 in RXENMASK
				SRXMASKBITCLR: Clears bit 5 in RXENMASK
				RXENABLE can be modified directly by the CPU by accessing registers RXMASKSET and RXMASKCLR.
				There might be conflicts between the CSP and CPU operations if both try to modify RXENMASK simultaneously. To handle the case of simultaneous access to RXENMASK, the following rules apply:
				<ul> <li>If two sources are not in conflict (they modify different parts of the register), both their requests to modify RXENMASK are processed.</li> </ul>
				<ul> <li>If both try to modify the mask simultaneously, bus-write operations to RXMASKSET and RXMASKCLR have priority over the CSP. It is strongly recommended to avoid this situation.</li> </ul>

#### RXMASKSET (0x618C) – RX Enabling

Bit No.	Name	Reset	R/W	Description
7:0	RXENMASKSET[7:0]	0x00	R0/W	When written, the written data is ORed with RXENMASK and stored in RXENMASK.

### RXMASKCLR (0x618D) – RX Disabling

Bit No.	Name	Reset	R/W	Description
7:0	RXENMASKCLR[7:0]	0x00	R0/W	When written, the written data is inverted and ANDed with RXENMASK and stored in RXENMASK.
				For example, if a 1 is written to one or more bit positions in this register, the corresponding bits are cleared in RXENMASK.



### Registers

### RFIRQM0 (0x61A3) – RF Interrupt Masks

Bit No.	Name	Reset	R/W	Description
7:0	RFIRQM[7:0]	0x00	R/W	Bit mask masking out interrupt sources
				Bit position
				7: RXMASKZERO
				6: RXPKTDONE
				5: FRAME_ACCEPTED
				4: SRC_MATCH_FOUND
				3: SRC_MATCH_DONE
				2: FIFOP
				1: SFD
				0: ACT_UNUSED

#### RFIRQM1 (0x61A4) – RF Interrupt Masks

Bit No.	Name	Reset	R/W	Description	
7:0	RFIRQM[14:8]	0x00	R/W	Bit mask masking out interrupt sources	
				Bit position	
				7: Reserved	
				6: Reserved	
				5: CSP_WAIT	
				4: CSP_STOP	
				3: CSP_MANINT	
				2: RF_IDLE	
				1: TXDONE	
				0: TXACKDONE	

#### RFERRM (0x61A5) - RF Error Interrupt Mask

Bit No.	Name	Reset	R/W	Description	
7:0	RFERRM[7:0]	0x00	R/W	Bit mask masking out interrupt sources	
				Bit position	
				7: Reserved	
				6: STROBE_ERR	
				5: TXUNDERF	
				4: TXOVERF	
				3: RXUNDERF	
				2: RXOVERF	
				1: RXABO	
				0: NLOCK	

### FREQCTRL (0x618F) – Controls the RF Frequency

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Read as zero
6:0	FREQ[6:0]	0x0B (2405 MHz)	R/W	Frequency control word $f_{RF} = f_{LO} = (2394 + FREQ[6:0]) MHz$
				The frequency word in $FREQ[6:0]$ is an offset value from 2394. The device supports the frequency range from 2394 MHz to 2507 MHz. The usable settings for $FREQ[6:0]$ are consequently 0 to 113. Settings outside this range (114–127) give a frequency of 2507 MHz.
				IEEE 802.15.4-2006 specifies a frequency range from 2405 MHz to 2480 MHz with 16 channels 5 MHz apart. The channels are numbered 11 through 26. For an IEEE 802.15.4-2006 compliant system, the only valid settings are thus $FREQ[6:0] = 11 + 5$ (channel number – 11).



### FREQTUNE (0x618E) – Crystal Oscillator Frequency Tuning

-			- J	
Bit No.	Name	Reset	R/W	Description
7:4	-	0x0	R0	Read as zero
3:0	XOSC32M_TUNE[3:0]	0xF	R/W	Tune crystal oscillator
				The default setting, 1111, leaves the XOSC not tuned. Changing the setting from default switches in extra capacitance to the oscillator, effectively lowering the XOSC frequency. Hence, a higher setting gives a higher frequency.

#### TXPOWER (0x6190) – Controls the Output Power

Bit No.	Name		Reset	R/W	Description
7:0	PA_POWER	[7:0]	0xF5	R/W	PA power control. <b>NOTE:</b> Before going to TX, this value should be updated. Please consult the device's data sheet (Appendix C) for recommended values; see also Section 23.8.13.

#### TXCTRL (0x6191) – Controls the TX Settings

	· · · · ·			
Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6:4	DAC_CURR[2:0]	110	R/W	Change the current in the DAC.
3:2	DAC_DC[1:0]	10	R/W	Adjusts the dc level to the TX mixer.
1:0	TXMIX_CURRENT[1:0]	01	R/W	Transmit mixers core current: current increases with increasing setting.

### FSMSTAT0 (0x6192) - Radio Status Register

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6	CAL_RUNNING	0	R	<ul> <li>Frequency synthesizer calibration status</li> <li>0: Calibration done or not started</li> <li>1: Calibration in progress</li> </ul>
5:0	FSM_FFCTRL_STATE[5:0]	-	R	Gives the current state of the FIFO and frame-control (FFCTRL) finite-state machine.



#### Registers

Bit	STAT1 (0x6193) – Radio Name	Reset	R/W	Description
No.	Name	Reset	N/W	Description
7	FIFO	0	R	FIFO is high whenever there is data in the RXFIFO. Low during RXFIFO overflow
6	FIFOP	0	R	FIFOP is set high when there are more than FIFOP_THR bytes of data in the RXFIFO that have passed frame filtering.
				FIFOP is set high when there is at least one complete frame in the RXFIFO. FIFOP is set low again when a byte is read from the RXFIFO and this leaves FIFOP_THR bytes in the FIFO.
				FIFOP is high during RXFIFO overflow.
5	SFD	0	R	In TX
				0: When a complete frame with SFD has been sent or no SFD has been sent
				1: SFD has been sent.
				In RX
				0: When a complete frame has been received or no SFD has been received
				1: SFD has been received.
4	CCA	0	R	Clear-channel assessment. Dependent on CCA_MODE settings. See the following description of CCACTRL1.
3	SAMPLED_CCA	0	R	Contains a sampled value of the CCA. The value is updated whenever a SSAMPLECCA or STXONCCA strobe is issued.
2	LOCK_STATUS	0	R	1 when PLL is in lock, otherwise 0.
1	TX_ACTIVE	0	R	Status signal, active when FFCTRL is in one of the transmit states
0	RX_ACTIVE	0	R	Status signal, active when FFCTRL is in one of the receive states

#### FIFOPCTRL (0x6194) – FIFOP Threshold

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Read as zero
6:0	FIFOP_THR[6:0]	100 0000	R/W	Threshold used when generating FIFOP signal

#### Bit Name Reset R/W Description No. 7:2 0000 00 R0 Read as zero 1 SLOTTED_ACK 0 R/W Controls timing of transmission of acknowledge frames 0: The acknowledge frame is sent 12 symbol periods after the end of the received frame which requests the aknowledge. The acknowledge frame is sent at the first backoff-slot boundary more than 12 symbol periods after the end of the received frame 1: which requests the aknowledge. 0 R/W RX2RX_TIME_OFF 1 Defines whether or not a 12-symbol time-out should be used after frame reception has ended. 0: No time-out 1: 12-symbol-period time-out

#### FSMCTRL (0x6195) - FSM Options

### CCACTRL0 (0x6196) - CCA Threshold

CCA	- CCA Three (0x6196) - CCA Three	siloiu		
Bit No.	Name	Reset	R/W	Description
7:0	CCA_THR[7:0]	0xE0	R/W	Clear-channel-assessment threshold value, signed 2s-complement number for comparison with the RSSI.
				The unit is 1 dB, offset is about 76 dBm. The CCA signal goes high when the received signal is below this value. The CCA signal is available on the CCA pin and in the FSMSTAT1 register.
				Note that the value should never be set lower than CCA_HYST – 128 in order to avoid erroneous behavior of the CCA signal.
				<b>NOTE:</b> The reset value translates to an input level of approximately $-32 - 76 = -108$ dBm, which is well below the sensitivity limit. That means the CCA signal never indicates a clear channel.
				This register should be updated to 0xF8, which translates to an input level of about $-8 - 76 = -84$ dBm.

#### CCACTRL1 (0x6197) – Other CCA Options

Bit No.	Name	Reset	R/W	Description
7:5	-	000	R0	Read as zero
4:3	CCA_MODE[1:0]	11	R/W	00: CCA always set to 1
				01: $CCA = 1$ when RSSI < CCA_THR - CCA_HYST; $CCA = 0$ when RSSI > CCA_THR
				10: $CCA = 1$ when not receiving a frame, else $CCA = 0$
				11: CCA = 1 when RSSI < CCA_THR - CCA_HYST and not receiving a frame; CCA = 0 when RSSI ≥ CCA_THR or when receiving a frame
2:0	CCA_HYST[2:0]	010	R/W	Sets the level of CCA hysteresis. Unsigned values given in dB

#### RSSI (0x6198) - RSSI Status Register

Bit No.	Name	Reset	R/W	Description
7:0	RSSI_VAL[7:0]	0x80	R	RSSI estimate on a logarithmic scale, signed number in 2s complement Unit is 1 dB. The offset to use in order to convert the register value into the real RSSI value can be found in the device's data sheet (Appendix C). The RSSI value is averaged over 8 symbol periods. The RSSI_VALID status bit should be checked before reading RSSI_VAL the first time. The reset value of -128 also indicates that the RSSI value is invalid.

#### RSSISTAT (0x6199) - RSSI Valid Status Register

Bit No.	Name	Reset	R/W	Description
7:1	-	000 000	R0	Read as zero
0	RSSI_VALID	0	R	RSSI value is valid. Occurs eight symbol periods after entering RX

#### RXFIRST (0x619A) - First Byte in RXFIFO

Bit No.	Name	Reset	R/W	Description
7:0	DATA[7:0]	0x00		First byte of the RXFIFO. Note: Reading this register does not modify the contents of the FIFO.

#### RXFIFOCNT (0x619B) – Number of Bytes in RXFIFO

Bit No.	Name	Reset	R/W	Description
7:0	RXFIFOCNT[7:0]	0x00	R	Number of bytes in the RXFIFO. Unsigned integer.

Registers

### Registers

TXFIF	TXFIFOCNT (0x619C) – Number of Bytes in TXFIFO							
Bit No.	Name	Reset	R/W	Description				
7:0	TXFIFOCNT[7:0]	0x00	R	Number of bytes in the TXFIFO. Unsigned integer				

### RXFIRST_PTR (0x619D) - RXFIFO Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	RXFIRST_PTR[6:0]	000 0000	R	RAM address offset of the first byte in the RXFIFO

#### RXLAST_PTR (0x619E) - RXFIFO Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	RXLAST_PTR[6:0]	000 0000	R	RAM address offset of the last byte +1 byte in the RXFIFO

#### RXP1_PTR (0x619F) – RXFIFO Pointer

Bit No.	Name	Reset	R/W	Description
7:0	RXP1_PTR[7:0]	0x00	R	RAM address offset of the first byte of the first frame in the RXFIFO

#### TXFIRST_PTR (0x61A1) – TXFIFO Pointer

Bit No.	Name	Reset	R/W	Description
7:0	TXFIRST_PTR[7:0]	0x00	R	RAM address offset of the next byte to be transmitted from the TXFIFO

### TXLAST_PTR (0x61A2) – TXFIFO Pointer

Bit No.	Name	Reset	R/W	Description
7:0	TXLAST_PTR[7:0]	0x00	R	RAM address offset of the last byte +1 byte of the TXFIFO

Bit No.	Name	Reset	R/W	Description
7:6	DEM_NUM_ZEROS[1:0] 10		R/W	Sets how many zero symbols must be detected before the sync word when searching for sync. Note that only one is required to have a correlation value above the correlation threshold set in the MDMCTRL1 register.
				00: Reserved
				01: 1 zero symbol
				10: 2 zero symbols
				11: 3 zero symbols
5	DEMOD_AVG_MODE	0	R/W	Defines the behavior or the frequency offset averaging filter.
				0: Lock average level after preamble match. Restart frequency offset calibration when searching for the next frame.
				1: Continuously update average level.
4:1	PREAMBLE_LENGTH [3:0]	TH [3:0] 0010 R/W		The number of preamble bytes (two zero-symbols) to be sent in TX mode prior to the SFD, encoded in steps of 2 symbols (1 byte). The reset value of 2 is compliant with IEEE 802.15.4.
				0000: 2 leading-zero bytes
				0001: 3 leading-zero bytes
				0010: 4 leading-zero bytes
				1111: 17 leading-zero bytes
0	TX_FILTER	1	R/W	Defines the kind of TX filter that is used. The normal TX filter is as defined by the IEEE802.15.4 standard. Extra filtering may be applied in order to lower the out-of-band emissions.
				0: Normal TX filtering
				1: Enable extra filtering

#### MDMCTRL1 (0x61A9) – Controls Modem

Bit No.	Name	Reset	R/W	Description
7:6	-	00	R0	Read as zero
5	CORR_THR_SFD	0	R/W	Defines requirements for SFD detection:
				0: The correlation value of one of the zero symbols of the preamble must be above the correlation threshold.
				1: The correlation value of one zero symbol of the preamble and both symbols in the SFD must be above the correlation threshold.
4:0	CORR_THR[4:0]	0x14	R/W	Demodulator correlator threshold value, required before SFD search.
				Threshold value adjusts how the receiver synchronizes to data from the radio. If the threshold is set too low, sync can more easily be found on noise. If set too high, the sensitivity is reduced, but sync is not likely to be found on noise.
				In combination with DEM_NUM_ZEROS, the system can be tuned so sensitivity is high with less sync found on noise.

#### FREQEST (0x61AA) – Estimated RF Frequency Offset

Bit No.	Name	Reset	R/W	Description
7:0	FREQEST[7:0]	0x00	R	Signed 2s-complement value. Contains an estimate of the frequency offset between carrier and the receiver LO. The offset frequency is FREQEST × 7800 Hz. DEM_AVG_MODE controls when this estimate is updated If DEM_AVG_MODE = 0, it is updated until sync is found. Then the frequency offset estimate is frozen until the end of the received frame. If DEM_AVG_MODE = 1, it is updated as long as the demodulator is enabled. To calculate the correct value, one must use an offset (FREQEST_offset), which can be found in the device's data sheet (Appendix C). Real FREQEST value = FREQEST – FREQEST offset.



#### Registers

#### RXCTRL (0x61AB) – Tune Receive Section

NAC I	RACTAE (0x0TAB) - Tulle Receive Section							
Bit No.	Name	Reset	R/W	Description				
7:6	-	00	R0	Reserved				
5:4	GBIAS_LNA2_REF[1:0]	11	R/W	Adjusts front-end LNA2/mixer PTAT current output (from M = 3 to M = 6), default: $M = 5$				
3:2	GBIAS_LNA_REF[1:0]	11	R/W	Adjusts front-end LNA PTAT current output (from $M = 3$ to $M = 6$ ), default: $M = 5$				
1:0	MIX_CURRENT[1:0]	11	R/W	Control of the receiver mixers output current. The current increases with increasing setting.				

#### FSCTRL (0x61AC) – Tune Frequency Synthesizer

Bit No.	Name	Reset	R/W	Description
7:6	PRE_CURRENT [1:0]	01	R/W	Prescaler current setting
5:4	LODIV_BUF_CURRENT_TX [1:0]	01	R/W	Adjusts current in mixer and PA buffers. Used when TX_ACTIVE = 1
3:2	LODIV_BUF_CURRENT_RX [1:0]	10	R/W	Adjusts current in mixer and PA buffers. Used when $TX_ACTIVE = 0$
1:0	LODIV_CURRENT [1:0]	10	R/W	Adjusts divider currents, except mixer and PA buffers.

#### FSCAL1 (0x61AE) – Tune Frequency Calibration

Bit No.	Name	Reset	R/W	Description
7:2	-	0010 10	R/W0	Reserved
1:0	VCO_CURR[1:0]	01	R/W	Defines current in VCO core. Sets the multiplier between calibrated current and VCO current. For the best value to use, see Table 23-6 in Section 23.15.1.

### FSCAL2 (0x61AF) – Tune Frequency Calibration

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	VCO_CAPARR_OE	0	R/W	Override the calibration result with the value from VCO_CAPARR[5:0].
5:0	VCO_CAPARR[5:0]	10 0000	R*/W	VCO capacitor array setting. Programmed during calibration. Override value when VCO_CAPARR_OE = 1.

#### FSCAL3 (0x61B0) – Tune Frequency Calibration

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	VCO_DAC_EN_OV	0	R/W	Enables the VCO DAC when 1
5:2	VCO_VC_DAC [3:0]	1010	R/W	Bit vector for programming varactor control voltage from VC DAC.
1:0	VCO_CAPARR_CAL_CTRL[1:0]	10	R/W	Calibration accuracy setting for the capacitor array part of the calibration 00: 80 XOSC periods 01: 100 XOSC periods 10: 125 XOSC periods 11: 250 XOSC periods

#### AGCCTRL0 (0x61B1) – AGC Dynamic Range Control

Bit No.	Name	Reset	R/W	Description
7	_	0	R0	Reserved. Read as 0
6	AGC_DR_XTND_EN	1	R/W	0: The AGC performs no adjustment of attenuation in the AAF.
				1: The AGC adjusts the gain in the AAF to achieve extra dynamic range for the receiver.
5:0	AGC_DR_XTND_THR[5:0]	01 1111	R/W	If the measured error between the AGC reference magnitude and the actual magnitude in dB is larger than this threshold, the extra attenuation is enabled in the front end. This threshold should be set higher than 0x0C.
				This feature is enabled by AGC_DR_XTND_EN.



AGCC	AGCCTRL1 (0x61B2) – AGC Reference Level						
Bit No.	Name	Reset	R/W	Description			
7:6	-	00	R0	Reserved. Read as 0			
5:0	AGC_REF[5:0]	01 0001	R/W	Target value for the AGC control loop, given in 1-dB steps. For the best value to use see Table 23-6 in Section 23.15.1.			

#### AGCCTRL2 (0x61B3) – AGC Gain Override

Bit No.	Name	Reset	R/W	Description
7:6	LNA1_CURRENT[1:0]	00	R*/W	Overrride value for LNA 1. Only used when LNA_CURRENT_OE = 1. When read, this register returns the current applied gain setting.
				00: 0-dB gain (reference level)
				01: 3-dB gain
				10: Reserved
				11: 6-dB gain
5:3	LNA2_CURRENT[2:0]	000	R*/W	Overrride value for LNA 2. Only used when LNA_CURRENT_OE = 1. When read, this register returns the current applied gain setting.
				000: 0-dB gain (reference level)
				001: 3-dB gain
				010: 6-dB gain
				011: 9-dB gain
				100: 12-dB gain
				101: 15-dB gain
				110: 18-dB gain
				111: 21-dB gain
2:1	LNA3_CURRENT[1:0]	00	R*/W	Overrride value for LNA 3. Only used when LNA_CURRENT_OE = 1. When read, this register returns the current applied gain setting.
				00: 0-dB gain (reference level)
				01: 3-dB gain
				10: 6-dB gain
				11: 9-dB gain
0	LNA_CURRENT_OE	0	R/W	Write 1 to override the AGC LNA current setting with the values above (LNA1_CURRENT, LNA2_CURRENT, and LNA3_CURRENT).

#### AGCCTRL3 (0x61B4) - AGC Control

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6:5	AGC_SETTLE_WAIT[1:0]	01	R/W	Time for AGC to wait for analog gain to settle after a gain change. During this period, the energy measurement in the AGC is paused.
				00: 15 periods
				01: 20 periods
				10: 25 periods
				11: 30 periods
4:3	AGC_WIN_SIZE[1:0]	01	R/W	Window size for the accumulate and dump function in the AGC
				00: 16 samples
				01: 32 samples
				10: 64 samples
				11: 128 samples
2:1	AAF_RP[1:0]	11	R*/W	Overrides the AGC control signals to AAF when AAF_RP_OE = 1. When read, it returns the applied signal to the AAF.
				00: 9-dB attenuation in AAF
				01: 6-dB attenuation in AAF
				10: 3-dB attenuation in AAF
				11: 0-dB attenuation in AAF (reference level)
0	AAF_RP_OE	0	R/W	Write 1 to override the AGC AAF control signals with the values stored in AAF_RP.



### Registers

### ADCTEST0 (0x61B5) – ADC Tuning

Bit No.	Name	Reset	R/W	Description		
7:6	ADC_VREF_ADJ[1:0]	00	R/W	Quantizer threshold control for test/debug		
5:4	ADC_QUANT_ADJ[1:0]	01	R/W	Quantizer threshold control for test/debug		
3:1	ADC_GM_ADJ[2:0]	000	R/W	Gm control for test/debug		
0	ADC_DAC2_EN	0	R/W	Enables DAC2 for enhanced ADC stability		

#### ADCTEST1 (0x61B6) – ADC Tuning

Bit No.	Name	Reset	R/W	Description
7:4	ADC_TEST_CTRL[3:0]	0000	R/W	ADC test mode selector
3:2	ADC_C2_ADJ[1:0]	11	R/W	Used to adjust capacitor values in ADC
1:0	ADC_C3_ADJ[1:0]	10	R/W	Used to adjust capacitor values in ADC

#### ADCTEST2 (0x61B7) – ADC Tuning

Bit No.	Name	Reset	R/W	Description
7	_	0	R0	Reserved. Read as 0
6:5	ADC_TEST_MODE	00	R/W	Test mode to enable output of ADC data from demodulator. When enabled, raw ADC data is clocked out on the GPIO pins.
				00: Test mode disabled
				01: Data from both I and Q ADCs is output, data rate 76 MHz
				10: Data from I ADC is output. Two and two ADC samples grouped, data rate 38 MHz
				11: Data from Q ADC is output. Two and two ADC samples grouped, data rate 38 MHz
4:3	AAF_RS[1:0]	00	R/W	Controls series resistance of AAF
2:1	ADC_FF_ADJ[1:0]	01	R/W	Adjust feedforward
0	ADC_DAC_ROT	1	R/W	Control of DAC DWA scheme
				0: DWA (scrambling) disabled
				1: DWA enabled



Bit No.	Name	Reset	R/W	Description
7:4	TX_TONE[3:0]	0111	R/W	Enables the possibility to transmit a baseband tone by picking samples from the sine tables with a controllable phase step between the samples. The step size is controlled by TX_TONE. If MDMTEST1.MOD_IF is 0, the tone is superpositioned on the modulated data, effectively giving modulation with an IF. If MDMTEST1.MOD_IF is 1, only the tone is transmitted.
				0000: –6 MHz
				0001: –4 MHz
				0010: –3 MHz
				0011: –2 MHz
				0100: –1 MHz
				0101: –500 kHz
				0110: –4 kHz
				0111: 0
				1000: 4 kHz
				1001: 500 kHz
				1010: 1 MHz
				1011: 2 MHz
				1100: 3 MHz
				1101: 4 MHz
				1110: 6 MHz
				Others: Reserved
3:2	DC_WIN_SIZE[1:0]	01	R/W	Controls the number of samples to be accumulated between each dump of the accumulate-and-dump filter used in dc removal.
				00: 32 samples
				01: 64 samples
				10: 128 samples
				11: 256 samples
1:0	DC_BLOCK_MODE[1:0]	01	R/W	Selects the mode of operation:
				00: The input signal to the dc blocker is passed on to the output without any attempt to remove dc.
				01: Enable dc cancellation. Normal operation
				10: Freeze estimates of dc when sync is found. Start estimating dc again when searching for the next frame
				11: Reserved

#### MDMTEST1 (0x61B9) – Test Register for Modem

Bit No.	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved. Read as 0
4	MOD_IF	0	R/W	0: Modulation is performed at an IF set by MDMTEST0.TX_TONE. 1: A tone is transmitted with frequency set by MDMTEST0.TX_TONE.
3	RAMP_AMP	1	R/W	<ol> <li>Enable ramping of DAC output amplitude during startup and finish.</li> <li>Disable ramping of DAC output amplitude</li> </ol>
2	RFC_SNIFF_EN	0	R/W	0: Packet sniffer module disabled
				1: Packet sniffer module enabled. The received and transmitted data can be observed on GPIO pins.
1	MODULATION_MODE	0	R/W	Set one of two RF modulation modes for RX/TX
				0: IEEE 802.15.4 compliant mode
				1: Reversed phase, non-IEEE compliant
0	RESERVED	0	R/W	Reserved. Do not write.



#### Registers

### DACTEST0 (0x61BA) – DAC Override Value

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6:0	DAC_Q_0[6:0]	000 0000	R/W	Q-branch DAC override value when DAC_SRC = 001
				If DAC_SRC is set to be ADC data, CORDIC magnitude, channel filtered data, then DAC_Q_O controls the part of the word in question that actually is muxed to the DAC as described in the following list.
				00 0110 ≥ bits 6:0
				00 0111 ≥ bits 7:1
				00 1000 ≥ bits 8:2
				and so on
				If an invalid setting is chosen, then the DAC outputs only zeros (minimum value).

#### DACTEST1 (0x61BB) – DAC Override Value

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6:0	DAC_I_0[6:0]	000 0000	R/W	I-branch DAC override value when DAC_SRC = 001
				If DAC_SRC is set to be ADC data, CORDIC magnitude, channel filtered data, then DAC_I_O controls the part of the word in question that actually is muxed to the DAC as described in the following list.
				00 0110 ≥ bits 6:0
				00 0111 ≥ bits 7:1
				00 1000 ≥ bits 8:2
				And so on
				If an invalid setting is chosen, then the DAC outputs only zeros (minimum value),

#### DACTEST2 (0x61BC) - DAC Test Setting

Bit No.	Name	Reset	R/W	Description
7:3	-	0010 1	R0	Reserved
2:0	DAC_SRC[2:0]	000	R/W	The TX DAC's data source is selected by DAC_SRC according to:
				000: Normal operation (from modulator).
				001: The DAC_I_O and DAC_Q_O override values
				010: ADC data after decimation, magnitude-controlled by DAC_I_O and DAC_Q_O
				011: I/Q after decimation, channel and dc filtering, magnitude-controlled by DAC_I_O and DAC_Q_O
				100: Cordic magnitude output and front-end gain is output, magnitude-controlled by DAC_I_O and DAC_Q_O
				101: RSSI I output on the I DAC
				111: Reserved

### ATEST (0x61BD) – Analog Test Control

Bit No.	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved. Read as 0
5:0	ATEST_CTRL[5:0]	00 0000	R/W	Controls the analog test mode: 00 0000: Disabled 00 0001: Enables the temperature sensor (see also the TR0 register description in Section 12.2.10). Other values reserved.



### RFRND (0x61A7) – Random Data

Bit No.	Name	Reset	R/W	Description
7:2	-	0000 00	R0	Reserved. Read as 0
1	QRND	0	R/W	Random bit from the Q channel of the receiver
0	IRND	0	R/W	Random bit from the I channel of the receiver

### PTEST0 (0x61BE) – Override Power-Down Register

Bit No.	Name	Reset	R/W	Description	
7	PRE_PD	0	R/W	Prescaler power-down signal when PD_OVERRIDE = 1	
6	CHP_PD	0	R/W	Charge-pump power-down signal when PD_OVERRIDE = 1	
5	ADC_PD	0	R/W	Analog-to-digital converter power-down signal when PD_OVERRIDE = 1	
4	DAC_PD	0	R/W	Digital-to-analog converter power-down signal when PD_OVERRIDE = 1	
3:2	LNA_PD[1:0]	0	R/W	Low-noise amplifier power-down signal. Defines LNA/mixer PD modes.         00:       Power up         01:       LNA off, mixer/regulator on         10:       LNA/mixer off, regulator on         11:       PD         When PD_OVERRIDE = 1	
1	TXMIX_PD	0	R/W	Transmit mixer power-down signal when PD_OVERRIDE = 1	
0	AAF_PD	0	R/W	Antialiasing filter power-down signal when PD_OVERRIDE = 1	

### PTEST1 (0x61BF) – Override Power-Down Register

Bit No.	Name	Reset	R/W	Description
7:4	-	0000	R0	Reserved. Read as 0
3	PD_OVERRIDE	0	R/W	Override enabling/disabling of various modules. For debug and testing only. It is impossible to override hard-coded BIAS_PD[1:0] dependency.
2	PA_PD	0	R/W	Power amplifier power-down signal when PD_OVERRIDE = 1
1	VCO_PD	0	R/W	Voltage-controlled oscillator power-down signal when PD_OVERRIDE = 1
0	LODIV_PD	0	R/W	LO power-down signal when PD_OVERRIDE = 1

### TEXAS INSTRUMENTS

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Rec	isters
ney	131613

Bit No.	Name	Reset	R/W	Description	
7	_	0	R0	Reserved. Read as 0	
6	RFC_OBS_POL0	0	R/W	The signal chosen by RFC_OBS_MUX0 is XORed with this bit.	
5:0	RFC_OBS_MUX0	00 0000	R/W	Controls which observable signal from RF Core is to be muxed out to rfc_obs_sigs[0].	
				00 0000: 0 – Constant value	
				00 0001: 1 – Constant value	
				00 1000: rfc_sniff_data – Data from packet sniffer. Sample data on rising edges of sniff_clk.	
				00 1001: rfc_sniff_clk – 250kHz clock for packet sniffer data.	
				00 1100: rssi_valid – Pin is high when the RSSI value has been updated at least once since RX was started. Cleared when leaving RX.	
				00 1101: demod_cca – Clear channel assessment. See FSMSTAT1 register for details on how to configure the behavior of this signal.	
				00 1110: sampled_cca – A sampled version of the CCA bit from demodulator. The value is updated whenever a SSAMPLECCA or STXONCCA strobe is issued.	
				00 1111: sfd_sync – Pin is high when a SFD has been received or transmitted. Cleared when leaving RX/TX respectively. Not to be confused with the SFD exception	
				01 0000: tx_active – Indicates that FFCTRL is in one of the TX states. Active-high. Note: This signal might have glitches, because it has no output flip-flop and is based on the current state register of the FFCTRL FSM.	
				01 0001: rx_active – Indicates that FFCTRL is in one of the RX states. Active-high. Note: This signal might have glitches, because it has no output flip-flop and is based on the current state register of the FFCTRL FSM.	
				01 0010: ffctrl_fifo – Pin is high when one or more bytes are in the RXFIFO. Low during RXFIFO overflow.	
				01 0011: ffctrl_fifop – Pin is high when the number of bytes in the RXFIFO exceeds th programmable threshold or at least one complete frame is in the RXFIFO. Also high during RXFIFO overflow. Not to be confused with the FIFOP exception.	
				01 0100: packet_done – A complete frame has been received. I.e., the number of bytes set by the frame-length field has been received.	
				01 0110: rfc_xor_rand_i_q – XOR between I and Q random outputs. Updated at 8 MHz.	
				01 0111: rfc_rand_q – Random data output from the Q channel of the receiver. Updated at 8 MHz.	
				01 1000: rfc_rand_i – Random data output from the I channel of the receiver. Updated at 8 MHz	
				01 1001: lock_status - 1 when PLL is in lock, otherwise 0	
				10 1000: pa_pd – Power amplifier power-down signal	
				10 1010: Ina_pd – LNA power-down signal	
				Others: Reserved	

### RFC_OBS_CTRL1 (0x61EC) – RF Observation Mux Control

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	RFC_OBS_POL1	0	R/W	The signal chosen by RFC_OBS_MUX1 is XORed with this bit.
5:0	RFC_OBS_MUX1	00 0000	R/W	Controls which observable signal from RF Core is to be muxed out to rfc_obs_sigs[1]. See description of RFC_OBS_CTRL0 for details.

### RFC_OBS_CTRL2 (0x61ED) - RF Observation Mux Control

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	RFC_OBS_POL2	0	R/W	The signal chosen by RFC_OBS_MUX2 is XORed with this bit.
5:0	RFC_OBS_MUX2	00 0000	R/W	Controls which observable signal from RF Core is to be muxed out to rfc_obs_sigs[2]. See description of RFC_OBS_CTRL0 for details.



TXFIL	TXFILTCFG (0x61FA) – TX Filter Configuration						
Bit No.	Name	Reset	R/W	Description			
7:4	-	0	R0	Reserved			
3:0	FC	0xF	R/W	Sets TX anti-aliasing filter to appropriate bandwidth. Reduces spurious emissions close to signal. For the best value to use, see Table 23-6 in Section 23.15.1.			

#### IVCTRL (0x6265) - Analog control register (CC2533 only)

Bit No.	Name	Reset	R/W	Description	
7:6	-	00	R0	Reserved. Always read as 0.	
5:4	DAC_CURR_CTRL	01	R/W	Controls bias current to DAC	
				00: 100% IVREF, 0% IREF bias	
				01: 60% IVREF, 40% IREF bias	
				10: 40% IVREF, 60% IREF bias	
				11: 0% IVREF, 100% IREF bias	
3	LODIV_BIAS_CTRL	0	R/W	Controls bias current to LODIV	
				1: PTAT bias	
				0: IVREF bias	
2	TXMIX_DC_CTRL	0	R/W	Controls dc bias in TXMIX	
1:0	PA_BIAS_CTRL	11	R/W	Controls bias current to PA	
				00: IREF bias	
				01: IREF and IVREF bias	
				10: PTAT bias	
				11: Increased PTAT slope bias	

Registers



# CC2540/CC2541 Bluetooth low energy Radio

The CC2540 and CC2541 provide a *Bluetooth* low energy compliant radio transceiver. On the CC2540/41, radio operation is controlled by the *Bluetooth* low energy stack. The application is not allowed to access the radio directly. The application interacts with the radio by sending API commands to the stack. The TI BLE stack with documentation is available at <u>www.ti.com/blestack</u>. The CC2541 may also be run in proprietary mode; see Chapter 25 for a description of the operation in that case.

Горіс		Page
24.1	Registers	288

Registers

### 24.1 Registers

The following status registers are available to the user:

#### RFSTAT (0x618D) RF Core Status

Bit	Name	Reset	R/W	Description
7	MOD_UNDERFLOW	0	R/W0	Modulator has underflowed. Must be cleared by software
6:5	DEM_STATUS	00	R	Demodulator status
				00: Idle
				01: Active
				10: Finishing
				11: Error
4	SFD	0	R	High when the access address has been sent in TX or when sync has been obtained in RX
3	CAL_RUNNING	0	R	Frequency synth calibration status
				0: Calibration done or not started
				1: Calibration in progress
2	LOCK_STATUS	0	R	1 when PLL is in lock, otherwise 0
1	TX_ACTIVE	0	R	Status signal, active when radio is in transmit state
0	RX_ACTIVE	0	R	Status signal, active when radio is in receive state

### RFC_OBS_CTRL0 (0x61AE) RF Observation Mux Control 0

Bit	Name	Reset	R/W	Description
7	_	0	R0	Reserved. Read as 0
6	RFC_OBS_POL0	0	R/W	The signal chosen by RFC_OBS_MUX0 is XORed with this bit.
5:0	RFC_OBS_MUX0	00 0000	R/W	Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(0).
				00 0000: 0 – Constant value
				00 0001: 1 – Constant value
				00 1001: TX active
				00 1010: RX_active
				11 0000: High from when receiver has found access address until packet is finished, low otherwise
				11 0001: High from the access address has been transmitted until end of packet, low otherwise
				Other values reserved

#### RFC_OBS_CTRL1 (0x61AF) RF Observation Mux Control 1

Bit	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	RFC_OBS_POL1	0	R/W	The signal chosen by RFC_OBS_MUX1 is XORed with this bit.
5:0	RFC_OBS_MUX1	00 0000	R/W	Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(1).
				00 0000: 0 – Constant value
				00 0001: 1 - Constant value
				00 1001: TX active
				00 1010: RX_active
				11 0000: High from when receiver has found access address until packet is finished, low otherwise
				11 0001: High from the access address has been transmitted until end of packet, low otherwise
				Other values reserved



#### RFC_OBS_CTRL2 (0x61B0) RF Observation Mux Control 2

Bit	Name	Reset	R/W	Description	
7	-	0	R0	Reserved. Read as 0	
6	RFC_OBS_POL2	0	R/W	The signal chosen by RFC_OBS_MUX2 is XORed with this bit.	
5:0	RFC_OBS_MUX2	00 0000	R/W	Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(2).	
				00 0000: 0 - Constant value	
				00 0001: 1 - Constant value	
				00 1001: TX active	
				00 1010: RX_active	
				11 0000: High from when receiver has found access address until packet is finished, low otherwise	
				11 0001: High from the access address has been transmitted until end of packet, low otherwise	
				Other values reserved	

#### ATEST (0x61A9) – Analog Test Control

Bit	Name	Reset	R/W	Description	
7:6	-	00	R0	Reserved. Read as 0	
5:0	ATEST_CTRL[5:0]	00 0000	R/W	Controls the analog test mode: 00 0000: Disabled 00 0001: Enables the temperature sensor (see also the TR0 register description in Section 12.2.10). Other values reserved.	



# CC2541 Proprietary Mode Radio

In proprietary mode, the CC2541 radio supports data rates up to 2 Mbps, and has extensive baseband automation, including auto-acknowledgment and address decoding. The **RF Core** controls the analog radio module and the RF transceiver state. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events.

It has 1 KB of dedicated RAM, which holds the 128-byte transmit and receive FIFO.

This chapter describes the proprietary mode operation of the CC2541 devices and features in the LLE program. For *Bluetooth* low energy operation, see Chapter 24.

Topic

25.1	RF Core	292
25.2	Interrupts	292
25.3	RF Core Data Memory	<b>293</b>
25.4	Bit-Stream Processor	304
25.5	Frequency and Channel Programming	309
25.6	Modulation Formats	309
25.7	Receiver	309
25.8	Packet Format	310
25.9	Link Layer Engine	314
25.10	Random Number Generation	330
25.11	Packet Sniffing	331
25.12	Registers	332



# RF Core

### 25.1 RF Core

The **RF core** contains several submodules that support and control the analog radio modules. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events.

The **link-layer engine** (LLE) controls the RF transceiver state and most of the dynamically controlled analog signals such as power up / down of analog modules. The LLE is used to provide the correct sequencing of events (such as performing an FS calibration before enabling the receiver). It handles packet assembly and decoding, including automatic length field handling, address insertion and filtering and CRC generation and checking.

The **radio data RAM** holds a FIFO for transmit data (Tx FIFO) and a FIFO for receive data (Rx FIFO). Both FIFOs are 128 bytes long and have hardware control of pointers when data is entered and removed from the FIFOs. In addition, the RAM contains six segments of 128 bytes, one of which is used for communication with the LLE.

The **bit-stream processor** is used for whitening and de-whitening transferred signals and CRC generation and check.

The modulator transforms raw data into I/Q signals to the transmitter DAC.

The **demodulator** is responsible for retrieving the over-the-air data from the received signal.

The frequency synthesizer (FS) generates the carrier wave for the RF signal.

### 25.2 Interrupts

The radio is associated with two **interrupt** vectors on the CPU. These are the RFERR interrupt (interrupt 0) and the RF interrupt (interrupt 12) with the following functions.

- RFERR: Error situations in the radio are signaled using this interrupt.
- RF: Interrupts coming from normal operation are signaled using this interrupt.

The RF interrupt vector combines the interrupts in RFIF. Note that these RF interrupts are rising-edge triggered. Thus, an interrupt is generated when, e.g., the SFD status flag goes from 0 to 1. The RFIF interrupt flags are described in Section 25.2.1.

#### 25.2.1 Interrupt Registers

Two main interrupt-control SFR registers are used to enable the RF and RFERR interrupts. These are the following:

- **RFERR**: IEN0.RFERRIE
- **RF**: IEN2.RFIE

Two main interrupt-flag SFR registers hold the RF and RFERR interrupt flags. These are the following:

- **RFERR**: TCON.RFERRIF
- **RF**:s1CON.RFIF

The two interrupts generated from the RF core are a combination of several sources within the RF core. Each of the individual sources has its own enable and interrupt flags in RF core. Flags can be found in RFIRQF0, RFIRQF1, and RFERRF. Interrupt enable masks can be found in RFIRQM0, RFIRQM1, and RFERRM.

The interrupt enable bits in the mask registers are used to enable individual interrupt sources. Note that masking an interrupt source does not affect the updating of the corresponding status in the flag registers.

Due to the use of individual interrupt masks in the RF core, the interrupts coming from the RF core have two-layered masking, and care must be taken when processing these interrupts. The procedure is described as follows.

To clear an interrupt from the RF core one needs to clear two flags, both the flag set in the RF core and the one set in the main interrupt flag SFR registers, SICON or TCON (depending on which interrupt is triggered). If a flag is cleared in the RF core and there are other unmasked flags standing, the main interrupt flag is set. Exiting the interrupt service routine with the main interrupt flag set causes the interrupt service routine to be executed again.

TIP: For proper handling of interrupts in ISRs, the following is advised:

- At the start of the ISR, read and store the RF core flags
- Process the interrupts
- Clear the main interrupt flag
- Clear the processed RF core flags. It is important that this is done in a single operation.

# 25.3 RF Core Data Memory

The radio core has 1024 bytes of data RAM divided into eight pages of 128 bytes each. The pages are to be used as shown in Table 25-1.

Page Number	Assignment
0	RAM-based registers
1	For Rx with auto ACK: ACK payload FIFO for addresses 2 and 3
2	For Rx with auto ACK: ACK payload FIFO for addresses 4 and 5
3	For Rx with auto ACK: ACK payload FIFO for addresses 6 and 7
4	Free for MCU use
5	Additional RAM-based registers/Reserved for LLE
6	Rx FIFO
7	Tx FIFO/for Rx with auto ACK: ACK payload FIFO for addresses 0 and 1

### Table 25-1. Radio RAM Pages

The active memory page is selected in register RFRAMCFG.PRE. The selected page is accessible at XDATA addresses 0x6000–0x607F. The Rx FIFO page (page 6) is also accessible at XDATA addresses 0x6080–0x60FF. The Tx FIFO page (page 7) is also accessible at XDATA addresses 0x6100–0x617F.

A page is used for transferring parameters to the LLE, see Section 25.3.3.

There is no hardware protection to prevent the MCU from overwriting memory used by the LLE and the FIFO. Thus the MCU should never write to page 5 (except for special dedicated registers). The MCU should write to pages 0, 1, 2, 3, and 7 only as specified in this chapter. Writes to the FIFO pages should only be done in ways compatible with FIFO operation, except for accessing the Tx FIFO page while running an Rx task with auto ACK.

Pages 0, 1, 6, and 7 have retention in all power modes, whereas the contents of pages 2–5 are lost in PM2 and PM3.

Radio core hardware registers are located at XDATA addresses 0x6180–0x61F7. Figure 25-1 shows the mapping of radio memory to MCU XDATA memory space.



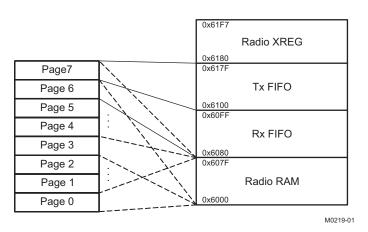


Figure 25-1. Mapping of Radio Memory to MCU XDATA Memory Space

### 25.3.1 FIFOs

The FIFOs are used for transporting data between the MCU and the radio. The FIFOs have hardware support for read and write pointer increment with circular buffering, overflow and underflow detection, and flushing of last entry or the entire FIFO.

The Rx and Tx FIFOs are fundamentally two similar modules. Each FIFO has four pointers: the write pointer (WP), the read pointer (RP), the start-of-packet write pointer (SWP), and the start-of-packet read pointer (SRP). WP and RP give the index in the FIFO where the next byte is to be written and read, respectively. SWP is used to indicate the start of the current packet being written, and SRP is used to indicate the start of the current packet being read. The use of the pointers is indicated in Figure 25-2.

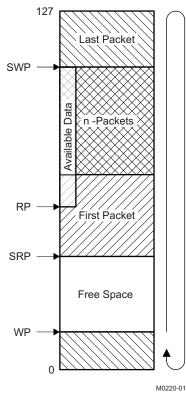


Figure 25-2. FIFO Pointers

The Tx FIFO and Rx FIFO may be accessed though the SFR register RFD (0xD9). Data is written to the Tx FIFO when writing to the RFD register. Data is read from the Rx FIFO when the RFD register is read. In addition, there are separate read and write registers for each FIFO (RFRXFRD, RFRXFWR, RFTXFRD, RFTXFWR).

The Rx FIFO or Tx FIFO can be cleared by issuing CMD_RXFIFO_RESET or CMD_TXFIFO_RESET (see Section 25.3.1.2), respectively. The contents of both FIFOs can be cleared by issuing CMD_FIFO_RESET.

Four operations are defined to handle the four pointers:

- *Deallocate* is setting SRP equal to RP. This should be done when the treatment of a packet that has been read from the FIFO is finished.
- *Retry* is setting RP equal to SRP. This is done to re-read a packet that has been read from the FIFO previously.
- *Discard* is setting WP equal to SWP. This is done to remove a packet that had been written to the FIFO.
- *Commit* is setting SWP equal to WP. This is done to confirm the writing of a packet to the FIFO and making it available to be read out.

Using the register RFFCFG, it is possible to set up auto-commit and auto-deallocate for each of the FIFOs. If auto-commit is enabled, SWP is set equal to WP each time a byte is written to the FIFO. If auto-deallocate is enabled, SRP is set equal to RP each time a byte is read from the FIFO. By default, auto-commit is enabled for the Tx FIFO and auto-deallocate is enabled for the Rx FIFO. This is also the recommended setting. However, if packets that exceed the FIFO size are to be supported, auto-commit must be enabled for the Rx FIFO and auto-deallocate for the Tx FIFO; see Section 25.8.1 and Section 25.8.2 for details. If auto-commit is disabled for the Tx FIFO, the MCU must issue a commit command after writing a packet to the Tx FIFO, and if auto-deallocate is disabled for the Rx FIFO, the MCU must issue a deallocate command after reading a packet from the Rx FIFO.

### 25.3.1.1 FIFO Status and Interrupts

The XREG registers RFRXFLEN and RFTXFLEN provide information on the amount of data in the FIFOs. This is the number of bytes between SRP and WP, i.e., the number of bytes that is not free space in Figure 25-2. The register RFFSTATUS contains status bits for each of the FIFOs. FIFO empty is defined as the length being 0, and FIFO full is defined as the length being 128. The amount of data between RP and SWP is known as available data, and there is a status bit in the RFFSTATUS register telling whether there is available data for each of the FIFOs.

An attempt to write to a full FIFO results in a FIFO overflow. The data written is then ignored and the RXOVERF or TXOVERF flag is set in the RFERRF register, causing an RFERR interrupt. An attempt to read from a FIFO when no data is available results in a FIFO underflow. The value read is then zero, and the RXUNDERF or TXUNDERF flag is set in the RFERRF register, causing an RFERR interrupt.

Registers RFTXFTHRS and RFRXFTHRS are used to set threshold points for the Tx and Rx FIFOs, respectively. Each FIFO has one status flag and two interrupt flags; when the amount of data in the FIFO crosses the threshold, an interrupt flag is set. The FIFO status flags are available in RFFSTATUS, and the interrupt flags are available in RFIRQF0.

When the amount of data in the FIFO is above the threshold, i.e. RFXXFLEN is greater than or equal to RFXXFTHRS, the status bit XXDTHEX of RFFSTATUS is 1, otherwise it is 0.

When data is written to the FIFO causing the FIFO threshold to be crossed, i.e., **XXDTHEX** going from 0 to 1, the corresponding interrupt flag is set.

When data is read from the FIFO causing the FIFO threshold to be crossed, i.e., **XXDTHEX** going from 1 to 0, the corresponding interrupt flag is set.

### 25.3.1.2 Command Register

The command register RFST can be used for sending commands to the FIFO. Commands in the range 0x80–0xFF are commands to the FIFO. Other commands are commands to the LLE; see Section 25.9.1.



The supported FIFO commands are listed in Table 25-2. A command in the range of 0x80–0xFF that does not match any of the listed commands is ignored.

Number	Command Name	Description
0x81	CMD_RXFIFO_RESET	Reset (empty) Rx FIFO. Set RFRXF* := 0
0x82	CMD_RXFIFO_DEALLOC	Deallocate Rx FIFO. This sets RFRXFSRP := RFRXFRP.
0x83	CMD_RXFIFO_RETRY	Retry Rx FIFO. This sets RFRXFRP := RFRXFSRP
0x84	CMD_RXFIFO_DISCARD	Discard Rx FIFO. This sets RFRXFWP := RFRXFSWP
0x85	CMD_RXFIFO_COMMIT	Commit Rx FIFO. This sets RFRXFSWP := RFRXFWP
0x91	CMD_TXFIFO_RESET	Reset (empty) Tx FIFO. Set RFRXF* := 0
0x92	CMD_TXFIFO_DEALLOC	Deallocate Tx FIFO. This sets RFTXFSRP := RFTXFRP.
0x93	CMD_TXFIFO_RETRY	Retry Tx FIFO. This sets RFTXFRP := RFTXFSRP
0x94	CMD_TXFIFO_DISCARD	Discard Tx FIFO. This sets RFTXFWP := RFTXFSWP
0x95	CMD_TXFIFO_COMMIT	Commit Tx FIFO. This sets RFTXFSWP := RFTXFWP
0xF1	CMD_FIFO_RESET	Reset both FIFOs
0xF2	CMD_FIFO_DEALLOC	Deallocate both FIFOs
0xF3	CMD_FIFO_RETRY	Retry both FIFOs
0xF4	CMD_FIFO_DISCARD	Discard both FIFOs
0xF5	CMD_FIFO_COMMIT	Commit both FIFOs

#### Table 25-2. Commands to FIFO via RFST Register

#### 25.3.1.3 FIFO Pointer Operations

The FIFO pointers can be accessed directly through registers RFFRXFWP, RFFRXFRP, RFFRXFSWP, RFFRXFSWP, RFFTXFSWP, RFFTXFSWP, and RFFTXFSRP.

Because the placement of the pointers may be the same for an empty and a full FIFO, there are internal states distinguishing between these situations. This means that while any value can be written to the pointer registers, certain rules must be observed for the FIFO to function reliably after the pointer write.

Any writes to a pointer must be considered to move that pointer up. Hence, writing N to a pointer already holding N is considered equivalent to moving that pointer up 128 places, writing N-1 is equivalent to moving the pointer up 127 places, and so on.

The pointers must maintain a specific ordering: (Going from lowest position to highest) SRP, RP, SWP, WP.

A lower pointer may be moved up to but not past a higher pointer, whereas the highest pointer (WP) may be moved down to, but not past the lower.

### 25.3.1.4 Cooperation With LLE

The LLE performs FIFO operations as part of its operation. In order to avoid conflicts between the LLE and the MCU, access to FIFO registers should be done according to Table 25-3. Read accesses can always be made, except for the data-read registers, which causes the read pointers to be modified. If the MCU reads a register, one must take into account that the value may change at any time due to accesses from the LLE. The reset FIFO commands should only be run by the MCU between LLE tasks. They are marked with an asterisk in Table 25-3.

#### Table 25-3. Access to FIFO Registers

Register	Read Access	Write Access
RFD	MCU	MCU
RFST (FIFO commands)	N/A	Depends on command: 0x81 : Reset Rx FIFO: MCU* 0x82 : Deallocate Rx FIFO: MCU 0x83 : Retry Rx FIFO: MCU 0x84: Discard Rx FIFO: LLE 0x85: Commit Rx FIFO: LLE 0x91 : Reset Tx FIFO: MCU* 0x92 : Deallocate Tx FIFO: LLE ⁽¹⁾ 0x93 : Retry Tx FIFO: LLE ⁽¹⁾ 0x94: Discard Tx FIFO: MCU 0x95: Commit Tx FIFO: MCU 0xF1: Reset both fifos: MCU* 0xF2: Deallocate both fifos: none 0xF4: Discard both fifos: none 0xF5: Commit both fifos: none
RFFDMA0	Both	MCU
RFFDMA1	Both	MCU
RFFSTATUS	Both	N/A
RFFCFG	Both	MCU
RFRXFLEN	Both	N/A
RFRXFTHRS	Both	MCU
RFRXFWR	N/A	LLE
RFRXFRD	MCU	N/A
RFRXFWP	Both	LLE
RFRXFRP	Both	MCU
RFRXFSWP	Both	LLE
RFRXFSRP	Both	MCU
RFTXFLEN	Both	N/A
RFTXFTHRS	Both	MCU
RFTXFWR	N/A	MCU
RFTXFRD	LLE	N/A
RFTXFWP	Both	MCU
RFTXFRP	Both	LLE
RFTXFSWP	Both	MCU
RFTXFSRP	Both	LLE

⁽¹⁾ MCU if PRF_ADDR_ENTRY*n*.CONF.RETRY is 1

## 25.3.2 DMA

It is possible to use direct memory access (DMA) to move data between memory and the radio. See Chapter 8 for a detailed description on how to set up and use DMA transfers.

There are two DMA triggers associated with the radio: the RADIO DMA triggers 0 and 1 (DMA triggers 19 and 11).

The radio DMA trigger source is selected in registers RFFDMA0 and RFFDMA1. See the register descriptions in Section 25.12 for details.

### 25.3.3 RAM-Based Registers

A list of the memory entries of the general radio RAM area used for parameter transfer is shown in Table 25-5. All these registers are in page 0 of the radio RAM. Each memory entry is considered a RAM-based register and has a name. Numeric values that are two bytes long are represented in little-endian format.

The radio RAM registers have no defined reset value and must therefore be initialized by the MCU.

The registers SEMAPHORE0 and SEMAPHORE1 can be used to verify data integrity. These registers are changed to 0 when they are read. If a semaphore register is read and the value was 1, the semaphore has been successfully taken, and subsequent reads of the register return 0 until the semaphore is released. If a semaphore register is read as 0, the semaphore was not free. A semaphore can be released by writing 1 to the semaphore register; this should only be done if the semaphore has previously been taken by the MCU. The LLE takes SEMAPHORE0 when a task starts and SEMAPHORE1 when the radio has been set up. Both semaphores are released by the LLE at the end of the task. SEMAPHORE2 is not used by the LLE. If the LLE is not granted the semaphore, it generates an error. If SEMAPHORE0 and SEMAPHORE1 are taken by the MCU before registers protected by these semaphores are modified by the MCU, data integrity is ensured, and an error occurs if the LLE is accidentally started while such an access is going on.

Where bit numbering is used, bit 0 is the LSB and bit 7 is the MSB. Multi-byte fields are little-endian.

The detailed breakdown of the address entries ADDR_ENTRY0-ADDR_ENTRY7 is shown in Table 25-4 or Table 25-6, depending on the operational mode.

The Prot columns of Table 25-4, Table 25-5, and Table 25-6 list the type of protection for each entry:

Sem0: Entries protected by SEMAPHOREO. Should only be written by the MCU while the LLE does not have SEMAPHOREO. Is not modified by the LLE.

Sem1: Entries protected by SEMAPHORE1. Should only be written by the MCU while the LLE does not have SEMAPHORE1. Is not modified by the LLE.

Sem1/R: Entries containing state variables and accumulative counters that are updated by the LLE. They may be read by the MCU after a receive or transmit interrupt to see how many packets have been received or transmitted. The MCU must take into account that at the time these values are read, some of them may have been updated for the next interrupt and some not. When the LLE does not have SEMAPHORE1, the MCU may write to them to initialize. The counters are not initialized by the LLE.

None: No semaphore protection; special rules apply for access.

Name	Addr	Prot	Description
PRF_CHAN	0x6000	Sem0	Bits 0–6: FREQ Frequency to use. 0: 2379 MHz 1 MHz steps 116: 2495 MHz 117–126: Reserved 127: The LLE does not program frequency; it is to be set up by the MCU through the FREQCTRL and MDMTEST1 registers. Bit 7: SYNTH ON
			0: Turn off synthesizer when task is done. 1: Leave synthesizer running after task is done.

#### Table 25-4. RAM-Based Registers



		Dased Registe	
Name	Addr	Prot	Description
			Configuration of task control <b>Bits 0–1</b> : MODE (operation mode) 00: Basic mode, fixed length 01: Basic mode, variable length 10: Auto mode, 9-bit header 11: Auto mode, 10-bit header
			<b>Bit 2</b> : REPEAT (repeated operation) 0: Single operation 1: Repeated operation
			<b>Bit 3</b> : START_CONF (start configuration) 0: Start each receive/transmit immediately 1: Start each receive/transmit on Timer 2 event 1
PRF_TASK_CONF	0x6001	Sem0	<ul> <li>Bits 4–5: STOP_CONF (stop configuration)</li> <li>00: No stop based on Timer 2.</li> <li>01: End task after current packet is done on Timer 2 event 2 (end immediately in sync search or wait)</li> <li>10: Stop transmit/receive immediately on Timer 2 event 2</li> <li>11: End task on Timer 2 event 2 in first sync search or clear channel assessment. No stop after first sync search or clear channel assessment.</li> </ul>
			Bit 6: TX_ON_CC_CONF 0: Listen until RSSI drops below given level, then start Tx. 1: End task if RSSI is above given level
			<b>Bit 7</b> : REPEAT_CONF For TX_ON_CC with REPEAT = 1: 0: Listen again on repeated operation and retransmissions 1: Listen only before the first transmission, then transmit every time For RX with REPEAT = 1: 0: Recalibrate the synthesizer before listening for new packets

### Table 25-4. RAM-Based Registers (continued)

1: Recalibrate the synthesizer only when the task

starts



Name	Addr	Prot	Description
PRF_FIFO_CONF	0×6002	Sem1	Configure FIFO use <b>Bit 0</b> : AUTOFLUSH_IGN Keep received packets with unexpected sequence number in the Rx FIFO. 0: Keep 1: Auto-flush <b>Bit 1</b> : AUTOFLUSH_CRC Keep received packets with CRC error in the Rx FIFO. 0: Keep 1: Auto-flush <b>Bit 2</b> : AUTOFLUSH_EMPTY Keep packets with no payload in the Rx FIFO. 0: Keep 1: Auto-flush <b>Bit 3</b> : RX_STATUS_CONF Rx FIFO channel information 0: Do not append RSSI and RES 1: Append RSSI and RES <b>Bits 4–5</b> : RX_ADDR_CONF Rx FIFO address and config byte configuration 00: Do not include address or config byte in Rx FIFO 01: Include received address in Rx FIFO (1-byte addresses only), but no config byte 10: Include config byte in Rx FIFO (1-byte addresses only), but no config byte 11: Include received address (1-byte addresses only) and config byte in Rx FIFO <b>Bits 6–7</b> : TX_ADDR_CONF Tx FIFO address and config byte configuration 00: No address or config byte addresses only) and config byte in Rx FIFO <b>Bits 6–7</b> : TX_ADDR_CONF Tx FIFO address and config byte configuration 00: No address or config byte; read address from PRF_ADDR_ENTRY0 01: Include address byte in Tx FIFO, no config byte 10: Include config byte and use address index in that byte to find address from PRF_ADDR_ENTRY1 11: Read address from Tx FIFO followed by config byte (where address information is ignored). Not allowed for PRF_TASK_CONF . MODE = 00 or 01.
PRF_PKT_CONF	0x6003	Sem0	Packet configuration Bit 0: ADDR_LEN. Number of address bytes (0 or 1). Bit 1: AGC_EN 0: Do not use AGC 1: Use AGC (Section 25.9.2.1) Bit 2: START_TONE 0: Ordinary transmission 1: Override extra preamble bytes with tone and reduce synthesizer calibration time accordingly (Section 25.9.2.2)
			Bits 3-7: Reserved, always write 0.
PRF_CRC_LEN	0x6004	Sem1	Number of CRC bytes. Permitted values: 0–4
PRF_RSSI_LIMIT	0x6005	Sem1	For transmit on clear channel. Start a transmit task by listening to the channel; start transmitting if the RSSI drops below the level (signed) given in this register.
PRF_RSSI_COUNT	0x6006–0x6007	Sem1	For transmit on clear channel. Number of additional RSSI measurements that must be below the RSSI limit before transmission takes place.

# Table 25-4. RAM-Based Registers (continued)



Table 25-4. RAM-Based Registers	(continued)
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Name	Addr	Prot	Description
PRF_CRC_INIT	0x6008–0x600B	Sem1	Initialization value for CRC. For less than a 4-byte CRC, the first bytes shall be 0 and the last bytes the desired value.
PRF_W_INIT	0x600C	Sem1	Byte to write to register BSP_W before a packet; initializes the PN7 whitener if that is used. If PN9 whitener is used, bit 7 should be 1.
PRF_RETRANS_CNT	0x600D	Sem1	Maximum number of retransmissions in automatic retransmit
PRF_TX_DELAY	0x600E-0x600F	Sem1	Time from end of transmission to new transmission of different payload, given in units of 62.5 ns
PRF_RETRANS_DELAY	0x6010–0x6011	Sem1	Time from end of transmission to retransmission in auto retransmit mode, given in units of 62.5 ns
PRF_SEARCH_TIME	0x6012–0x6013	Sem1	Time to perform search before giving up or retransmitting, given in 31.25-ns units. 0: Never give up. Must be at least 256 if not 0.
PRF_RX_TX_TIME	0x6014–0x6015	Sem1	Time to add to Rx-Tx turnaround time in Rx with auto ACK, given in 31.25-ns units
PRF_TX_RX_TIME	0x6016–0x6017	Sem1	Time to add to Tx-Rx turnaround time in Tx with auto retransmission, given in 31.25-ns units
PRF_ADDR_ENTRY0	0x6018–0x6023		Address structure for address number 0. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY1	0x6024–0x602F		Address structure for address number 1. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY2	0x6030-0x603B		Address structure for address number 2. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY3	0x603C-0x6047		Address structure for address number 3. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY4	0x6048–0x6053		Address structure for address number 4. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY5	0x6054–0x605F		Address structure for address number 5. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY6	0x6060–0x606B		Address structure for address number 6. See Table 25-5 and Table 25-6 for details.
PRF_ADDR_ENTRY7	0x606C-0x6077		Address structure for address number 7. See Table 25-5 and Table 25-6 for details.
PRF_N_TX	0x6078	Sem1/R	Total number of packets transmitted
PRF_LAST_RSSI	0x6079	Sem1/R	RSSI of last received packet
PRF_LAST_DCOFF	0x607A-0x607D	Sem1/R	DC offset of last received packet

Name	Addr	Prot	Description
Name	Addr	Prot	Description           Configure radio hardware           Bits 0–1: RXCAP           00: Do not capture on Rx packets           01: Capture start of every Rx packet           10: Capture end of every Rx packet           11: Capture start of first Rx packet only           Bits 2–3: TXCAP           00: Do not capture on Tx packets           01: Capture start of every Tx packet           10: Capture start of every Tx packet           11: Capture start of every Tx packet           11: Capture start of first Tx packet
PRF_RADIO_CONF	0x607E	Sem0	Bits 4–5: TXIF: TX IF configuration (for 2 Mbps only) 00: Zero IF 01: ±1 MHz IF 10: ±2 MHz IF 11: ±3 MHz IF
			<b>Bit 6</b> : DCOFF: Special dc offset handling 0: Standard dc offset 1: Use special dc offset routine measuring dc offset right after Rx start
			<b>Bit 7</b> : DCWB: Write back dc offset estimate to override registers 0: Do not write back 1: Write back after each received packet with CRC OK
PRF_ENDCAUSE	0x607F	None	Reason why LLE ended task

# Table 25-4. RAM-Based Registers (continued)

### Table 25-5. Address Structure for Auto Mode

Name	Index	Prot	Description
			<b>Bit 0</b> : ENA0 (Enable for primary sync word – Rx task only) 0: Disable address entry for primary sync word 1: Enable address entry for primary sync word
			<b>Bit 1</b> : ENA1 (Enable for secondary sync word – Rx task only) 0: Disable address entry for secondary sync word 1: Enable address entry for secondary sync word
			<b>Bit 2</b> : REUSE (Allow reuse of transmitted packet) 0: LLE deallocates packet after it has been acknowledged 1: LLE does not deallocate packet after it has been acknowledged (this is up to the MCU)
CONF 0x00	0x00	Sem1	<b>Bit 3</b> : AA (Enable auto acknowledgement/auto retranmsmission) 0: Disable auto ack (Rx) or auto retransmission (Tx) for this address 1: Enable auto ack (Rx) or auto retransmission (Tx) for this address
			<b>Bit 4</b> : VARLEN (variable length support) 0: Use fixed length given by RXLENGTH in receiver when receiving packets or ACKs 1: Use variable length up to RXLENGTH in receiver when receiving packets or ACKs
			<b>Bit 5</b> : FIXEDSEQ (fixed sequence number – Tx task only) 0: Insert sequence number from SEQSTAT.SEQ 1: Read sequence number from Tx FIFO
			<b>Bit 6</b> : TXLEN 0: Insert packet length in header when transmitting 1: Used fixed-length word when transmitting Note: Must not be set to 1 unless the peer uses fixed length
RXLENGTH	0x01	Sem1	Maximum length of received packet (0-127)
ADDRESS	0x02	Sem1	Address of packet



Name	Index	Prot	Description			
			<ul> <li>Bit 0: VALID (Rx task only)</li> <li>0: The status is not valid. Any packet is viewed as new. On successful reception of a packet, the LLE sets this bit.</li> <li>1: The status is valid. Only packets with a sequence number and CRC different from the previous one are accepted.</li> </ul>			
			<b>Bits 1–2</b> : SEQ (sequence number). For Rx, the sequence number of the last successfully received packet. For Tx, the sequence number of the next or current packet to be transmitted			
SEQSTAT	0x03	Sem1/R	<b>Bits 3–4:</b> ACKSEQ (ACK sequence number – Rx task only) For Rx with auto ACK, the sequence number of the next or current ACK to be transmitted			
			Bit 5: ACK_PAYLOAD_SENT (Rx task only) 0: The last received packet was not acknowledged with payload. 1: The last received packet was acknowledged with payload.			
			<b>Bit 6</b> : NEXTACK (next ACK buffer to use – Rx task only) 0: Use ACK buffer 0. 1: Use ACK buffer 1.			
ACKLENGTH0	0x04	None	For Rx with auto ACK: Length of payload to be transmitted from buffer 0. When 0, the buffer is free. After the payload has been transmitted and a packet with a new sequence number is receive the value is set to 0 by the LLE. The MCU only writes to the regis when it is zero; the LLE only writes it to zero when it is non-zero.			
ACKLENGTH1	0x05	None	For Rx with auto ACK: Length of payload to be transmitted from buffer 1. When 0, the buffer is free. After the payload has been transmitted and a packet with a new sequence number is received, the value is set to 0 by the LLE. The MCU only writes to the register when it is zero; the LLE only writes it to zero when it is non-zero.			
CRCVAL	0x06–0x07	Sem1/R	CRC value (last two bytes if more than 2 CRC bytes) of last successfully received packet			
N_TXDONE	0x08	Sem1/R	Number of packets transmitted. For auto retransmission, only acknowledged packets with new sequence number are counted. For auto ACK, only packets with new payload are counted when the payload has been confirmed.			
N_RXIGNORED	0x09	Sem1/R	Number of retransmitted packets received with CRC OK			
N_RXOK	0x0A	Sem1/R	Number of new packets received with CRC OK or ACK packets without payload received			
N_RXNOK	0x0B	Sem1/R	Number of packets received with CRC error			

### Table 25-5. Address Structure for Auto Mode (continued)

### Table 25-6. Address Structure for Basic Mode

Name	Index	Prot	Description
			<b>Bit 0</b> : ENA0 (enable for primary sync word – Rx task only) 0: Disable address entry for primary sync word 1: Enable address entry for primary sync word
CONF	0x00	Sem1	<b>Bit 1</b> : ENA1 (enable for secondary sync word – Rx task only) 0: Disable address entry for secondary sync word 1: Enable address entry for secondary sync word
			<b>Bit 2</b> : REUSE (allow reuse of transmitted packet) 0: LLE deallocates packet after it has been transmitted 1: LLE does not deallocate packet after it has been transmitted (this is up to the MCU)
RXLENGTH	0x01	Sem1	Maximum length of received packet (0-255)
ADDRESS	0x02	Sem1	Address of packet
	0x03–0x09		Reserved
N_RXOK	0x0A	Sem1/R	Number of packets received with CRC OK
N_RXNOK	0x0B	Sem1/R	Number of packets received with CRC error



### 25.3.4 Variables in RAM Page 5

Some additional RAM registers are placed in page 5 of the RFCORE RAM. These variables have the prefix PRFX and are listed in Table 25-7. The addresses overlap other RAM registers, and to access them page 5 must be selected using the RFRAMCFG register; see Section 25.3. Some of the registers have a reset value. This value is written by the LLE shortly after it has been taken out of reset by LLECTRL.LLE_EN being set to 1. If the MCU must modify these registers, the modification must be done each time the LLE is reset. After taking the LLE out of reset, the MCU may modify the registers after LLASTAT.LLE_IDLE has gone high.

Name	Addr	Prot	Reset Val	Description
PRFX_LAST_FREQEST	0x6006	Sem1/R	_	Last frequency offset estimate, read from the FREQEST register at the end of receiving each packet
PRFX_RSSI_LIM_LOWER	0x6008	Sem1	0x20	Lower RSSI limit for use in AGC algorithm
PRFX_RSSI_LIM_UPPER	0x6009	Sem1	0x3C	Upper RSSI limit for use in AGC algorithm
PRFX_RSSI_DIFF	0x600A	Sem1	0x14	Difference between high and low RSSI gain
PRFX_LNAGAIN_SAT	0x600B	Sem1	0x4A	LNAGAIN setting to use while close to saturation
PRFX_TONE_DURATION	0x600C-0x600D	Sem1	0x064A	Duration of tone in start of packet if PRF_PKT_CONF . START_TONE = 1, given in 31.25-ns units
PRFX_TONE_OFFSET	0x600E-0x600F	Sem0	0x0600	Time to subtract from Tx synthesizer calibration time if PRF_PKT_CONF . START_TONE = 1, given in 31.25-ns units

⁽¹⁾ Note that the LLE is reset when the device enters PM2 or PM3. This means that the PRFX registers must be re-initialized after coming up from one of these power modes.

The parts of RAM page 5 that are not listed in Table 25-7 are reserved for use by the LLE and should not be written by the MCU.

### 25.4 Bit-Stream Processor

The **bit-stream processor** (BSP) supports automatic insertion of CRC and detection of CRC error with a programmable polynomial of 8, 16, 24, or 32 bits.

The **bit-stream processor** also supports whitening and de-whitening. The whitening sequences supported are a PN7 sequence and a PN 9 sequence compatible with CC2500/CC2510.

The bit-stream processor is used by the LLE to do the whitening and CRC generation and checking. This operation is based on the configuration set up by the MCU. The BSP can also be run in a coprocessor mode to calculate whitened sequences and CRCs. This must only be done while the LLE is not running.

### 25.4.1 Whitening

The BSP supports two whiteners, a PN7 and a PN9 whitener. The register BSP_MODE is used to enable or disable each whitener. When no whitener is enabled, it outputs zero. The whitener sequence is XORed with the transmitted or received signal.

It is possible to enable both whiteners. This is useful, e.g., in conjunction with the test command CMD_TX_TEST (Section 25.9.3) to transmit a white test signal.

### 25.4.1.1 PN7 Whitening

The PN7 whitener is shown in Figure 25-3. It has a 7-bit whitening shift register **w** used for calculating the PN sequence given by the polynomial  $x^7 + x^4 + 1$ . The output is the same as the shift register feedback.

The **w** register must be initialized by writing **w** into register BSP_W.W before starting receiving or transmitting a packet. Doing this sets w ₆ to BSP_W[0], w ₅ to BSP_W[1] and so on up to w ₁ to BSP_W[5]; w ₀ is set to 1.

When running normal receive or transmit tasks, writing to BSP_W is done by the LLE, which writes the value in PRF_W_INIT to this register, but for test commands and co-processor mode, the BSP_W register must be written by the MCU.

 $w_4$ 

 $x^4$ 

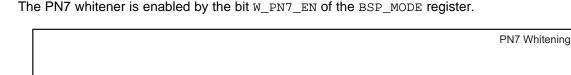
w₅

 $w_6$ 

x⁷

BSP_MODE.W_PN7_EN

B0466-01



 $W_2$ 



w₃

# 25.4.2 CC2500-Compatible PN9 Whitening

w₀

x⁰

Output

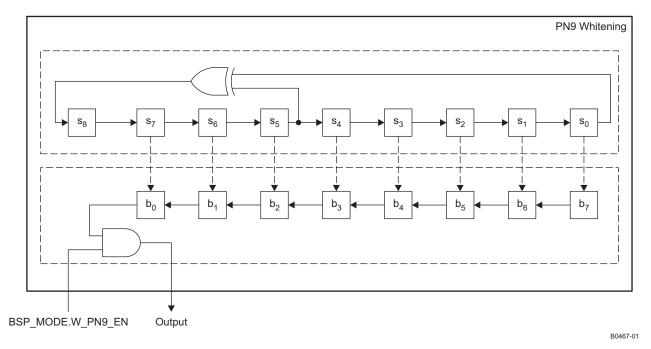
W₁

The CC2500-compatible PN9 whitener is shown in Figure 25-4. It has a 9-bit whitening shift register s and an 8-bit output register b. It produces a whitening sequence compatible with CC2500, CC2510 and other TI devices. These devices use the polynomial  $x^9 + x^4 + 1$ . The whitening sequence is produced one byte at a time, and the byte is bit-reversed before being XORed with a received or transmitted byte. Before starting reception or transmission of a packet, the s and b registers must be initialized to all ones by writing a 1 to register BSP_W.W_PN9_RESET. As for the PN7 whitener, this is done by the LLE for normal receive and transmit tasks, provided that bit 7 of PRF_W_INIT is 1.

In Figure 25-4, the dashed arrows going from the s blocks to the b blocks denote a copy that takes place after the whitening of one byte is done. This means that the first byte is whitened by the 8 bits that are in the b register after initialization (all ones). As this byte is being whitened, the s register is updated. After the first byte is whitened, the value of the s register is copied into the b register and used for whitening the second byte.

The CC2500-compatible whitener is enabled by bit W_PN9_EN of the BSP_MODE register.





#### Figure 25-4. CC2500-Compatible Whitening

### 25.4.3 CRC

A block diagram showing the operation of the CRC module is given in Figure 25-5. The CRC sub-module has two registers:

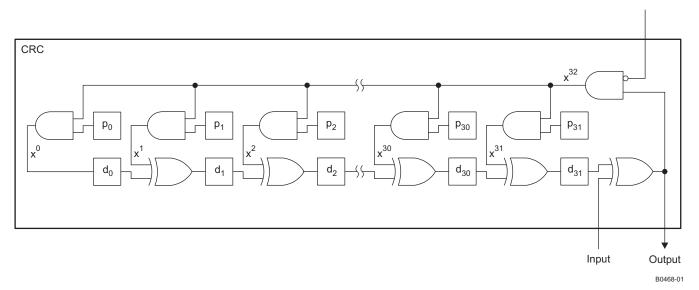
- A 32-bit data shift register d
- A 32-bit register **p** for holding the polynomial

The **p** register defines the shift register used for calculating CRC. There is a feedback tap in the locations where the corresponding bit of **p** is set to 1. The module input is XORed by the output of the shift register, and this becomes the feedback of the shift register.

The current value of the data shift register **d** is the CRC value. Prior to the start of CRC calculation, the **d** and **p** registers should be initialized by writing **d** to registers  $BSP_D[0-3]$  and **p** to registers  $BSP_P[0-3]$ . The  $BSP_P[0-3]$  registers only must be set once, whereas the  $BSP_D[0-3]$  registers should be set again for each packet. In normal transmit and receive modes, this is handled by the LLE, which writes the value of  $PRF_CRC_INIT[0-3]$  to  $BSP_D[0-3]$ . At the end of CRC calculation, the value of the register is serially shifted out on the output. When performing CRC checking, all the  $BSP_D[0-3]$  registers.

If whitening is enabled, calculated CRC bytes are whitened before transmission, and received CRC bytes are de-whitened before CRC checking.





### Figure 25-5. CRC Module

A 32-bit CRC polynomial can be described by the equation  $x^{32} + a_{31}x^{31} + ... + a_{1}x^{1} + 1$ , where all  $a_n$  are 0 or 1. To represent this, each P[n] bit in the BSP_P0-BSP_P3 registers should be set to  $a_n$ , and P[0] should be set to 1. To reduce the size of the polynomial to k, set the bits P[33 - k:0] to 0 and P[32 - k] to 1. In this case, the initialization value must have zeros at D[33 - k:0]. In practice, only polynomials of order 8, 16, 24, and 32 are supported, as the number of CRC bits produced in the transmitter and checked in the receiver is always a multiple of 8. The number of CRC bytes produced in normal transmit tasks is given by RAM register PRF_CRC_LEN.

This is summarized in Table 25-8 for the four CRC polynomial orders supported. In the BSP_Px column, the numbers are binary, with the most significant bit at the left. In the PRF_CRC_INIT column, an X indicates the initialization value to use (each X does not have to be the same). Some examples are shown in Table 25-9.

Order	PRF_CRC_LEN	Polynomial	BSP_Px	PRF_CRC_INIT
8	1	$x^8 + a_7 x^7 + \dots + a_1 x^1 + 1$	BSP_P0 = 0000 0000 BSP_P1 = 0000 0000 BSP_P2 = 0000 0000 BSP_P3 = a ₇ a ₆ a ₅ a ₄ a ₃ a ₂ a ₁ 1	<pre>PRF_CRC_INIT[0] = 0 PRF_CRC_INIT[1] = 0 PRF_CRC_INIT[2] = 0 PRF_CRC_INIT[3] = X</pre>
16	2	$x^{16} + a_{15}x^{15} + \dots + a_1x^1 + 1$	$BSP_P0 = 0000 \ 0000$ $BSP_P1 = 0000 \ 0000$ $BSP_P2 = a_7 \ a_6 \ a_5 \ a_4 \ a_3 \ a_2 \ a_11$ $BSP_P3 = a_{15} \ a_{14} \ a_{13} \ a_{12} \ a_{11} \ a_{10}$ $a_9 \ a_8$	<pre>PRF_CRC_INIT[0] = 0 PRF_CRC_INIT[1] = 0 PRF_CRC_INIT[2] = X PRF_CRC_INIT[3] = X</pre>
24	3	$x^{24} + a_{23} x^{23} + \dots + a_1 x^1 + 1$	$BSP_P0 = 0000 \ 0000$ $BSP_P1 = a_7 \ a_6 \ a_5 \ a_4 \ a_3 \ a_2 \ a_11$ $BSP_P2 = a_{15} \ a_{14} \ a_{13} \ a_{12} \ a_{11} \ a_{10}$ $a_9 \ a_8$ $BSP_P3 = a_{23} \ a_{22} \ a_{21} \ a_{20} \ a_{19} \ a_{18}$ $a_{17} \ a_{16}$	<pre>PRF_CRC_INIT[0] = 0 PRF_CRC_INIT[1] = X PRF_CRC_INIT[2] = X PRF_CRC_INIT[3] = X</pre>
32	4	$x^{32} + a_{31} x^{31} + \dots + a_1 x^1 + 1$	$BSP_P = a_7 a_6 a_5 a_4 a_3 a_2 a_1 1$ $BSP_P = a_{15} a_{14} a_{13} a_{12} a_{11} a_{10}$ $a_9 a_8$ $BSP_P = a_{23} a_{22} a_{21} a_{20} a_{19} a_{18}$ $a_{17} a_{16}$ $BSP_P = a_{31} a_{30} a_{29} a_{28} a_{27} a_{26}$ $a_{25} a_{24}$	<pre>PRF_CRC_INIT[0] = X PRF_CRC_INIT[1] = X PRF_CRC_INIT[2] = X PRF_CRC_INIT[3] = X</pre>

### Table 25-8. Register Settings for Different CRCs

Order	PRF_CRC_LEN	CRC	$BSP_Px$	PRF_CRC_INIT
8	1	CRC-8-ATM $x^8 + x^2 + x + 1$	BSP_P0 = 0x00 BSP_P1 = 0x00 BSP_P2 = 0x00 BSP_P3 = 0x07	PRF_CRC_INIT[0] = 0x00 PRF_CRC_INIT[1] = 0x00 PRF_CRC_INIT[2] = 0x00 PRF_CRC_INIT[2] = 0x00 PRF_CRC_INIT[3] = 0xFF
8	1	CRC-8 $x^8 + x^7 + x^6 + x^4 + x^2 + 1$	BSP_P0 = 0x00 BSP_P1 = 0x00 BSP_P2 = 0x00 BSP_P3 = 0xD3	PRF_CRC_INIT[0] = 0x00 PRF_CRC_INIT[1] = 0x00 PRF_CRC_INIT[2] = 0x00 PRF_CRC_INIT[3] = 0xFF
16	2	CRC-16 (used in CC2500) $x^{16} + x^{15} + x^{2} + 1$	BSP_P0 = 0x00 BSP_P1 = 0x00 BSP_P2 = 0x05 BSP_P3 = 0x80	PRF_CRC_INIT[0] = 0x00 PRF_CRC_INIT[1] = 0x00 PRF_CRC_INIT[2] = 0xFF PRF_CRC_INIT[3] = 0xFF
16	2	CRC-16-CCITT x ¹⁶ + x ¹² + x ⁵ + 1	BSP_P0 = 0x00 BSP_P1 = 0x00 BSP_P2 = 0x21 BSP_P3 = 0x10	PRF_CRC_INIT[0] = 0x00 PRF_CRC_INIT[1] = 0x00 PRF_CRC_INIT[2] = 0xFF PRF_CRC_INIT[3] = 0xFF
24	3	CRC-24 $x^{24} + x^{22} + x^{20} + x^{19} + x^{18} + x^{16}$ + $x^{14} + x^{13} + x^{11} + x^{10} + x^8 + x^7 + x^6 + x^3$ + $x + 1$	BSP_P0 = 0x00 BSP_P1 = 0xCB BSP_P2 = 0x6D BSP_P3 = 0x5D	PRF_CRC_INIT[0] = 0x00 PRF_CRC_INIT[1] = 0xFF PRF_CRC_INIT[2] = 0xFF PRF_CRC_INIT[3] = 0xFF
32	4	CRC-32-IEEE 802.3 $x^{32} + x^{26} + x^{23} + x^{22} + x^{16} + x^{12} + x^{11} + x^{10} + x^8 + x^7 + x^5 + x^4 + x^2 + x + 1$	BSP_P0 = 0xB7 BSP_P1 = 0x1D BSP_P2 = 0xC1 BSP_P3 = 0x04	PRF_CRC_INIT[0] = 0xFF PRF_CRC_INIT[1] = 0xFF PRF_CRC_INIT[2] = 0xFF PRF_CRC_INIT[3] = 0xFF

Table 25-9. Register Settings for Some Commonly Used CRCs, Assuming Initialization With All 1s

### 25.4.4 Coprocessor Mode

The coprocessor mode is used to run the BSP as a stand-alone and not part of the signal path. It must not be used while the LLE is running. Coprocessor mode is selected by setting BSP_MODE.CP_MODE to 01 or 11. In these modes, one byte to be processed is written to the BSP_DATA register, and the result of processing this byte can later be read back from the same register. When BSP_MODE.CP_MODE is 01, the coprocessor is in receive mode, where the whitener is applied before the CRC. When BSP_MODE.CP_MODE is 11, the coprocessor is in transmit mode, where the whitener is applied after the CRC.

To apply the BSP operations to a byte, write it to the BSP_DATA register. When this register is written to, the BSP_MODE.CP_BUSY bit goes high.

If CP_MODE.CP_END is 0, the first bit provided is the LSB and the last bit is the MSB. If CP_MODE.CP_END is 1, the first bit provided is the MSB and the last bit is the LSB.

When BSP_MODE.CP_BUSY goes low, the processed data can be read from the BSP_DATA register. If one or both whiteners are enabled, this byte is whitened or de-whitened. Otherwise, it is the same as the byte written, except if the CRC is being read as described in the following text.

To read out a CRC in transmit mode, set BSP_MODE.CP_READOUT to 1. A zero must be written to the BSP_DATA register, and when BSP_MODE.CP_BUSY goes low, a CRC byte can be read from BSP_DATA. This should be repeated for each CRC byte. If whitening is enabled, the read back CRC bytes are whitened.

The BSP must not be set in coprocessor mode while the LLE is processing a packet.



### 25.5 Frequency and Channel Programming

For normal transmit and receive tasks, the carrier frequency is set by using the PRF_CHAN.FREQ register. The carrier frequency is 2379 + n MHz, where *n* is the value of this register, and *n* can be from 0 to 116. This gives a frequency range from 2379 MHz to 2495 MHz. Note that this frequency range extends beyond the ISM band.

If PRF_CHAN.FREQ is set to 127, and for the Rx and Tx test commands, the frequency must be programmed directly in hardware registers. In this case, the synthesizer frequency is set by programming the 7-bit frequency word located in FREQCTRL.FREQ[6:0]. The synthesizer frequency is given by 2379 + n MHz, where n is the value of FREQCTRL.FREQ[6:0], and is programmable in 1-MHz steps. The device supports synthesizer frequencies in the range from 2379 MHz to 2495 MHz. The usable settings for FREQ[6:0] is consequently 0 to 116.

In Rx, the system operates on a low intermediate frequency (IF) of 1 MHz for data rates up to 1 Mbps, and on a zero IF for 2 Mbps. In Tx, the system supports operating on low IF or zero IF. The IF to be used for Tx can be programmed in the register MDMTEST1.TX_TONE. The receiver may operate on a positive or negative IF when the data rate is 1 Mbps and lower; this is controlled with MDMTEST1.RX_IF.

When the symbol rate is 1 Mbps or lower and the LLE programs the frequency, it uses a  $\pm$ 1 MHz IF on Tx. For both Rx and Tx, a negative IF is used when PRF_CHAN.FREQ < 62, and a positive IF is used when PRF_CHAN.FREQ < 62

When the symbol rate is 2 Mbps and the LLE programs the frequency, it uses an IF on Tx as specified in  $PRF_RADIO_CONF.TXIF$ . This IF may be zero, or ±1 MHz, ±2 MHz, or ±3 MHz. The recommended setting is ±1 MHz. A negative IF is used when  $PRF_CHAN.FREQ < 62$ , and a positive IF is used when  $PRF_CHAN.FREQ < 62$ , and a positive IF is used when  $PRF_CHAN.FREQ > 62$ .

For all data rates, the setting of MDMCTRL1.PHASE_INVERT is taken into account by the LLE when finding the setting for MDMTEST1.TX_TONE. The FREQCTRL register is programmed corresponding to the programmed IF in order to operate on the channel specified by PRF_CHAN.FREQ.

### 25.6 Modulation Formats

The CC2541 supports GFSK and MSK modulation formats. For GFSK modulation, the deviation can be set to 160 kHz or 250 kHz (320 kHz or 500 kHz for 2 Mbps). The data rate can be set to 250 kbps, 500 kbps, 1 Mbps, or 2 Mbps. The desired modulation scheme is set in the MDMCTRL0.MODULATION register.

Not all combinations of modulation format, data rate and deviation are supported. Table 25-10 gives an overview of supported combinations.

Modulation Format	Data Rate	Deviation	MDMCTRL0.MODULATION
GFSK	2 Mbps	500 kHz	0011
GFSK	2 Mbps	320 kHz	0111
GFSK	1 Mbps	250 kHz	0010
GFSK	1 Mbps	160 kHz	0110
GFSK	250 kbps	160 kHz	0100
MSK	500 kbps	_	1001
MSK	250 kbps	-	1000

Table 25-10. Supported Modulation Formats, Data Rates, and Deviations

### 25.7 Receiver

When the receiver is started, it searches for the preamble and the sync word. These are used for frequency offset compensation and bit and byte synchronization. The sync word can be programmed to be from 16 to 32 bits.

Checking the sync word is done in a two-stage process. First, a correlation value is calculated. If this



correlation is above a programmable threshold, a data decision of the received sync word is done. It can be programmed in MDMCTRL3.SYNC_MODE whether this data decision is to be ignored, no bit errors are to be accepted, or one bit error is to be accepted. The correlation threshold value is programmed in MDMCTRL1.CORR_THR. This threshold value should depend on the sync word length. As a rule of thumb, a value of 0.25 times the number of bits (rounded down) can be used.

For the bit synchronization to work well, some guidelines should be followed for the sync word. It should have enough transitions, but not long runs of 10 1010... or other short, repeated patterns. Generally, a longer sync word gives better performance.

The CC2541 devices have support for two independent sync words. The primary and secondary sync words are specified in two sets of registers. The secondary sync word can be enabled by the SW_CONF.DUAL_RX bit, and if enabled, the received signal is correlated against both sync words. If the correlation with one of the sync words is above the threshold, data decision is done against that sync word.

While the receiver is running, a received signal strength indicator (RSSI) is updated. The RSSI is available some time after the receiver is started, regardless of whether sync is found. It can be read from the RSSI register, which is 0x80 when no RSSI is available. The value given is in the range 0 to approximately 64, with a change of 1 corresponding to a 1-dB change. The offset from a true dBm value depends on the receiver mode and can be found in the device data sheet. For high received signal levels, the reported RSSI saturates at one of the highest possible reported values. The accuracy and update time of the RSSI can be traded off using MDMTEST0.RSSI_ACC. The RSSI can be calculated over a window of 5.33  $\mu$ s or 21.3  $\mu$ s, and 1, 2, or 4 such windows can be averaged to give the result. Using a longer average time gives higher accuracy, but it takes longer before a result is ready, and doing the average over a longer time means that the result may be wrong for short packets. An average of *n* windows of length *t*_{RSSI}should only be used for packets lasting longer than (*n* + 1) *t*_{RSSI} (including preamble, sync word, and CRC).

The receiver must run dc offset estimation and removal. The dc offset estimation mode can be controlled with MDMTEST0.DC_BLOCK_MODE. For data rates of 1 Mbps and lower, where the receiver runs on a low IF, it is recommended to use the default setting for this register (continuous estimation). For 2 Mbps, where the receiver runs on zero IF, delayed dc offset estimation should normally be used. This causes the dc offset estimation to be done in front of the packet. The delay can be controlled through MDMTEST0.DC_BLOCK_LENGTH and MDMTEST1.DC_DELAY. The recommendation is to set MDMTEST0.DC_BLOCK_LENGTH to 11 (128 samples) and MDMTEST1.DC_DELAY to 00 (5 delays), which allows for up to approximately 105  $\mu$ s of energy in front of the packet payload, including the preamble and sync word. As an alternative for 2 Mbps, dc offset estimation can be turned off, and a previously found value can be used, written into the DC_I_L, DC_I_H, DC_Q_L, and DC_Q_H registers. Values can be found in advance, but differ for each frequency. For auto acknowledgments and other packets that are received at a known time, the LLE can perform a special dc offset algorithm as described in Section 25.9.2.

### 25.8 Packet Format

The packet format is configurable. There are two operation modes for radio packet control, basic mode and auto mode. Of these, only auto mode supports automatic acknowledgment and retransmissions. The LLE-controlled part of the packet format is also different for the two modes. In basic mode, there is an optional length field followed by an optional address of 1 byte, as shown in Figure 25-6. In auto mode, there is a 9-bit or 10-bit header field containing length and sequence number information. This format is shown in Figure 25-7. The figures show the packet formats with their configurability. The fields with a header in gray are controlled directly by the modem and are used in the acquisition of received packets. The fields with header in white are controlled by the LLE.

Preamble	Sync word	Length	Address	Payload	CRC	
1–16 bytes	16–32 bits	0 1 byto	0 1 byte	0.255 bytes	0 4 bytes	
Handled by modem		0–1 byte	0–1 byte	0–255 bytes	0–4 bytes	

R0009-01

### Figure 25-6. Air Interface Packet Format for Basic Mode



Preamble	Sync word	Address	Header	Payload	CRC I
1–16 bytes	16–32 bits	0–1 byte	9–10 bits	0–127 bytes	0–4 bytes
Handle	d by modem	o i byto	3-10 bits	0 121 byteo	0 4 byteo
					R0010-02

Figure 25-7. Air Interface Packet Format for Auto Mode

The preamble is a sequence of 1010 1010 or 0101 0101. It can be from 1 to 16 bytes. The type of preamble and the number of bytes can be set up in the MDMCTRL2 register.

The Sync word field is a synchronization word that can have any length from 16 to 32 bits. The length is programmed in the SW_CONF.SW_LEN register. The sync word itself is programmed in the SW0, SW1, SW2, and SW3 registers for the primary sync word, and SW4, SW5, SW6, and SW7 for the secondary sync word. The bit ordering of the sync word is set up with MDMCTRL2.SW_BIT_ORDER. If SW_BIT_ORDER is 0, the LSB of SW0 (SW4) is transmitted first and the MSB of SW3 (SW7) is transmitted last. If SW_BIT_ORDER is 1, the MSB of SW3 (SW7) is transmitted first and the LSB of SW0 (SW4) is transmitted last. The first bit transmitted is always the same regardless of the sync word length; the unused bits for sync word length of less than 32 bits are the ones that would have been transmitted last.

The optional length byte in basic mode (see Figure 25-6) is present if PRF_TASK_CONF.MODE = 01. It indicates the number of address and payload bytes following the length byte. If the length field is not present, the length is fixed as described in Section 25.9.2.

The optional address is 1 byte if present; the length is configured with the PRF_PKT_CONF.ADDR_LEN register. In the transmitter, the address can be used for identification or to direct the message to a particular receiver, and in the receiver, the address can be used to filter out messages from unknown or unwanted transmitters and to distinguish between messages from different transmitters. See Section 25.9.2 for details on how the address is used. Note that for the packet format in Figure 25-7 or if a length field is not used, the address field immediately follows the sync word, and can thus be seen as an extension of it.

The 9-bit or 10-bit header shown in Figure 25-7 is shown in more detail in Figure 25-8 and Figure 25-9. This field consists of a 6-bit or 7-bit length followed by a 2-bit sequence number and a flag called NO_ACK (NOA in Figure 25-8 and Figure 25-9) to inform that acknowledgment of the packet is not expected. If the configuration is to use a fixed length, the value of the length field is ignored in the receiver. It can be configured always to set the length field to 110011 in the transmitter for fixed-length packets.

Length						SE	EQ	NOA
Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB

R0011-01

### Figure 25-8. Bits of 9-Bit Header

Length						SE	EQ	NOA	
Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB

R0012-01

### Figure 25-9. Bits of 10-Bit Header

The payload can be from zero to 255 bytes in basic mode, but the sum of the number of address and payload bytes must not exceed 255. In auto mode, the payload can be from 0 to 63 bytes with a 9-bit header or 0 to 127 bytes with a 10-bit header. The maximum packet length can be limited, see Section 25.9.2.3.1 and Section 25.9.2.3.2



#### Packet Format

www.ti.com

The bit ordering when transmitting the length, address, payload, and CRC bytes is set up with the ENDIANNESS bit of the FRMCTRL0 register; if 0, the LSB of each byte is transmitted first and if 1, the MSB is transmitted first. Normally, FRMCTRL0.ENDIANNESS and MDMCTRL2.SW_BIT_ORDER should have the same value. Note that for correct operation in auto mode, FRMCTRL0.ENDIANNESS must be set to 1 so that MSB is transmitted first.

The CRC field contains 0 to 4 bytes and is used to check the packet for errors if present. See Section 25.4.3 on how to set up the CRC generation and checking.

### 25.8.1 Rx FIFO Packet Organization

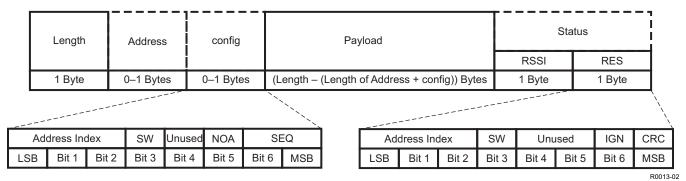


Figure 25-10. Structure of Packets in the Rx FIFO

The structure of a packet in the Rx FIFO is shown in Figure 25-10. All packets start with a length byte, regardless of whether a length byte is present on the air. The length is the number of bytes in the address, config, and payload fields following the length byte, and it may be modified compared to the length received on the air or configured as fixed-length. If packets are longer than what can fit in the FIFO, packets must be read from the FIFO while reception takes place, either by DMA or directly by the MCU. The auto-flush options in PRF_FIFO_CONF cannot be used in this case, and auto-commit and auto-deallocate must be enabled for the Rx FIFO in RFFCFG.

The address byte is placed after the length byte and is present if configured in PRF_FIFO_CONF.RX_ADDR_CONF. The address is written in the FIFO as it was received on the air.

The config byte following the length byte and address byte is present if configured in PRF_FIFO_CONF.RX_ADDR_CONF. In this case, the index *n* to the PRF_ADDR_ENTRYn containing the received address is present in bits 0–2, and bit 3 is 0 if the primary sync word was received and 1 if the secondary sync word was received. In auto mode, the 3 MSBs of the config byte are set to the 3 LSBs of the received header.

The payload is as received on the air. In case of an empty packet, there is no payload.

The status field consists of 2 bytes appended to the FIFO entry if configured in

PRF_FIFO_CONF.RX_STATUS_CONF. The presence of a status field is not reflected in the value of the length byte, so if a status field is present, the MCU must read 2 extra bytes. It is possible to configure this even using DMA with automatic length extraction. The status bytes are:

- · RSSI is the received signal-strength indication from the demodulator.
- · RES contains information on the address and CRC result.
  - The 3 LSBs contain the address index as in the config byte.
  - Bit 3 is 0 if the primary sync word was received and 1 if the secondary sync word was received.
  - IGN is 1 for packets that may be ignored by the MCU due to repeated sequence number and 0 otherwise.
  - CRC is 1 if there was a CRC error and 0 otherwise.

### 25.8.2 Tx FIFO Packet Organization

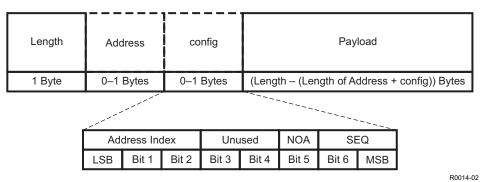


Figure 25-11. Structure of Packets in the Tx FIFO

The structure of a packet in the Tx FIFO is shown in Figure 25-11. All packets start with a length byte, regardless of whether a length byte is present on the air. The length is the number of bytes in the address, config, and payload fields following the length byte, and it may be modified before being transmitted on the air. If a fixed length is used, it is up to the MCU to ensure that the length is correct given the fixed length expected by the receiver. If packets are longer than what can fit in the FIFO, packets must be written to the FIFO while transmission takes place, either by DMA or directly by the MCU. Auto-commit and auto-deallocate must then be enabled for the Tx FIFO in RFFCFG.

The address byte is placed after the length byte and is present if configured in

PRF_FIFO_CONF.TX_ADDR_CONF. If it is included, the address is transmitted on the air as it is read from the FIFO. If it is not included, but a config byte is included, the three LSBs of the config byte tell the index *n* of PRF_ADDR_ENTRYn from which the address is inserted. If neither an address nor a config byte is included, the address is inserted from PRF_ADDR_ENTRY0.ADDRESS.

The config byte following the length byte and optional address byte is present if configured in PRF_FIFO_CONF.TX_ADDR_CONF. This byte contains an address index which is used to determine the address if no address byte is included as explained previously. If an address byte is included, the address index is used to determine which address entry to read the configuration from, but the ADDRESS field in that address entry is ignored. In auto mode, the NO_ACK bit (LSB) of the transmitted header is set to bit 5 of the config byte. If PRF_ADDR_ENTRYn.CONF.FIXEDSEQ, where *n* is the index of the address used, is 1, the SEQ field of the transmitted header is taken from the SEQ field (bits 6–7) of the config byte; otherwise, the sequence number on the air is inserted from PRF_ADDR_ENTRYn.SEQSTAT.SEQ. If the config byte is not included, the NO_ACK bit is always sent as 0 and

PRF_ADDR_ENTRYn.CONF.FIXEDSEQ should be 0 (otherwise the SEQ field always remains 0).The payload is transmitted as present in the FIFO.

### 25.8.3 Tx Buffers for ACK Payload

The hardware Tx FIFO is not used for ACK payload in Rx tasks in auto mode. Instead, an acknowledgment packet for each address can be placed in one of two dedicated buffers for that address. These two buffers constitute a FIFO capable of holding two packets. The buffers for the first two addresses are placed in the RAM page normally used for the hardware Tx FIFO. These four buffers can either be accessed from the Tx FIFO space at 0x6100 or by selecting page 7 through RFRAMCFG, but the Tx FIFO registers should not be used. The other twelve buffers must be addressed from the configurable radio memory bank through the RFRAMCFG register. The mapping of each buffer is shown in Table 25-11.

Address Entry Number	Buffer Number	Setting of RFRAMCFG	Start Address
0	0	7 or X	0x6000 <i>or</i> 0x6100
0	1	7 or X	0x6020 or 0x6120
1	0	7 or X	0x6040 or 0x6140
1	1	7 or X	0x6060 <i>or</i> 0x6160
2	0	1	0x6000

Table 25-11. Segments for Holding ACK Payload for Each Address Entry

Address Entry Number	Buffer Number	Setting of RFRAMCFG	Start Address
2	1	1	0x6020
3	0	1	0x6040
3	1	1	0x6060
4	0	2	0x6000
4	1	2	0x6020
5	0	2	0x6040
5	1	2	0x6060
6	0	3	0x6000
6	1	3	0x6020
7	0	3	0x6040
7	1	3	0x6060

Table 25-11. Segments for Holding ACK Payload for Each Address Entry (continued)

The status of buffer *k* for address *n* is contained in the PRF_ADDR_ENTRYn.ACKLENGTHk register. If the value is 0, the buffer is free.

In order to enter a payload for address *n*, the MCU must follow the following procedure:

- 1. Read PRF_ADDR_ENTRYn.ACKLENGTH0 and PRF_ADDR_ENTRYn.ACKLENGTH1. Call the values len_0 and len_1, respectively.
- 2. Read PRF_ADDR_ENTRYn.SEQSTAT.NEXTACK and call this value k. Let m be NOT k (i.e., 1 k).
- 3. Check if len_k is 0. If so, write the payload to buffer k for address entry n (see Table 25-11), then write the payload length to PRF_ADDR_ENTRYn.ACKLENGTHk. End the procedure.
- 4. Otherwise, check whether len_ *m* is 0. If so, write the payload to buffer *m* for address entry *n* (see Table 25-11), then write the payload length to PRF_ADDR_ENTRYn.ACKLENGTHm. End the procedure.
- 5. Otherwise, no ACK payload buffer for that address is free, and no payload can be entered at this time.

The ACK payload length can be 1–32. When a buffer becomes free, the LLE writes the PRF_ADDR_ENTRYn.ACKLENGTHk to 0 and raises a TXDONE interrupt.

A buffer contains only the payload to be transmitted. The length is given by PRF_ADDR_ENTRYn.ACKLENGTHk, and the address and sequence number are as described in Section 25.9.2.3.2.

In order to flush the buffers for address *n*, issue the command CMD_FLUSH_ACK *n* (see Table 25-12). This causes the LLE to write PRF_ADDR_ENTRYn.ACKLENGTH0 and PRF_ADDR_ENTRYn.ACKLENGTH1 to 0 and clear PRF_ADDR_ENTRYn.SEQSTAT.ACK_PAYLOAD_SENT. If no task is running, the LLE takes SEMAPHORE1; if it fails, it does not write to PRF_ADDR_ENTRYn.SEQSTAT.ACK_PAYLOAD_SENT. If the transmission of an acknowledgment with payload had started on that address, flushing happens after the transmission is finished. After the flushing is done, the LLE raises a TXFLUSHED interrupt.

### 25.9 Link Layer Engine

The link layer engine controls radio operation. It is started by setting the LLECTRL.LLE_EN bit to 1. The LLE must be started before the radio can be operated.

The LLE can be reset by clearing and setting LLECTRL.LLE_EN. The LLE should not be reset while the radio is active. The MCU should not enter PM1, PM2, or PM3 while the LLE is running a task. The mode of the LLE is selected with LLECTRL.LLE_MODE_SEL. For the proprietary-mode operation described in this chapter, this field must be written to 00. For BLE operation, this field is 01; that value should only be written by the TI BLE stack. In order to switch modes, the LLE must be reset; writing to LLECTRL.LLE_MODE_SEL while LLECTRL.LLE_EN is 1 has no effect.



### 25.9.1 Command Register

The command register RFST can be used for sending commands to the LLE and the FIFOs. Commands in the range 0x80-0xFF are commands to the FIFOs, see Section 25.3.1. Other commands are commands to the LLE.

The commands are listed in Table 25-12. There are commands for starting receive and transmit modes. In addition, there is a command CMD_SHUTDOWN to stop the radio operation and end the task directly. The commands CMD_SEND_EVENT1 and CMD_SEND_EVENT2 do the same action as receiving a Timer 2 event 1 or event 2

If an unknown command is entered, the LLE responds by generating an LLEERR interrupt. If a task is running, it stops.

When sending a command to the LLE, the RFST register retains its value until the LLE has received the command (but not necessarily executed it) and then is set to 0. Commands should not be sent to the LLE unless RFST is 0.

Number	Command Name	Description
0x01	CMD_SHUTDOWN	Stop operation immediately
0x02	CMD_DEMOD_TEST	Start demodulator without sync search
0x03	CMD_RX_TEST	Start demodulator and sync search
0x04	CMD_TX_TEST	Start transmitter and transmit zeros
0x05	CMD_TX_FIFO_TEST	Start transmitter and transmit from Tx FIFO
0x06	CMD_PING	Respond with a PINGRSP interrupt
0x08	CMD_RX	Start receive operation
0x09	CMD_TX	Start transmit operation
0x0A	CMD_TX_ON_CC	Start transmit operation on clear channel
0x0B	CMD_STOP	Gracefully stop radio task
0x21	CMD_SEND_EVENT1	Do the same action as if Timer 2 event 1 was observed
0x22	CMD_SEND_EVENT2	Do the same action as if Timer 2 event 2 was observed
0x30	CMD_FLUSH_ACK0	Flush the ACK payload buffers for address 0
0x31	CMD_FLUSH_ACK1	Flush the ACK payload buffers for address 1
0x32	CMD_FLUSH_ACK2	Flush the ACK payload buffers for address 2
0x33	CMD_FLUSH_ACK3	Flush the ACK payload buffers for address 3
0x34	CMD_FLUSH_ACK4	Flush the ACK payload buffers for address 4
0x35	CMD_FLUSH_ACK5	Flush the ACK payload buffers for address 5
0x36	CMD_FLUSH_ACK6	Flush the ACK payload buffers for address 6
0x37	CMD_FLUSH_ACK7	Flush the ACK payload buffers for address 7

### Table 25-12. Commands From MCU to LL Engine via RFST Register

### 25.9.2 Radio Tasks

Before starting a task, radio registers should be set up with the desired packet format, and the desired input sensitivity and output power should be programmed. Furthermore, the sync word in use must be programmed in the SW0, SW1, SW2, and SW3 registers. If a secondary sync word is used, it must be programmed in the SW4, SW5, SW6, and SW7 registers. The RAM registers must be programmed to configure the task. The way the task runs depends on the PRF_TASK_CONF register. The operation mode is set up by the MODE bits of this register. A value of 00 or 01 gives basic mode and thus disables auto ACK or auto retransmission. A value of 10 or 11 gives auto mode where auto acknowledgment or auto retransmission can be enabled per the address in PRF_ADDR_ENTRYn.CONF.AA.

All tasks start with a start-of-task command from the MCU. The LLE takes SEMAPHORE0 at this time; if the semaphore is not available, the task ends with an error. Depending on the configuration in PRF_TASK_CONF.START_CONF, the LLE either starts the task immediately or waits for a Timer 2 event 1 before starting. Note that a Timer 2 event 1 may be pending from before the LLE starts waiting; in that case, the task starts immediately. To clear a pending Timer 2 event 1, reset the LLE. To prevent



Link Layer Engine

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unwanted events from reaching the LLE, Timer 2 event 1 can be disabled in the T2EVTCFG register; see Chapter 22. The frequency word is programmed according to the setting of PRF_CHAN.FREQ, except if it is 127, in which case no frequency programming is done and any value written by the MCU is retained. The LLE starts configuring the transmitter or receiver, depending on the type of task. After the transmitter or receiver has been set up, the LLE takes SEMAPHORE1 to gain access to the remaining RAM-based registers, read the parameters, and start transmission or reception.

Programming of frequency is done as described in Section 25.5. For symbol rates of 1 Mbps and lower, Rx and Tx are done on the same synthesizer frequency, whereas for a symbol rate of 2 Mbps, the synthesizer frequency changes between Rx and Tx. This change is done without a recalibration of the synthesizer.

At the end of a packet, the LLE reads the RSSI register and writes the value to the PRF_LAST_RSSI register and, if so configured, to the RSSI byte of the Rx FIFO. This read is done after the next-to-last byte has been obtained from the demodulator. Note that for a bit rate of 2 Mbps and for sync words shorter than 32 bits, MDMCTRL3.RSSI_MODE should be set to 11 to ensure a correct reading. Before turning off the demodulator, the LLE reads the dc offset from the DC_I_L, DC_I_H, DC_Q_L, and DC_Q_H registers and writes the result to PRF_LAST_DCOFF (in the byte order listed for the register read). The LLE also reads the frequency offset from the FREQEST register and writes the result to PRFX_LAST_FREQEST (see Table 25-7).

If PRF_RADIO_CONF.DCOFF is 1, the LLE runs a procedure that estimates the dc offset right after receiver startup. This mode is suitable for packets that are known to be received at a certain time, such as acknowledgment packets. In this mode, the LLE starts the receiver with normal dc cancellation mode and forces the LNA gain to minimum. After a short time, the LLE reads out the value of the dc offset estimate, writes it into the override registers, and selects manual override mode for dc offset estimation. It sets the LNA gain back to the programmed value and after a waiting time to allow the LNA to stabilize, starts sync search. The time to start Rx with this mode is the same as for ordinary start of Rx.

If PRF_RADIO_CONF.DCWB is 1, the LLE writes the dc offset estimate read out at the end of the packet into the dc offset override register, provided that the received packet did not have a CRC error. This is suited for the delayed dc offset mode, where the override value for dc offset is used before a delayed dc offset is available.

Some of the RAM registers are checked by the LLE to verify that their values are permitted. This applies to PRF_CHAN.FREQ, PRF_FIFO_CONF.TX_ADDR_CONF, and PRF_CRC_LEN. If any of these registers have values that are not permitted, the task ends with an error.

A CMD_SHUTDOWN command, undefined command, or any command starting a new task, ends the task immediately. If a packet was being transmitted or received, an RXTXABO interrupt to the MCU is raised. If a CMD_STOP command is received, the task ends after the current reception or transmission is done. Timer 2 event 2 can be configured to end a task: If PRF_TASK_CONF.STOP_CONF is 01, Timer 2 event 2 behaves as a CMD_STOP, and if PRF_TASK_CONF.STOP_CONF is 10, Timer 2 event 2 behaves as a CMD_STOP, and if PRF_TASK_CONF.STOP_CONF is 10, Timer 2 event 2 behaves as a CMD_SHUTDOWN. Setting PRF_TASK_CONF.STOP_CONF to 00 disables Timer 2 event 2 as a stop event. With the 11 setting, Timer 2 event 2 only applies to sync search or listen right after a CMD_RX or CMD_TX_ON_CC (this setting is not meaningful for a CMD_TX task) or a start by Timer 2 event 1. This is explained in later subsections.

Timer 2 may capture the time of a packet based on the setting in PRF_RADIO_CONF. The fields TXCAP and RXCAP decide how capture is configured for Tx and Rx, respectively; see Table 25-13. The captured value can be read from the registers T2M0, T2M1, T2MOVF0, T2MOVF1, and T2MOVF2 when t2_cap and t2ovf_cap are selected using the T2MSELregister; see Chapter 22.

TXCAP	Description		
00	Capture of transmitted packets off		
01	Capture the start (after the sync word) of every transmitted packet		
10	Capture the end of every transmitted packet		
11	Capture the start of the first transmitted packet, i.e., capture of transmitted packets is turned off after a packet has been transmitted.		
RXCAP	Description		

### Table 25-13. Timer 2 Capture Settings



00	Capture of received packets off
01	Capture the start (after the sync word) of every received packet
10	Capture the end of every received packet
11	Capture the start of the first received packet, i.e., capture of received packets is turned off after a packet has been fully received.

#### Table 25-13. Timer 2 Capture Settings (continued)

When capture is done at the beginning of a packet, the time captured is the time right after the sync word has been received or transmitted. Setting TXCAP or RXCAP to 11 enables capture at the start of a packet, but the capture is turned off after a packet has been transmitted or fully received in a task, so it is the start of the first packet in the task that is captured. The MCU should normally only read the captured value after a task is done; otherwise, the captured value may be overwritten with a new value. The user must take into account that a timer value may be captured on a received packet that does not match the address or that has a length which is not permitted, and that is thus not reported. It is possible to turn on capture for both received and transmitted packets in the same task. If so, it is up to the user to determine if the captured value was from a received or transmitted packet.

When a task is finished, the LLE writes an end-of-task cause in PRF_ENDCAUSE, frees the semaphores, raises a TASKDONE interrupt, and halts its operation. The possible values of PRF_ENDCAUSE are listed in Table 25-14.

If PRF_CHAN.SYNTH_ON is 1, the synthesizer is not turned off after the task ends. This can be used to start a new task immediately on the same channel and get faster start of Rx or Tx. To do so, the next task should be started with PRF_CHAN.FREQ set to 127. Note that the synthesizer should not be allowed to run for a long time after a task has ended, as this causes excessive power consumption. The synthesizer can be stopped by sending a CMD_SHUTDOWN command.

Number	Name	Description		
Normal task ending				
0	TASK_ENDOK	Task ended normally		
1	TASK_RXTIMEOUT	Timer 2 event 2 or CMD_STOP observed while waiting for Rx sync		
2	TASK_NOSYNC	Sync was not obtained in the specified time		
3	TASK_NOCC	TX_ON_CC ended because channel was not clear		
4	TASK_MAXRT	Task ended because maximum number of retransmissions was reached		
5	TASK_STOP	Task ended after transmission or reception by Timer 2 event 2 or CMD_STOP while transmitting or receiving or with ACK or retransmission in progress		
6	TASK_ABORT	Task aborted by command		
MCU interface error				
255	TASKERR_INTERNAL	Internal program error		
254	TASKERR_CMD	Unknown command		
253	TASKERR_SEM	Unable to obtain semaphore		
252	TASKERR_PAR	Unpermitted parameter		
251	TASKERR_TXFIFO	Tx FIFO without available data when not permitted		
250	TASKERR_RXFIFO	Overfull Rx FIFO in Tx task		
249	TASKERR_MODUNF	Modulator underflow observed		

#### Table 25-14. End-of-Task Causes



Link Layer Engine

#### 25.9.2.1 AGC Algorithm

If PRF_PKT_CONF.AGC_EN is 1, an automatic gain control (AGC) algorithm is run while the receiver is looking for sync. The AGC algorithm switches between two different front-end gain settings in the LNAGAIN register. It is recommended to use AGC when running on 2 Mbps to improve the saturation performance.

Parameters for control of the AGC algorithm are found in page 5 of the radio RAM; see Table 25-7. This table lists reset values that the LLE sets for these parameters.

The LLE polls the RSSI value at every update and compares it to the values of PRFX_RSSI_LIM_LOWER and PRFX_RSSI_LIM_UPPER. If the observed RSSI is below PRFX_RSSI_LIM_LOWER, the LNA gain is set to the high gain setting. If the observed RSSI is above PRFX_RSSI_LIM_UPPER, the LNA gain is set to the low gain setting. If the observed RSSI is between these limits, the LNA gain is not changed.

The high gain to use is the value found in the LNAGAIN register when the task is started. The low gain to use is the value found in the PRFX_LNAGAIN_SAT RAM register.

The PRFX_RSSI_LIM_LOWER and PRFX_RSSI_LIM_UPPER values must differ in order to account for the difference that is observed from the RSSI register when the LNA gain is changed and to have hysteresis to avoid too-frequent gain changes.

When sync is obtained on the receiver, the AGC algorithm stops updating the LNA gain, which remains at the value last set. When the receiver is switched off, the LNAGAIN register is set back to the value it had when the task started, i.e., the high gain setting.

When the gain is reduced during the reception of a packet, the value found in the PRF_LAST_RSSI register and (if configured) in the RSSI byte of the Rx FIFO is updated to reflect this. This update is done by adding the value of the register PRFX_RSSI_DIFF to the value found in the RSSI register. PRFX_RSSI_DIFF should therefore contain the difference between the RSSI offset for the two LNA gain settings in use, available from the device data sheet. Note that the hardware RSSI register is not updated this way.

For the AGC algorithm to operate correctly, it requires some signal, having the same power as the packet, transmitted in the band in front of the packet. That signal can be extra preamble bytes or tone. The length required for this signal depends on the RSSI accuracy setting in MDMTEST0.RSSI_ACC, see Section 25.7. An average of *n* windows of length  $t_{RSSI}$  requires the extra signal to last at least  $(n + 1) t_{RSSI}$ . Extra preamble bytes can be set up using MDMCTRL2.NUM_PREAM_BYTES. Note that the extra signal required comes in addition to the 1 preamble byte always used in a packet. When adding extra preamble bytes, this must be accounted for in PRF_TX_DELAY, PRF_RETRANS_DELAY and PRF_RX_TX_TIME. The Rx requires  $n \times t_{RSSI}$  extra time to start when using the AGC. In the dc offset estimation, the extra signal must be accounted for when setting the delay.

#### 25.9.2.2 Tone in Front of Packet

In order to get the transmission format to resemble that of other vendors, a tone may be sent in front of the preamble. This tone can be used by the AGC algorithm on the receiver side. If PRF_PKT_CONF.START_TONE is 1, such a tone is transmitted as a replacement of the first preamble bytes. This means that this feature must only be used in combination with increasing the number of preamble bytes. The tone lasts for a time given by the RAM register PRFX_TONE_DURATION. In order to get a smooth transition from tone to preamble, it is recommended to set PRFX_TONE_DURATION as given in Table 25-15.

Tone transmission is allowed to coincide with the synthesizer stabilizing (this may, however, cause the start of the tone to have larger frequency variations than the packet). For this reason, when  $PRF_PKT_CONF.START_TONE$  is 1, the synthesizer startup time is reduced by the value of the register  $PRFX_TONE_OFFSET$ . This should normally correspond to the time of the extra preambles, but it must not be larger than 4096 (corresponding to 128 µs).  $PRFX_TONE_OFFSET$  can thus be used to compensate for the extra time added by the extra preamble bytes used for tone generation. However, the duration of the extra preamble bytes configured must be accounted for in  $PRF_TX_DELAY$ ,  $PRF_RETRANS_DELAY$ , and  $PRF_RX_TX_TIME$ .

The default values of PRFX_TONE_DURATION and PRFX_TONE_OFFSET correspond to 48 µs and are tuned for using 12 extra preamble bytes (13 in total) on 2 Mbps. When using the reset values, MDMCTRL2.NUM_PREAM_BYTES should thus be set to 0x0C.

If PRFX_TONE_DURATION is set too large compared to the number of preamble bytes configured, the modulator underflows. If this happens, the task ends with TASKERR_MODUNF as end cause.

Data Rate	PRFX_TONE_DURATION	PRFX_TONE_OFFSET
2 Mbps	MDMCTRL2.NUM_PREAM_BYTES × 0x80 + 0x4A	MDMCTRL2.NUM_PREAM_BYTES × 0x80
1 Mbps	MDMCTRL2.NUM_PREAM_BYTES × 0x100 + 0x52	MDMCTRL2.NUM_PREAM_BYTES × 0x100
500 kbps	MDMCTRL2.NUM_PREAM_BYTES × 0x200 + 0x62	<pre>min(MDMCTRL2.NUM_PREAM_BYTES × 0x200, 0x1000)</pre>
250 kbps	MDMCTRL2.NUM_PREAM_BYTES × 0x400 + 0x82	<pre>min(MDMCTRL2.NUM_PREAM_BYTES × 0x400, 0x1000)</pre>

Table 25-15	Recommended	<b>DAM</b> Pogistor	Sottings for	Start Tono
Table 25-15.	Recommended	RAW REGISLE	Settings for	Start rone

### 25.9.2.3 Receive Task

When a CMD_RX command is received, the LLE configures the radio on the channel given by PRF_CHAN.FREQ and starts listening for a sync word.

The LLE can set up an internal time-out for the sync search in the PRF_SEARCH_TIME register. If this register is non-zero and no sync has been obtained in the number of 32 MHz cycles given by this register, the task ends with a TASK_NOSYNC end cause. Note that the value of this register must be large enough to have time for the duration of the sync word and one preamble byte, in addition to some margin, in order to get sync. The task can also be set up to end on Timer 2 event 2, based on

PRF_TASK_CONF.STOP_CONF. If this bit field is 11, the Timer 2 event 2 time-out applies only during the first sync search after a CMD_RX command has been issued if PRF_TASK_CONF.START_CONF is 0. In this case, the time-out in PRF_SEARCH_TIME does not apply to the first sync search, but it still applies to subsequent sync searches in the same task. If PRF_TASK_CONF.STOP_CONF is 11 and PRF_TASK_CONF.START_CONF is 1, the time-out applies to every sync search and PRF_SEARCH_TIME

never applies, but the Timer 2 event 2 timeout does not apply after sync is obtained or while waiting for Timer 2 event 1 to restart listening. If sync is obtained, the LLE starts reading the packet.

If sync is found on a packet, the time of sync is captured by the Timer 2 capture function (see Section 22.1.10).

### 25.9.2.3.1 Basic Mode

This section describes the receive operation if PRF_TASK_CONF.MODE is 00 or 01.

If PRF_TASK_CONF.MODE is 01, the length byte is read first. It gives the number of bytes between the length byte and the CRC, including the address. If the length is too small to contain the address, the reception of the packet is stopped and the device goes back to sync search (regardless of the setting in PRF_TASK_CONF.REPEAT).

Next, the address byte is read if PRF_PKT_CONF.ADDR_LEN is 1. It is compared against PRF_ADDR_ENTRYn.ADDRESS for the values of *n* where the entry is enabled for the received sync word. If there is a matching entry, this entry is used when receiving the packet; otherwise, reception is stopped and the device goes back to sync search. If PRF_PKT_CONF.ADDR_LEN is 0, the first entry that is enabled for the received sync word is used. If PRF_TASK_CONF.MODE is 00, the packet length is then read from PRF_ADDR_ENTRYn.RXLENGTH. This length includes the address, so it must be greater than or equal to the number of address bytes. If PRF_TASK_CONF.MODE is 01, the received length byte is compared against PRF_ADDR_ENTRYn.RXLENGTH. If it is greater than that value, reception is stopped and the device goes back to sync search.

If reception is stopped due to address mismatch or invalid length, the time-out given by PRF_SEARCH_TIME or Timer 2 event 2 still applies. If the first packet of the task is being received and PRF_TASK_CONF.STOP_CONF is 11, the next packet still counts as the first packet.



If a CRC field is present, it is checked using the polynomial configured in the BSP and the initialization value from PRF_CRC_INIT. The result of the CRC check is written in the MSB of the RES byte in the status field if a status field is configured. If the CRC is not correct and PRF_FIFO_CONF.AUTOFLUSH_CRC is 1, the LLE sends a discard Rx FIFO command to remove the packet from the Rx FIFO.

A packet where the length is equal to the address size contains no payload. Such a packet is known as an empty packet. If PRF_FIFO_CONF.AUTOFLUSH_EMPTY is 1 and an empty packet is received, the LLE sends a discard Rx FIFO command to remove the empty packet from the Rx FIFO.

Note that if the CRC length is 1 byte, the CRC check is not correct for empty packets if fixed length is configured or no address bytes are used.

If the Rx FIFO becomes full while receiving a packet, the packet is discarded from the FIFO and no more bytes are stored in the Rx FIFO, but the packet is received to its end. After that, it is checked whether the packet would be discarded from the Rx FIFO anyway due to the setting of PRF_FIFO_CONF. If so, the task proceeds as normally. Otherwise, an RXFIFOFULL error interrupt is raised in lieu of the normal interrupt for received packets.

After receiving a packet, the LLE raises an interrupt to the MCU. Depending on CRC result and whether the packet was empty, the interrupts are generated as shown in Table 25-16, provided an RXFIFOFULL interrupt is not raised as described previously. The table also shows which of the counters among the RAM registers are to be updated.

CRC Result	Payload Length > Address Length	Counter Incremented	Interrupt Raised
ОК	No	PRF_ADDR_ENTRYn.N_RXOK	RXEMPTY
ОК	Yes	PRF_ADDR_ENTRYn.N_RXOK	RXOK
NOK	No	PRF_ADDR_ENTRYn.N_RXNOK	RXNOK
NOK	Yes	PRF_ADDR_ENTRYn.N_RXNOK	RXNOK

 Table 25-16. Interrupt and Counter Operation for Received Messages

An address entry should not be modified while the receiver is running. In order to modify, stop the receiver, modify the entry or entries, and restart the receiver.

### 25.9.2.3.2 Auto Mode

This section describes the receive operation if PRF_TASK_CONF.MODE is 10 or 11.

If PRF_PKT_CONF.ADDR_LEN is 1, the address byte is compared against PRF_ADDR_ENTRYn.ADDRESS, where *n* ranges from 0 to 7. It is compared against PRF_ADDR_ENTRYn.ADDRESS for the values of *n* where the entry is enabled for the received sync word. If there is a matching entry, this entry is used when receiving the packet, otherwise reception is stopped and the device goes back to sync search. If PRF_PKT_CONF.ADDR_LEN is 0, the first entry that is enabled for the received sync word is used.

Next, the 9-bit or 10-bit header is read. If PRF_ADDR_ENTRYn.CONF.VARLEN is 1, the length is fetched from the header and compared against PRF_ADDR_ENRYn.RXLENGTH. If it is greater than that value, reception is stopped and the device goes back to sync search. If PRF_ADDR_ENTRYn.CONF.VARLEN is 0, the length field in the received header is ignored and the packet length is read from PRF_ADDR_ENTRYn.RXLENGTH. In both cases, the length is the number of bytes after the header and before the CRC. The length must be less than or equal to 63 for a 9-bit header and 127 for a 10-bit header. When a 10-bit header is used, the MCU must ensure that an entire packet can fit in the Rx FIFO for auto ACK to be possible. This limits the maximum packet size based on the settings in PRF_FIFO_CONF.

If reception is stopped due to address mismatch or invalid length, the time-out given by PRF_SEARCH_TIME or Timer 2 event 2 still applies. If the first packet of the task is being received and PRF_TASK_CONF.STOP_CONF is 11, this still counts as the first packet.

If a CRC field is present, it is checked using the polynomial configured in the BSP and the initialization value from PRF_CRC_INIT. The result of the CRC is written in the MSB of the RES byte in the status field if a status field is configured. If the CRC is not correct and PRF_FIFO_CONF.AUTOFLUSH_CRC is 1, the LLE sends a discard Rx FIFO command to remove the packet from the Rx FIFO.



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321

If the CRC is correct, the sequence number is checked against the sequence number stored in PRF_ADDR_ENTRYn.SEQSTAT.SEQ. If the sequence numbers are equal and PRF_ADDR_ENTRYn.SEQSTAT.VALID is 1, the two last received CRC bytes are compared against the 2 bytes in PRF_ADDR_ENTRYn.LASTCRC. If they are equal, the packet is determined to be a retransmission which can be ignored. If the CRC is 1 byte only, the received CRC byte is compared to PRF_ADDR_ENTRYn.LASTCRC[0] only, and if there is no CRC, the comparison is always viewed as equal. If the packet was a retransmission, the IGN bit of the RES byte in the status field is set if a status field is configured. After reception of a packet with CRC OK and which fit in the Rx FIFO, PRF_ADDR_ENTRYn.SEQSTAT.VALID is set to 1, PRF_ADDR_ENTRYn.SEQSTAT.SEQ is set to the sequence number of the header of the received packet, and PRF_ADDR_ENTRYn.LASTCRC is set to the value of the last two received CRC bytes.

If the Rx FIFO becomes full while receiving a packet, the packet is discarded from the FIFO and no more bytes are stored in the Rx FIFO, but the packet is received to its end. After that, it is checked whether the packet would be discarded from the Rx FIFO anyway due to the setting of PRF_FIFO_CONF. If so, the task proceeds as normally. Otherwise, an RXFIFOFULL error interrupt is raised, and no acknowledgment is transmitted. The sequence number is not updated so that a retransmission of the packet is not ignored.

If the received packet was not a retransmission and PRF_ADDR_ENTRYn.SEQSTAT.ACK_PAYLOAD_SENT is 1, the packet is seen as a confirmation of the last transmitted acknowledgment payload. If so, PRF_ADDR_ENTRYn.SEQSTAT.ACK_PAYLOAD_SENT is set to 0, a TXDONE interrupt is raised, and the PRF_ADDR_ENTRYn.NTXDONE counter is incremented.PRF_ADDR_ENTRYn.ACKLENGTHk is set to 0 for the *k* found in PRF_ADDR_ENTRYn.SEQSTAT.NEXTACK, and PRF_ADDR_ENTRYn.SEQSTAT.NEXTACK is inverted.

After receiving a packet, the LLE raises an interrupt to the MCU. Depending on the CRC result, the payload length, and whether the received packet is a retransmission to be ignored, the interrupts are generated as shown in Table 25-17. The table also shows which of the counters among the RAM registers are to be updated.

CRC Result	Ignore	Length	Counter Incremented	Interrupt Raised
ОК	No	> 0	PRF_ADDR_ENTRYn.N_RXOK	RXOK
ОК	No	= 0	PRF_ADDR_ENTRYn.N_RXOK	RXEMPTY
ОК	Yes	Х	PRF_ADDR_ENTRYn.N_RXIGNORED	RXIGNORED
NOK	Х	Х	PRF_ADDR_ENTRYn.N_RXNOK	RXNOK

Table 25-17. Interrupt and Counter Operation for Received Messages

After reception of a packet, the next action is determined as follows:

- If the CRC of the received packet was not correct, the treatment of the packet is finished and the next action is as described in Section 25.9.2.3.3.
- If PRF_ADDR_ENTRYn.CONF.AA is 0, the treatment of the packet is finished and the next action is as described in Section 25.9.2.3.3.
- If the NO_ACK bit of the received header is 1 and the CRC was correct, the treatment of the packet is finished and the next action is as described in Section 25.9.2.3.3.
- If the packet did not fit in the Rx FIFO and was not otherwise to be discarded, the treatment of the packet is finished and the next action is as described in Section 25.9.2.3.3.
- Otherwise, an acknowledgment is transmitted as described in the following text.

After receiving a packet where the CRC is correct and where an acknowledgment is supposed to be sent, the transmitter is configured. The transmission starts at a time given by the PRF_RX_TX_TIME register. Synthesizer recalibration is performed only if there is time. The LLE checks

PRF_ADDR_ENTRYn.SEQSTAT.NEXTACK to find *k*. If PRF_ADDR_ENTRYn.ACKLENGTHk is nonzero, payload is included the packet. In this case, PRF_ADDR_ENTRYn.SEQSTAT.ACK_PAYLOAD_SENT is set to 1 by the LLE. The transmitted packet has the same sync word and address as the received packet. If PRF_ADDR_ENTRYn.CONF.TXLEN is 0, the length field in the header is set equal to PRF_ADDR_ENTRYn.ACKLENGTHk. If PRF_ADDR_ENTRYn.CONF.TXLEN is 1, the length field is set to 11 0011 for a 9-bit header and to 011 0011 for a 10-bit header. Note that a value of 0 for



Link Layer Engine

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PRF_ADDR_ENTRYn.CONF.TXLEN may be used regardless of the VARLEN setting in the peer device, as the length field is ignored for fixed length. A value of 1 must only be used if the peer is configured to use fixed length for the ACK payload, and should only be used with ACKs without payload. The sequence number is set to the value of PRF_ADDR_ENTRYn.SEQSTAT.ACKSEQ, and NO_ACK is set to 0. If there is payload, it is read from the buffer as described in Section 25.8.3.

After the acknowledgement has been transmitted, PRF_ADDR_ENTRYn.SEQSTAT.ACKSEQ is incremented modulo 4, the PRF_N_TX counter is incremented, and the next action is as described in Section 25.9.2.3.3.

### 25.9.2.3.3 Continuation and Ending of Receive Tasks

When a task ends, a TASKDONE interrupt is raised and an end cause is then available in PRF_ENDCAUSE.

After a packet has been received and potentially an acknowledgment has been transmitted, the next action depends on PRF_TASK_CONF.REPEAT. If this value is 0, the task ends. In this case, the PRF_ENDCAUSE register is set to TASK_ENDOK.

If PRF_TASK_CONF.REPEAT is 1, reception restarts. If PRF_TASK_CONF.START_CONF is 1, the LLE behaves as if the task was started again, with the LLE waiting for Timer 2 event 1; then starting to listen. If PRF_TASK_CONF.START_CONF is 0, the receiver restarts as soon as possible, as starting a new task (except for the behavior of Timer 2 event 2 if PRF_TASK_CONF.STOP_CONF is 11). In both cases, synthesizer recalibration is done if PRF_TASK_CONF.REPEAT_CONF is 0, otherwise not. Skipping synthesizer recalibration reduces the time before listening is restarted.

If a CMD_SHUTDOWN or a command starting a new task is observed while the task is running, it ends immediately with TASK_ABORT as the end cause. If the receiver or transmitter was running, an RXTXABO interrupt is also raised.

If CMD_STOP is received while in sync search, the task ends immediately with TASK_RXTIMEOUT as the end cause. If CMD_STOP is received while receiving or while transmitting an ACK or in the transition between those, the task ends with TASK_STOP as the end cause after the packet is fully received and (if ACK is to be sent) the ACK is sent. If CMD_STOP is received while waiting for Timer 2 event 1 to restart reception, the task ends immediately with TASK_STOP as the end cause

If Timer 2 event 2 (either from Timer 2 or from CMD_SEND_EVENT2) is observed during the task, the behavior depends on PRF_TASK_CONF.STOP_CONF:

- 00: Nothing happens
- 01: Behaves as if a CMD_STOP was received
- 10: Behaves as if a CMD_SHUTDOWN was received
- 11: If received while in sync search for the first packet after the task was started, or if PRF_TASK_CONF.START_CONF is 1 while in sync search for any packet, the task ends immediately with TASK_RXTIMEOUT as the end cause. Otherwise, nothing happens.

In addition, the task can end due to an internal time-out as described in the beginning of Section 25.9.2.1, or it can end due to an error condition. The full list of possible end causes is summarized in Table 25-18.

Condition	End-of-Task Cause	Comment
Received packet (and potentially sent ACK) with PRF_TASK_CONF . REPEAT = 0	TASK_ENDOK	
Received packet (and potentially sent ACK) with PRF_TASK_CONF.REPEAT = 1 after having observed CMD_STOP or Timer 2 event 2 with PRF_TASK_CONF.STOP_CONF = 01 and PRF_TASK_CONF.REPEAT = 1	TASK_STOP	
While in sync search, observed CMD_STOP or Timer 2 event 2 with PRF_TASK_CONF . STOP_CONF = 01	TASK_RXTIMEOUT	
Observed Timer 2 event 2 while in sync search of the first packet with PRF_TASK_CONF . STOP_CONF = 11	TASK_RXTIMEOUT	

#### Table 25-18. End-of-Receive Tasks

Condition	End-of-Task Cause	Comment
Did not get sync in the time specified by PRF_SEARCH_TIME	TASK_NOSYNC	
Received command for starting new task or CMD_SHUTDOWN or observed Timer 2 event 2 with PRF_TASK_CONF . STOP_CONF = 10	TASK_ABORT	If transmitter was running or receiver was running and had obtained sync, an RXTXABO interrupt is also raised.
Received unknown command	TASKERR_CMD	LLEERR interrupt is also raised. If transmitter was running or receiver was running and had obtained sync, an RXTXABO interrupt is also raised.
Semaphore is not free when expected	TASKERR_SEM	Task ends without any radio operation. LLEERR interrupt is also raised.
Unpermitted value of RAM register	TASKERR_PAR	LLEERR interrupt is also raised.
For PRF_TASK_CONF . MODE = 00: PRF_ADDR_ENTRYn . RXLENGTH corresponding to the received address is smaller than address length	TASKERR_PAR	LLEERR interrupt is also raised.
For auto mode: PRF_ADDR_ENTRYn.ACKLENGTGHm for the ACK payload to be transmitted exceeded 32	TASKERR_PAR	LLEERR interrupt is also raised.

### Table 25-18. End-of-Receive Tasks (continued)

### 25.9.2.4 Transmit Task

When a CMD_TX command is received, the LLE configures the radio on the channel given by PRF_CHAN.FREQ and starts transmitting the packet from the Tx FIFO.

If the Tx FIFO has no available data, the task ends with TASKERR_TXFIFO as the end cause. Otherwise, the number of bytes given by the length byte in the Tx FIFO is read from the Tx FIFO and transmitted or otherwise handled as described in following sections. No check of data availability is done after the length byte is read, so if the FIFO contains fewer bytes than indicated in the length field, a Tx FIFO underflow interrupt is raised by the FIFO hardware.

### 25.9.2.4.1 Basic Mode

This section describes the transmit operation if PRF_TASK_CONF.MODE is 00 or 01.

If PRF_TASK_CONF.MODE is 01, the length field is calculated from the length field in the FIFO and transmitted. It is up to the MCU to ensure that the calculated length field does not exceed 255. If PRF_TASK_CONF.MODE is 00, no length field is transmitted.

If an address is configured, it is found based on the setting in PRF_FIFO_CONF.TX_ADDR_CONF. It can be set to take the address from PRF_ADDR_ENTRY0, to read it from the Tx FIFO (which for the transmitter is equivalent to not having an address configured), or to read an index *n* from the config byte in the FIFO and read the address from PRF_ADDR_ENTRYn.ADDRESS. The values of ENA0 and ENA1 in PRF_ADDR_ENTRYn.CONF are ignored for the transmitter; the primary sync word is always transmitted.

The payload (if any) is transmitted as given in the FIFO.

If configured, a CRC with the number of bytes given by PRF_CRC_LEN is transmitted at the end.

When a packet has been transmitted, the LLE sends a deallocate Tx FIFO command if PRF_ADDR_ENTRYn.REUSE is 0. Otherwise, the MCU must issue either a deallocate Tx FIFO (to send a new packet) or a retry Tx FIFO (to reuse) before sending again. The PRF_N_TX counter is incremented. A TXDONE interrupt to the MCU is raised. The next action is as given in Section 25.9.2.3.3.

### 25.9.2.4.2 Auto Mode

This section describes the transmit operation if PRF_TASK_CONF.MODE is 10 or 11.



Link Layer Engine

When a 10-bit header is used, the MCU must ensure that an entire packet can fit in the Tx FIFO for auto retransmission to be possible. This limits the maximum packet size based on the settings in PRF_FIFO_CONF.

If an address is configured, it is the first byte transmitted. It is found based on the setting in PRF_FIFO_CONF.TX_ADDR_CONF. It can be set to take the address from PRF_ADDR_ENTRY0, to read it from the Tx FIFO, or to read an index *n* from the config byte in the FIFO and read the address from PRF_ADDR_ENTRYn.ADDRESS. In other cases, *n* is always be assumed to be 0 in the following text. The values of ENA0 and ENA1 in PRF_ADDR_ENTRYn.CONF are ignored for the transmitter; the primary sync word is always transmitted.

The 9-bit or 10-bit header is transmitted next. If PRF_ADDR_ENTRYn.CONF.TXLEN is 0, the length field is set to the number of payload bytes after the header, which is calculated from the length byte in the Tx FIFO. If PRF_ADDR_ENTRYn.CONF.TXLEN is 1, the length field is set to 11 0011 for a 9-bit header and to 011 0011 for a 10-bit header. Note that a value of 0 for PRF_ADDR_ENTRYn.CONF.TXLEN may be used regardless of the VARLEN setting in the receiver, as a receiver configured to use fixed length ignores the length field. A value of 1 must only be used if the receiver is configured to use fixed length. The NO_ACK bit transmitted is set according to bit 5 of the config byte read from the Tx FIFO if present, otherwise to 0. If PRF_ADDR_ENTRYn.CONF.FIXEDSEQ is 1, the SEQ bits transmitted are set equal to bits 6 and 7 of the config byte read from the FIFO. Otherwise, the SEQ bits are set to the value of PRF_ADDR_ENTRYn.SEQSTAT.SEQ.

The payload (if any) is transmitted as given in the FIFO.

If configured, a CRC with the number of bytes given by PRF_CRC_LEN is transmitted at the end.

When a packet has been transmitted, the N_TX counter is incremented.

After transmission of a packet, the action depends on PRF_ADDR_ENTRYn.CONF.AA and the NO_ACK bit in the transmitted header. If PRF_ADDR_ENTRYn.CONF.AA = 0 or NO_ACK = 1, no acknowledgment is expected, and the action is as if a valid acknowledgment had been received.

If PRF_ADDR_ENTRYn.CONF.AA is 1 and the transmitted NO_ACK bit was 0, the LLE configures Rx to listen for an acknowledgment. To listen for acknowledgment, the receiver is configured at a time given by the PRF_TX_RX_TIME register. Synthesizer recalibration is performed only if there is time. The unit looks for sync. The sync search times out at the time given by PRF_SEARCH_TIME. If sync is found, the packet is received into the Rx FIFO. If PRF_PKT_CONF.ADDR_LEN is 1, the address byte is compared against PRF_ADDR_ENTRYn.ADDRESS for the *n* that was used in transmission. If not matching, reception is stopped.

Next, the 9-bit or 10-bit header is read. If PRF_ADDR_ENTRYn.CONF.VARLEN is 1, the length is fetched from the header and compared against PRF_ADDR_ENTRYn.RXLENGTH. The maximum allowed value of this register is 32. If the received length is greater than PRF_ADDR_ENTRYn.RXLENGTH, reception is stopped and the device goes back to sync search. If PRF_ADDR_ENTRYn.CONF.VARLEN is 0, the length field in the received header is ignored and the packet length is read from PRF_ADDR_ENTRYn.RXLENGTH, which should normally be 0 in this case. The length is the number of bytes after the header and before the CRC.

If a CRC field is present, it is checked using the polynomial configured in the BSP and the initialization value from PRF_CRC_INIT. The result of the CRC is written in the MSB of the RES byte in the status field if a status field is configured. If the CRC is not correct and PRF_FIFO_CONF.AUTOFLUSH_CRC is set, the LLE sends a discard Rx FIFO command to remove the packet from the Rx FIFO.

The received sequence number is written to the config byte of the Rx FIFO if configured, but is otherwise ignored.

If the Rx FIFO goes full while receiving an acknowledgment packet, the packet is discarded from the FIFO and no more bytes are stored in the Rx FIFO, but the packet is received to its end. After that, it is checked to see whether the packet would be discarded from the Rx FIFO anyway due to the setting of PRF_FIFO_CONF. If so, the task proceeds as normally. Otherwise, the task ends after the packet is received and an RXFIFOFULL error interrupt is raised. In this case, the treatment of the packet is as if the acknowledgment were not successfully received. This means that the next time a transmit task is started, the packet is retransmitted so that the receiver retransmits the ACK payload.



After receiving an acknowledgment, the LLE raises an interrupt to the MCU. Depending on the CRC result, the payload length, and whether the received packet had the same sequence number as the transmitted one, the interrupts are generated as shown in Table 25-19. It also shows which of the counters among the RAM registers that are to be updated.

CRC Result	Length	Counter Incremented	Interrupt Raised
OK	> 0	PRF_ADDR_ENTRYn.N_RXOK	RXOK
ОК	= 0	PRF_ADDR_ENTRYn.N_RXIGNORED	RXEMPTY
NOK	Х	PRF_ADDR_ENTRYn.N_RXNOK	RXNOK

Table 25-19. Interru	pt and Counter C	peration for Received	ACK Packets
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If an acknowledgment was not received (because no sync was obtained in time, the address did not match, the sequence number was wrong, the CRC check failed, or the ACK did not fit in the Rx FIFO and was not otherwise to be discarded) the LLE sends a retry Tx FIFO command. If the number of retransmissions already performed (not including the original transmission) is equal to PRF_RETRANS_CNT, the task ends. Otherwise, the packet is retransmitted. The time from the end of the previous transmission to the start of the retransmission is given in units of 62.5 ns by PRF_RETRANS_DELAY.

If the received packet was a valid acknowledgment, or if a packet was transmitted and no acknowledgment was expected, the LLE sends a deallocate Tx FIFO command if PRF_ADDR_ENTRYn.REUSE is 0. Otherwise, the MCU must issue either a deallocate Tx FIFO (to send a new packet) or a retry Tx FIFO (to reuse) before sending again. The PRF_ADDR_ENTRYn.NTXDONE counter is incremented. A TXDONE interrupt to the MCU is raised. If PRF_ADDR_ENTRYn.CONF.FIXEDSEQ = 0, PRF_ADDR_ENTRYn.SEQSTAT.SEQ is incremented by 1 modulo 4. The next action is as given in Section 25.9.2.3.3.

If the task ends because of a maximum number of retransmissions, a retry Tx FIFO command is sent before the task ends, and PRF_ADDR_ENTRYn.SEQSTAT.SEQ is not incremented. This means that by default, the packet retransmission is attempted in the next task. If this is not desired, the packet must be removed from the FIFO. This can be done either by issuing a CMD_TXFIFO_RESET (this also removes any subsequent packets in the Tx FIFO), by reading out the packet using the RFTXFRD register and issuing a CMD_TX_FIFO_DEALLOC command, or by Tx FIFO pointer manipulation (Section 25.3.1.3). PRF_ADDR_ENTRYn.SEQSTAT.SEQ should then be incremented by one. These operations should only take place between tasks (i.e., while the LLE does not have SEMAPHORE1).

# 25.9.2.4.3 Continuation and Ending of Transmit Tasks

When a task ends, a TASKDONE interrupt is raised and an end cause is then available in PRF_ENDCAUSE.

After a packet has been transmitted and potentially a valid acknowledgment has been received, the next action depends on PRF_TASK_CONF.REPEAT. If this value is 0, the task ends. In this case, the PRF_ENDCAUSE register is set to TASK_ENDOK.

If PRF_TASK_CONF.REPEAT is 1, the Tx FIFO status is checked. If the Tx FIFO has no available data, the task ends with TASK_ENDOK as the end cause. Otherwise, transmission restarts. If PRF_TASK_CONF.START_CONF is 1, it behaves as if the task was started again with the LLE waiting for Timer 2 event 1, then performing a synthesizer calibration and starting to transmit. If PRF_TASK_CONF.START_CONF is 0, the transmitter restarts PRF_TX_DELAY after the end of the previously transmitted packet, with synthesizer recalibration only if there is enough time, but in other respects as starting a new task. The PRF_TX_DELAY register gives the wait time in units of 62.5 ns. If the value is too small to fulfill, the transmission starts as soon as possible.

If a CMD_SHUTDOWN or a command starting a new task is observed while the task is running, it ends immediately with TASK_ABORT as the end cause. If the transmitter or receiver was running, an RXTXABO interrupt is also raised.



Link Layer Engine

If CMD_STOP is received while transmitting a packet, waiting for transmission of another packet, waiting for an ACK, receiving an ACK, or in the transition between those, the task ends with TASK_STOP as the end cause after the packet is fully transmitted and (if ACK is expected) the ACK is received or given up. If CMD_STOP is received while waiting for Timer 2 event 1 to restart reception, the task ends immediately with TASK_STOP as the end cause.

If Timer 2 event 2 (either from Timer 2 or from CMD_SEND_EVENT2) is observed during the task, the behavior depends on PRF_TASK_CONF.STOP_CONF.

- 00: Nothing happens.
- 01: Behaves as if a CMD_STOP was received
- 10: Behaves as if a CMD_SHUTDOWN was received
- 11: Nothing happens.

In addition, the task can end for reasons described earlier, or it can end due to an error condition. The full list of possible end causes is summarized in Table 25-20.

Condition	End-of-Task Cause	Comment
Transmitted packet (and potentially received ACK) with PRF_TASK_CONF . REPEAT = 0	TASK_ENDOK	
Transmitted packet (and potentially received ACK) and observed Tx FIFO with no available data	TASK_ENDOK	
Transmitted packet (and potentially received ACK) after having observed CMD_STOP or Timer 2 event 2 with PRF_TASK_CONF . STOP_CONF = 01	TASK_STOP	
Did not get valid acknowledgment after having retransmitted the number of times given by PRF_RETRANS_CNT	TASK_MAXRT	
Observed empty Tx FIFO when packet transmission is supposed to start, or Tx FIFO is in an invalid state	TASKERR_TXFIFO	LLEERR and RXTXABO interrupts are also raised.
Rx FIFO went overfull while receiving an ACK that was not otherwise to be discarded	TASKERR_RXFIFO	RXFIFOFULL interrupt is also raised.
Received command for starting new task or CMD_SHUTDOWN or observed Timer 2 event 2 with PRF_TASK_CONF . STOP_CONF = 10	TASK_ABORT	If transmitter was running, an RXTXABO interrupt is also raised.
Received unknown command	TASKERR_CMD	LLEERR interrupt is also raised. If transmitter or receiver was running and had obtained sync, an RXTXABO interrupt is also raised.
Semaphore is not free when expected	TASKERR_SEM	Task ends without any radio operation. LLEERR interrupt is also raised.
Invalid value of RAM register	TASKERR_PAR	LLEERR interrupt is also raised.
Length field to be transmitted exceeded maximum allowed value (255 in basic mode, 127 or 63 in auto mode with variable length)	TASKERR_PAR	LLEERR interrupt is also raised.

#### Table 25-20. End-of-Transmit Tasks

#### 25.9.2.5 Transmit on Clear-Channel Task

When a CMD_TX_ON_CC command is received, the LLE configures the receiver on the channel given by PRF_CHAN.FREQ, but sync search is not enabled. The LLE polls the RSSI register every 5.33 µs and compares it to the value of PRF_RSSI_LIMIT. If a valid RSSI below the value of PRF_RSSI_LIMIT is observed more than PRF_RSSI_COUNT times in a row, the system starts transmitting. From there, the operation is as a normal transmit task, see Section 25.9.2.4, except for the operation after a packet has been transmitted and potentially acknowledged, which is described in Section 25.9.2.5.1.



# 25.9.2.5.1 Continuation and Ending of Transmit on Clear-Channel Tasks

If  $PRF_TASK_CONF.TX_ON_CC_CONF$  is 1, the task ends if a valid RSSI value is not below the limit. If  $PRF_TASK_CONF.TX_ON_CC_CONF$  is 0, the device keeps listening until an RSSI below the given value is found.

If PRF_TASK_CONF.STOP_CONF is 11, Timer 2 event 2 may give a time-out while listening for a clear channel the first time, but not after the first transmission has been started.

If PRF_TASK_CONF.TX_ON_CC_CONF is 0, the clear-channel assessment must not be used as the only medium access control scheme in a multiuser environment, as this may cause all the devices to start transmission at the same time.

If retransmission is enabled, the LLE listens for acknowledgment and retransmits if needed as for normal Tx tasks. However, if PRF_TASK_CONF.REPEAT_CONF is 0 after applying the retransmit delay, the device returns to listening, performing the same operation as when the task started, before possibly retransmitting.

If PRF_TASK_CONF.REPEAT is 0, the task ends after transmission as for normal Tx tasks.

If PRF_TASK_CONF.REPEAT is 1, the Tx FIFO status is checked. If the Tx FIFO has no available data, the task ends with TASK_ENDOK as the end cause. Otherwise, transmission restarts. If PRF_TASK_CONF.REPEAT_CONF is 0, the task returns to listening, whereas if it is 1, the task restarts as if it were a standard transmit task. If PRF_TASK_CONF.START_CONF is 1, it behaves as if the task was started again with the LLE waiting for Timer 2 event 1, then performing a synthesizer calibration and starting to listen or transmit. If PRF_TASK_CONF.START_CONF is 0, the listening or transmission restarts PRF_TX_DELAY after the end of the previously transmitted packet, with synthesizer recalibration only if there is enough time, but in other respects as starting a new task.

If Timer 2 event 2 (either from Timer 2 or from CMD_SEND_EVENT2) is observed during the task, the behavior depends on PRF_TASK_CONF.STOP_CONF.

- 00: Nothing happens.
- 01: If received while transmitting a packet or waiting for or receiving an ACK or in the transition between those, the task ends with TASK_STOP as the end cause after the packet is fully transmitted and (if ACK is expected) the ACK is received or given up. If received while waiting for Timer 2 event 1 to restart reception, the task ends immediately with TASK_STOP as the end cause.
- 10: Behaves as if a CMD_SHUTDOWN was received
- 11: If received while listening for RSSI below the level before the first packet after the task was started, or if PRF_TASK_CONF.START_CONF is 1 while listening before any packet, the task ends immediately with TASK_NOCC as the end cause. Otherwise, nothing happens.

The task can end for all the same reasons as a normal transmit task summarized in Table 25-20. In addition it can end for the reasons listed in Table 25-21.

Condition	End-of-Task Cause	Comment
Observed valid RSSI above the threshold with PRF_TASK_CONF . TX_ON_CC_CONF = 1	TASK_NOCC	
Observed CMD_STOP or Timer 2 event 2 with PRF_TASK_CONF . STOP_CONF = 01 or 11 while listening for RSSI below the threshold	TASK_NOCC	

#### Table 25-21. Additional Reasons for End-of-Transmit on Clear-Channel Tasks

#### 25.9.2.6 Timing

The timing in tasks is given the registers PRF_TX_DELAY, PRF_SEARCH_TIME, and PRF_RETRANS_DELAY, PRF_SEARCH_TIME, PRF_RX_TX_TIME, and PRF_TX_RX_TIME. The first two of these registers are multiplied by 2 and then represent the number of 32-MHz samples, while the rest directly give the number of 32-MHz samples. PRF_TX_DELAY gives the time from the end of the previous transmission in the task to the start of the next transmission. Some examples of these delays are shown in Figure 25-12 and Figure 25-13 for Rx and Tx tasks, respectively. The time from the end of a received packet to the beginning of a transmitted packet is 130 µs in an Rx task with auto ACK.

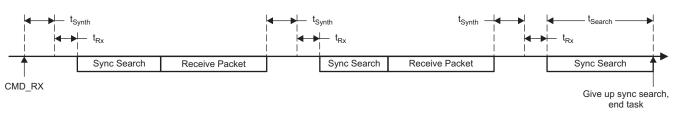


#### Link Layer Engine

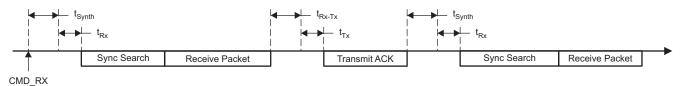
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When sync search takes place, either for receiving a normal packet or for receiving ACK, a time-out can be set up for when to give up the search. This time-out, given in 32-MHz cycles, is set up in the PRF_SEARCH_TIME register. Setting this register to 0 disables the time-out. In case of a time-out, the task ends for a normal sync search, or a packet is retransmitted in case of an ACK sync search.

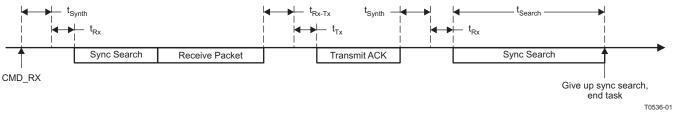
Receive task, PRF_TASK_CONF.MODE = 0X, .START_CONF = 0, .REPEAT = 1:



Receive task, PRF_TASK_CONF.MODE = 1X, .START_CONF = 0, .REPEAT = 1:



Receive task, PRF_TASK_CONF.MODE = 1X, .START_CONF = 0, .REPEAT = 1:



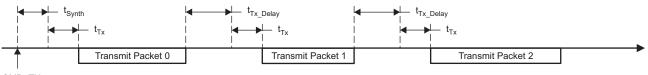
NOTE: The time given by PRF_SEARCH_TIME is denoted  $t_{\text{Search}}$  and the time given by PRF_RX_TX is denoted  $t_{\text{Rx-Tx}}$ . The setup and wait time for the synthesizer, receiver, and transmitter are denoted  $t_{\text{Synth}}$ ,  $t_{\text{Tx}}$ , and  $t_{\text{Rx}}$ , respectively.

# Figure 25-12. Timing of Packets in Rx Tasks

For auto retransmit tasks, the time PRF_RETRANS_DELAY is the time from the end of a transmission to the retransmission of the packet in case an ACK is not found or there is a CRC error; see Figure 25-13. The values of PRF_SEARCH_TIME and the maximum packet length in PRF_PAYLOAD_LEN should be set such that this time can always be achieved. If it is not possible to achieve the retransmission time, the packet is retransmitted as early as possible.

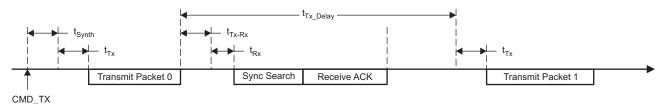


Transmit task, PRF_TASK_CONF.MODE = 0X, .START_CONF = 0, .REPEAT = 1:

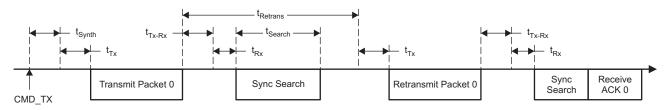


CMD_TX

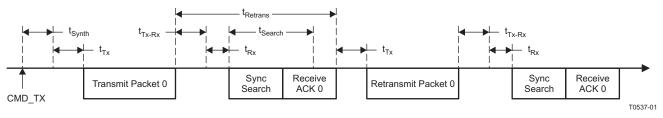
Transmit task, PRF_TASK_CONF.MODE = 0X, .START_CONF = 0, .REPEAT = 1:



Transmit task, PRF_TASK_CONF.MODE = 1X, .START_CONF = 0, .REPEAT = 1, no ACK found first time:



Transmit task, PRF_TASK_CONF.MODE = 1X, .START_CONF = 0, .REPEAT = 1, CRC error on ACK:



NOTE: The time given by PRF_TX_DELAY is denoted  $t_{Tx_Delay}$ , the time given by PRF_SEARCH_TIME is denoted  $t_{Search}$ , the time given by PRF_RETRANS_DELAY is denoted  $t_{Retrans}$ , and the time given by PRF_TX_RX is denoted  $t_{Tx_Rx}$ . The setup and wait times for the synthesizer, receiver, and transmitter are denoted  $t_{Synth}$ ,  $t_{Tx}$ , and  $t_{Rx}$ , respectively.

# Figure 25-13. Timing of Packets in Tx Tasks

# 25.9.3 RF Test Commands

In Table 25-12, there are listed some commands for modem test purposes. No registers are programmed by the LLE based on these commands, so all registers must be set up by the MCU. This includes the frequency word, which is otherwise written by the LLE; see Section 25.5 on how to program this. These commands take no parameters, and do not cause the LLE to create any interrupts or to write any end cause.



#### Random Number Generation

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When receiving CMD_DEMOD_TEST, the LLE starts the receiver, but does not start sync search. The receiver runs until a CMD_SHUTDOWN command (or a command starting another task) is sent.

When receiving CMD_RX_TEST, the LLE starts the receiver and starts sync search. Any received data is discarded. The receiver runs until a CMD_SHUTDOWN command (or a command starting another task) is sent.

When receiving a CMD_TX_TEST, the LLE starts the transmitter and starts sending all zeros. The Tx test command is normally combined with configuration of the modem to send a tone or by the BSP to send a whitening sequence. The transmitter runs until a CMD_SHUTDOWN command (or a command starting another task) is sent. In order to send random modulated data for test purposes, it is recommended to set BSP_MODE to 0x03 to enable both whiteners.

In order to send a continuous wave (CW), MDMCTRL0.TX_IF can be set to 1 before the CMD_TX_TEST command is issued. In this case, the radio outputs a tone with the frequency given in MDMTEST1.TX_TONE. In most cases, a tone at the synthesizer frequency is wanted (e.g., to measure phase noise), in which case MDMTEST.TX_TONE should be set to 0x0A. The frequency synthesizer must be programmed to the carrier frequency with no offset in this case; see Section 25.5.

When receiving a CMD_TX_FIFO_TEST, the LLE starts the transmitter and starts sending from the Tx FIFO; otherwise, the command is as CMD_TX_TEST. The MCU must feed data into the Tx FIFO to avoid underflow, and the Tx FIFO must be set up with auto commit and auto deallocate.

When receiving a CMD_PING command, the LLE responds with a PINGRSP. This can be used for checking that the LLE code is running.

#### 25.10 Random Number Generation

The CC2541 has a hardware pseudo-random register, as explained in Section 14.1. The RF core register bank provides a second interface to this register. Reading the RFPSRND register is equivalent to reading RNDL, then writing 01 to ADCCON1.RCTL.

For seeding the pseudo-random number generator and for tasks where higher entropy of the random numbers is needed, the radio can be used as a true-random generator. The register RFRND provides access to the least-significant bits of the radio ADC, which is random when noise is received. In order to get values on this register, the receiver must be turned on. The value in RFRND is updated every 0.17 µs, so the sampling of that register must be slower than that in order to get a new value with every sample.

To get true random numbers, the following procedure can be followed:

- 1. Program FREQCTRL to a channel that is not likely to contain a narrow-band signal. A frequency outside the ISM band, such as a setting of 0, is recommended.
- 2. Program LNAGAIN to 0 to have minimum reception of a signal on the air.
- 3. Start the receiver in test mode by issuing a CMD_DEMOD_TEST command.
- 4. Wait until ADC data are ready. This can be seen by the RFRND register having a value different from 0.
- Read the number of values needed from RFRND. Make sure that there is at least 0.17 µs between each read (i.e., at least 6 cycles if running on 32 MHz). For instance, to seed the pseudo-random generator, two values are needed.
- 6. Shut down the receiver by issuing a CMD_SHUTDOWN command

The values read from the RFRND register do not have a perfectly uniform distribution. In order to improve this, several random numbers can be combined to produce one random number. One way of doing this is to use the pseudo-random generator in CRC mode and combine 8 numbers into one. An example of how this can be done is given in the C code below:

```
// Store LNA gain setting and set minimum LNA gain
lnagain_stored = LNAGAIN;
LNAGAIN = 0x00;
// Set lowest possible frequency to avoid signals in ISM band
FREQCTRL = 0x00;
// Enable radio in Rx without sync search
while (RFST != 0);
```



```
RFST = CMD_DEMOD_TEST;
// Wait for modem to be running
while (RFRND == 0);
// Seed RNG
RNDL = RFRND;
RNDL = RFRND;
for (j=0; j<ARRAY_SZ; j++) {</pre>
    // Read 8 random bytes into CRC generation
    RNDH = RFRND;
    RNDH = RFRND;
   RNDH = RFRND;
   RNDH = RFRND;
   RNDH = RFRND;
   RNDH = RFRND;
    RNDH = RFRND;
   RNDH = RFRND;
    // Read out LSB of CRC state
    rndarray[j] = RNDL;
}
            // Shut down radio
while (RFST != 0);
RFST = CMD_SHUTDOWN;
// Restore LNA gain setting
LNAGAIN = lnagain_stored;
```

# 25.11 Packet Sniffing

Packet sniffing is a nonintrusive way of observing data that is either being transmitted or received. The packet sniffer outputs a clock and a data signal, which should be sampled on the rising edges of the clock. The two packet-sniffer signals are observable as GPIO outputs. For accurate time stamping, the SFD signal should also be output.

The packet-sniffer mode is selected in register MDMCTRL3.RFC_SNIFF_CTRL; see Table 25-22 for a description of the different modes of operation.

MDMCTRL3.RFC_SNIFF_CTRL	Description		
00	Packet sniffer disabled		
01	Data output from BSP. Tx data, including CRC, is whitened if the whitener is enabled. Rx data, including CRC, is always de-whitened.		
10	Data output from modulator. Only Tx data before whitening is output. Any CRC bytes are 0.		
11	Data output from the demodulator. Only Rx data before de-whitening is output.		

Table 25-22	. Packet-Sniffer	Modes	of	Operation
-------------	------------------	-------	----	-----------

The packet-sniffer clock frequency is equal to the RF data rate. The data is output serially, in the received/transmitted order. It is possible to use a SPI slave to receive the data stream.

When a complete packet has been transferred, the packet sniffer appends a status byte that tells which value of the FREQCTRL register was used and if it was a Tx or Rx packet (bit 0 is high if it was a Tx packet). The appended byte is formatted as follows (first transmitted bit to the left):

FREQ(6)	FREQ(5)	FREQ(4)	FREQ(3)	FREQ(2)	FREQ(1)	FREQ(0)	ТΧ
---------	---------	---------	---------	---------	---------	---------	----

R0015-01

331

# Figure 25-14. Complete Appended Packet

This allows for the external receiver to differentiate between Rx and Tx packets.

To set up the packet-sniffer signals or some of the other RF core observation outputs (in total maximum 3; rfc_obs_sig0, rfc_obs_sig1, and rfc_obs_sig2), the user must perform the following steps:

**Step1**: Determine which signal (RFC_OBS_CTRL[0-2]) to output on which GPIO pin (P1[0:5]). This is done using the OBSSELx control registers (OBSSEL0-OBSSEL5) that control the observation output to the pins P1[0:5] (overriding the standard GPIO behavior for those pins).

**Step2**: Set the  $(RFC_OBS_CTRL[0-2])$  control registers to select the correct signals (rfc_obs_sig); e.g., for packet sniffing one needs the rfc_sniff_data for the packet-sniffer data signal.and rfc_sniff_clk for the corresponding clock signal.

**Step3**: Enable the packet-sniffer module in the MDMCTRL3 register.

# 25.12 Registers

# 25.12.1 Register Overview

#### 25.12.1.1 SFR Registers

- 1 RFIRQF0 (0xE9) RF interrupt flags
- 2 RFIRQF1 (0x91) RF interrupt flags
- 3 RFERRF (0xBF) RF error interrupt flags
- 4 RFD (0xD9) RF data
- 5 RFST (0x6189) LLE and FIFO commands

#### 25.12.1.2 XREG Registers

Address (Hex)	+ 0x0000	+ 0x001	+ 0x002	+ 0x003
0x6180	FRMCTRL0	RFIRQM0	RFIRQM1	RFERRM
0x6184	FREQCTRL	FREQTUNE	TXPOWER	TXCTRL
0x6188	LLESTAT		SEMAPHORE0	SEMAPHORE1
0x618C	SEMAPHORE2	RFSTAT	RSSI	RFPSRND
0x6190	MDMCTRL0	MDMCTRL1	MDMCTRL2	MDMCTRL3
0x6194	SW_CONF	SW0	SW1	SW2
0x6198	SW3	FREQEST	RXCTRL	FSCTRL
0x619C				
0x61A0	LNAGAIN	AAFGAIN	ADCTEST0	
0x61A4		MDMTEST0	MDMTEST1	
0x61A8		ATEST		
0x61AC			RFC_OBS_CTRL0	RFC_OBS_CTRL1
0x61B0	RFC_OBS_CTRL2	LLECTRL		
0x61BC	TXFILTCFG			RFRND
0x61C0	RFRAMCFG			RFFDMA0
0x61C4	RFFDMA1	RFFSTATUS	RFFCFG	
0x61C8	RFRXFLEN	RFRXFTHRS	RFRXFWR	RFRXFRD
0x61CC	RFRXFWP	RFRXFRP	RFRXFSWP	RFRXFSRP
0x61D0	RFTXFLEN	RFTXFTHRS	RFTXFWR	RFTXFRD
0x61D4	RFTXFWP	RFTXFRP	RFTXFSWP	RFTXFSRP

#### Table 25-23. XREG Register Overview

Address (Hex)	+ 0x0000	+ 0x001	+ 0x002	+ 0x003
0x61E0	BSP_P0	BSP_P1	BSP_P2	BSP_P3
0x61E4	BSP_D0	BSP_D1	BSP_D2	BSP_D3
0x61E8	BSP_W	BSP_MODE	BSP_DATA	
0x61F8	SW4	SW5	SW6	SW7
0x61FC	DC_I_L	DC_I_H	DC_Q_L	DC_Q_H

Table 25-23. XREG Register Overview (continued)

# 25.12.2 Register Settings Update

This section contains a summary of the register settings that should be updated from their default value to have optimal performance. For some of the registers, the setting depends on the required gain in the receiver chain for bit rates of 1 Mbps and lower. For 2 Mbps, other values are needed, and different values should be used for Rx and Tx tasks. Note that registers that are listed in only one of the following two tables should have their reset value in the other case.

 
 Table 25-24. Registers That Should Be Updated From Their Default Value, Bit Rates 1 Mbps and Lower

Register Name	Address (Hex)	Standard Gain: New Value (Hex) ⁽¹⁾	High Gain: New Value (Hex)	Description
FRMCTRL0	6180	43	43	Amplitude weight in frequency offset compensation (assuming sync word included in CRC and MSB first)
MDMCTRL1	6191	48	48	Sync word correlation threshold (32-bit sync word)
MDMCTRL2	6192	CO	C0	Use inverse of preamble for frequency offset estimation (assuming MSB first)
MDMCTRL3	6193	63	63	Set RSSI mode to peak detect after sync
RXCTRL	619A	33	3F	Receiver currents
FSCTRL	619B	55	5A	Prescaler and mixer currents
LNAGAIN	61A0	3A	7F	LNA gain
TXFILTCFG	61BC	03	03	Sets Tx anti-aliasing filter to appropriate bandwidth
TXPOWER	6186	E1	E1	Tx power (0 dBm)
TXCTRL	6187	19	19	DAC current
IVCTRL	6265	1B	1B	PA, mixer, and DAC bias

⁽¹⁾ Not all modulation types are characterized for the standard gain setting; see the *CC2541 2.4-GHz* Bluetooth *low energy and Proprietary System-on-Chip* data sheet (SWRS110).

Table 25-25. Registers That Should Be Updated Fron	m Their Default Value, Bit Rate 2 Mbps
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Register Name	Address (Hex)	Rx Tasks	Tx Tasks	Description
FRMCTRL0	6180	43	43	Amplitude weight in frequency offset compensation (assuming sync word is included in CRC and MSB first)
MDMCTRL1	6191	48	48	Sync word correlation threshold (32-bit sync word)
MDMCTRL2	6192	CC	СС	Use inverse of preamble for frequency offset estimation (assuming MSB first); set extra preamble bytes
MDMCTRL3	6193	63	63	Set RSSI mode to peak detect after sync
RXCTRL	619A	29	29	Receiver currents

Register Name	Address (Hex)	Rx Tasks	Tx Tasks	Description					
FSCTRL	619B	5A	5A	Prescaler and mixer currents					
ADCTEST0	61A2	66	66	Reduce ADC gain					
MDMTEST0	61A5	0F	01	Select dc offset compensation method; change RSSI averaging					
TXFILTCFG	61BC	03	03	Transmit filter bandwidth					
PRF_PKT_CONF	6003	06	06	Enable AGC and start tone					
PRF_RADIO_CONF	607E	90	D0	Set 1-MHz Tx IF and dc write-back; for Tx tasks also special dc offset compensation					
TXPOWER	6186	E1	E1	Tx power (0 dBm)					
TXCTRL	6187	19	19	DAC current					
IVCTRL	6265	1B	1B	PA, mixer, and DAC bias					

# Table 25-25. Registers That Should Be Updated From Their Default Value, Bit Rate 2 Mbps (continued)

The values for FRMCTRL0, MDMCTRL2, and PRF_PKT_CONF may require further adjustment based on the frame format, and the correlation threshold in MDMCTRL1 should be adjusted according to the sync word length, see Section 25.7.

In addition to these modifications, registers must be set in order to set up the modulation format, packet handling, etc., as explained throughout this chapter.

# 25.12.3 SFR Register Descriptions

Bit No.	Name	Reset	R/W	Description
7:4	-	0000	R0	Reserved
3	RXTHSHUP	0	R/W0	Rx FIFO goes above its upper threshold. 0: No interrupt pending 1: Interrupt pending
2	TXTHSHUP	0	R/W0	Tx FIFO goes above its upper threshold. 0: No interrupt pending 1: Interrupt pending
1	RXTHSHDN	0	R/W0	Rx FIFO goes below its lower threshold. 0: No interrupt pending 1: Interrupt pending
0	TXTHSHDN	0	R/W0	Tx FIFO goes below its lower threshold. 0: No interrupt pending 1: Interrupt pending

# RFIRQF1 (0x91) – RF Interrupt Flags

Bit No.	Name	Reset	R/W	Description
7	PINGRSP	0	R/W0	When receiving a CMD_PING command, the LLE responds with a PINGRSP. This can be used for checking that the LLE is running. 0: No interrupt pending 1: Interrupt pending
6	TASKDONE	0	R/W0	Task ended 0: No interrupt pending 1: Interrupt pending
5	TXDONE	0	R/W0	Packet transmitted 0: No interrupt pending 1: Interrupt pending
4	RXEMPTY	0	R/W0	Empty packet received 0: No interrupt pending 1: Interrupt pending
3	RXIGNORED	0	R/W0	Packet received with unexpected sequence number 0: No interrupt pending 1: Interrupt pending
2	RXNOK	0	R/W0	Packet received with CRC error 0: No interrupt pending 1: Interrupt pending
1	TXFLUSHED	0	R/W0	Tx ACK buffer flushed 0: No interrupt pending 1: Interrupt pending
0	RXOK	0	R/W0	Packet received correctly 0: No interrupt pending 1: Interrupt pending

### RFERRF (0xBF) - RF Error Interrupt Flags

Bit No.	Name	Reset	R/W	Description
7	-	0	R/W0	Reserved
6	RXFIFOFULL	0	R/W0	Rx FIFO is full when trying to store received data 0: No interrupt pending 1: Interrupt pending
5	LLEERR	0	R/W0	LLE command or parameter error 0: No interrupt pending 1: Interrupt pending
4	RXTXABO	0	R/W0	Receive or transmit operation aborted 0: No interrupt pending 1: Interrupt pending
3	RXOVERF	0	R/W0	Rx FIFO overflow 0: No interrupt pending 1: Interrupt pending
2	TXOVERF	0	R/W0	Tx FIFO overflow 0: No interrupt pending 1: Interrupt pending
1	RXUNDERF	0	R/W0	Rx FIFO underflow 0: No interrupt pending 1: Interrupt pending
0	TXUNDERF	0	R/W0	Tx FIFO underflow 0: No interrupt pending 1: Interrupt pending

#### RFD (0xD9) - RF data

Bit No.	Name	Reset	R/W	Description
7:0	RFD[7:0]	0x00		Data written to the register is written to the Tx FIFO. When reading this register, data from the Rx FIFO is read.



Registers

RFST (	RFST (0xE1) – LLE and FIFO commands						
Bit No.	Name	Reset	R/W	Description			
7:0	RFST[7:0]	0x00	R/WH0	Commands to radio are written to this register. The register is cleared (set to 0x00) when the radio is ready for a new command.			

# 25.12.3.1 XREG Register Descriptions

#### FRMCTRL0 (0x6180) – Frame Control

Bit No.	Name	Reset	R/W	Description
7:5	FOC_MAGN_CONT[2:0]	000	R/W	Controls how signal amplitude is weighted into the frequency offset compensation scheme. 000: Magnitude has no effect 001 to 111: Low-to-high weighting of the magnitude
4:2	-	000	R/W	Reserved always write 0.
1	SW_CRC_MODE	0	R/W	0: The sync word is not included in the crc calculation.
				1: The sync word is included in the crc calculation. Only to be used with whitening disabled.
0	ENDIANNESS	0	R/W	0: The data goes LSB-first over the air.
				1: The data goes MSB-first over the air.

#### RFIRQM0 (0x6181) – RF Interrupt Masks

Bit No.	Name	Reset	R/W	Description
7:4	-	0000	R0	Reserved
3	RXTHSHUP	0	R/W	Rx FIFO goes above its upper threshold. 0: Interrupt disabled 1: Interrupt enabled
2	TXTHSHUP	0	R/W	Tx FIFO goes above its upper threshold. 0: Interrupt disabled 1: Interrupt enabled
1	RXTHSHDN	0	R/W	Rx FIFO goes below its lower threshold. 0: Interrupt disabled 1: Interrupt enabled
0	TXTHSHDN	0	R/W	Tx FIFO goes below its lower threshold. 0: Interrupt disabled 1: Interrupt enabled

# RFIRQM1 (0x6182) – RF Interrupt Masks

Bit No.	Name	Reset	R/W	Description
7	PINGRSP	0	R/W	When receiving a CMD_PING command, the LLE responds with a PINGRSP. This can be used for checking that the LLE is running. 0: Interrupt disabled 1: Interrupt enabled
6	TASKDONE	0	R/W	Task ended 0: Interrupt disabled 1: Interrupt enabled
5	TXDONE	0	R/W	Packet transmitted 0: Interrupt disabled 1: Interrupt enabled
4	RXEMPTY	0	R/W	Empty packet received 0: Interrupt disabled 1: Interrupt enabled
3	RXIGNORED	0	R/W	Packet received with unexpected sequence number 0: Interrupt disabled 1: Interrupt enabled
2	RXNOK	0	R/W	Packet received with CRC error 0: Interrupt disabled 1: Interrupt enabled
1	TXFLUSHED	0	R/W	Tx ACK buffer flushed 0: Interrupt disabled 1: Interrupt enabled
0	RXOK	0	R/W	Packet received correctly 0: Interrupt disabled 1: Interrupt enabled

# RFERRM (0x6183) - RF Error Interrupt Masks

Bit No.	Name	Reset	R/W	Description
7	-	0	R/W	Reserved
6	RXFIFOFULL	0	R/W	RX FIFO is full when trying to store received data. 0: Interrupt disabled 1: Interrupt enabled
5	LLEERR	0	R/W	LLE command or parameter error 0: Interrupt disabled 1: Interrupt enabled
4	RXTXABO	0	R/W	Receive or transmit operation aborted 0: Interrupt disabled 1: Interrupt enabled
3	RXOVERF	0	R/W	Rx FIFO overflow 0: Interrupt disabled 1: Interrupt enabled
2	TXOVERF	0	R/W	Tx FIFO overflow 0: Interrupt disabled 1: Interrupt enabled
1	RXUNDERF	0	R/W	Rx FIFO underflow 0: Interrupt disabled 1: Interrupt enabled
0	TXUNDERF	0	R/W	Tx FIFO underflow 0: Interrupt disabled 1: Interrupt enabled



#### Registers

FREQ	FREQCTRL (0x6184) – Synthesizer Frequency Control						
Bit No.	Name	Reset	R/W	Description			
7	-	0	R0	Reserved. Read as zero			
6:0	FREQ[6:0]	0x16	R/W	Frequency control word. Controls frequency of local oscillator. See Section 25.5 for the relation between the LO frequency and the RF frequency. $f_{1,0} = (2379 + FREQ[6:0])$ MHz			
				The frequency word in FREQ[6:0] is an offset value from 2379 MHz. The			
				device supports the frequency range from 2379 MHz to 2495 MHz. The usable settings for FREQ[6:0] are consequently 0 to 116. Settings outside this range (117–127) give a frequency of 2495 MHz.			

FREQTUNE (0x6185) – Crystal Oscillator Frequency Tuning

Bit No.	Name	Reset	R/W	Description			
7:4	-	0x0	R0	Reserved. Read as zero			
3:0	XOSC32M_TUNE[3:0]	0xF	R/W	Tune crystal oscillator. The default setting of 1111 leaves the XOSC not tuned. Changing setting from default switches in extra capacitance to the oscillator, effectively lowering the XOSC frequency. Hence, a higher setting gives a higher frequency.			

#### TXPOWER (0x6186) – Output Power Control

Bit No.	Name	Reset	R/W	Description
7:0	PA_POWER[7:0]	0xF5	R/W	PA power control. NOTE: This value should be updated. See the device data sheet (Appendix C) for recommended values.

#### TXCTRL (0x6187) - Tx Settings

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved
6	-	1	R/W	Reserved
5:4	DAC_CURR[1:0]	10	R/W	Change the current in the DAC to change the current steps
3:2	DAC_DC[1:0]	10	R/W	Adjusts the dc level to the Tx mixer.
1:0	TXMIX_CURRENT[1:0]	01	R/W	Transmit mixers core current: Current increases with increasing setting.

#### LLESTAT (0x6188) - LLE Status

Bit No.	Name	Reset	R/W	Description
7:5	-	000	R0	Reserved
4	AGC_LOWGAIN	0	R	1 if the AGC algorithm has reduced the front-end gain; 0 otherwise
3	WAIT_T2E1	0	R	Indication on the LLE waiting for Timer 2 event 1 to start a task 0: Not waiting for Timer 2 event 1 1: Command processed, event 1 not yet received
2	LLE_IDLE	0	R	Link-layer engine idle 0: The LLE is busy processing or finishing a command, or in reset. 1: The LLE is idle waiting for a command to start a new task.
1	SYNC_SEARCH	0	R	Rx search for sync 0: The modem is not ready to receive a packet 1: The modem is in search for a sync word or is receiving a packet
0	VCO_ON	0	R	VCO on 0: The VCO is powered down, so the next Rx or Tx operation must start and calibrate the synthesizer before transmitting or receiving 1: The VCO is powered up. If the LLE is idle, it means the next task starts quickly if frequency programming is disabled (PRF_CHAN.FREQ = 127)



#### SEMAPHORE0 (0x618A) – Semaphore for Accessing RF Data Memory

Bit No.	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved, read as 0
0	SEMAPHORE	1	R/W1	When SEMAPHORE = 1 and SEMAPHORE0 is read, SEMAPHORE is set to 0. SEMAPHORE can only be set to 1 by a reset or by writing 1 to it.

#### SEMAPHORE1 (0x618B) – Semaphore for Accessing RF Data Memory

Bit No.	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved, read as 0
0	SEMAPHORE	1	R/W1	When SEMAPHORE = 1 and SEMAPHORE1 is read, SEMAPHORE is set to 0. SEMAPHORE can only be set to 1 by a reset or by writing 1 to it.

# SEMAPHORE2 (0x618C) – Semaphore

Bit No.	Name	Reset	R/W	Description
7:1	-	0000 000	R0	Reserved, read as 0
0	SEMAPHORE	1	R/W1	When SEMAPHORE = 1 and SEMAPHORE2 is read, SEMAPHORE is set to 0. SEMAPHORE can only be set to 1 by a reset or by writing 1 to it.

# RFSTAT (0x618D) - RF Core Status

Bit No.	Name	Reset	R/W	Description
7	MOD_UNDERFLOW	0	R/W0	Modulator has underflowed. Must be cleared by software
6:5	DEM_STATUS	00	R	Demodulator status 00: Idle 01: Active 10: Finishing 11: Error
4	SFD	0	R	High when the sync word has been sent in TX or when sync has been obtained in RX
3	CAL_RUNNING	0	R	Frequency synthesizer calibration status. 0: Calibration done or not started 1: Calibration in progress.
2	LOCK_STATUS	0	R	1 when PLL is in lock; 0 otherwise
1	TX_ACTIVE	0	R	Status signal, active when the LLE is in one of the transmit states
0	RX_ACTIVE	0	R	Status signal, active when the LLE is in one of the receive states

#### RSSI (0x618E) - Received Signal Strength Indicator

Bit No.	Name	Reset	R/W	Description
7:0	RSSI_VAL[7:0]	0x80	R	RSSI estimate on a logarithmic scale. Unit is 1 dB; offset depends on the gain of the Rx chain, including external components; see the device data sheet. The reset value of 0x80 also indicates that the RSSI value is invalid/measurement not yet complete.

#### RFPSRND (0x618F) – Pseudorandom Number Generator

Bit No.	Name	Reset	R/W	Description
7:0	RNG_DOUT[7:0]	0x00	R	The value read from the pseudorandom number generator, see Chapter 14. Reading this register generates causes the shift register to be updated with 13 times rollout.

#### Registers

MDM	CTRL0 (0x6190) – Modem Co	onfiguration		
Bit No.	Name	Reset	R/W	Description
7:6	FOC_DECAY[1:0]	00	R/W	Controls decay ratio of frequency offset compensation mechanism. Value by which to increment preamble cost function at each decay 00: 8 01: 16 10: 32 11: 64
5	TX_IF	0	R/W	0: Modulation is done at an IF set by rfr_tx_tone. 1: Modulator outputs tone set by rfr_tx_tone.
4:1	MODULATION[3:0]	0010	R/W	Modulation scheme 0010: GFSK 250-kHz deviation, 1-Mbps data rate 0011: GFSK 500-kHz deviation, 2-Mbps data rate 0100: GFSK 160-kHz deviation, 250-kbps data rate 0110: GFSK 160-kHz deviation, 1-Mbps data rate 0111: GFSK 320-kHz deviation, 2-Mbps data rate 1000: MSK, 250-kbps data rate 1001: MSK, 500-kbps data rate Others: Reserved
0	PHASE_INVERT	0	R/W	<ul> <li>Set one of two RF modulation modes for Rx / Tx</li> <li>0: Normal (binary 0 represented with negative frequency deviation, binary 1 represented with positive frequency deviation)</li> <li>1: Inverted phase (binary 0 represented with positive frequency deviation, binary 1 represented with negative frequency deviation)</li> </ul>

#### MDMCTRL1 (0x6191) – Modem Configuration

Bit No.	Name	Reset	R/W	Description
7:6	FOC_MODE	01	R/W	Frequency-offset average filter behavior 00: No frequency-offset compensation done 01: Freeze frequency-offset estimate after sync 10: Continuously estimate and remove frequency offset 11: Freeze the frequency-offset estimate after sync, use double decay rate
5	-	0	R0	Reserved
4:0	CORR_THR[4:0]	0x0F	R/W	Demodulator correlator threshold value, used in sync search. Optimal threshold value depends on SW_CONF . SW_LEN. CORR_THR adjusts how the receiver synchronizes to data from the radio. If threshold is set too low, sync can more easily be found on noise. If set too high, the sensitivity is reduced but sync is not likely to be found on noise.

Bit No.	CTRL2 (0x6192) – Modem Config Name	Reset	R/W	Description
7	SW_BIT_ORDER	0	R/W	<ul> <li>0: The sync word is transmitted LSB to MSB (from SYNC_WORD[0] to SYNC_WORD[31]), and in receive the correlator expects this bit ordering.</li> <li>1: The sync word is transmitted MSB to LSB (from SYNC_WORD[31]) to SYNC_WORD[0]), and in receive the correlator expects this bit ordering.</li> </ul>
6	DEM_PREAM_MODE	0	R/W	Use PREAM_SEL[1:0] or 1s complement of PREAM_SEL[1:0] for frequency offset estimation.
5:4	PREAM_SEL[1:0]	00	R/W	<ul> <li>00: Select preamble based on first bit of sync word; last bit of preamble is inverse of first bit of sync word.</li> <li>01: Select preamble based on first bit of sync word; last bit of preamble is same as first bit of sync word.</li> <li>10: Use preamble 0101 0101</li> <li>11: Use preamble 1010 1010</li> </ul>
3:0	NUM_PREAM_BYTES[3:0]	0000	R/W	The number of preamble bytes to be sent in Tx mode prior to the sync word 0000: 1 leading preamble byte 0001: 2 leading preamble bytes 0010: 3 leading preamble bytes 0011: 4 leading preamble bytes
				 1111: 16 leading preamble bytes

#### MDMCTRL3 (0x6193) – Modem Configuration

Bit No.	Name	Reset	R/W	Description
7:6	SYNC_MODE[1:0]	01	R/W	<ul> <li>00: Correlation above threshold is sufficient as sync criterion.</li> <li>01: Correlation value above threshold and data decision on all symbols of sync word is used as sync criterion.</li> <li>10: Correlation value above threshold and data decision on all symbols of sync word is used as sync criterion. Accept one bit error in sync word 11: Reserved</li> </ul>
5	RAMP_AMP	1	R/W	<ol> <li>Enable ramping of DAC output amplitude during startup and finish.</li> <li>Disable ramping of DAC output amplitude.</li> </ol>
4:3	RFC_SNIFF_CTRL[1:0]	00	R/W	Enable / disable <i>rfc_sniff.</i> 00: Sniffer disabled 01: Output data out of the BSP 10: Output data out of the modulator before the BSP 11: Output data out of the demodulator before the BSP
2	-	0	R0	Reserved. Read as 0.
1:0	RSSI_MODE[1:0]	00	R/W	Controls mode of RSSI 00 : Continuous mode 01 : Freeze estimate at sync 10 : Peak detect 11 : Continuous before sync, peak detect after sync

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#### Registers

SW_CONF (0x6194) -	Sync Word	Configuration
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Bit No.	Name	Reset	R/W	Description
7	DUAL_RX	0	R/W	0: Only search for primary SW 1: Search for both primary and secondary SW
6	-	0	R/W	Reserved. Always write 0.
5	SW_RX	0	R	0: Primary SW received 1: Secondary SW received Valid only when RFSTAT . SFD is 1
4:0	SW_LEN[4:0]	0 0000	R/W	Determines how many of the bits in SYNC_WORD are to be used. This allows for arbitrary sync word lengths. 0 0000: 32-bit SW 0 0001 to 0 11111: Reserved 1 0000: 16-bit SW 1 0001: 17-bit SW 1 0010: 18-bit SW 1 0011: 19-bit SW  

#### SW0 (0x6195) – Primary Sync Word Byte 0

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD[7:0]	0x00	R/W	Contains bits 7:0 of the primary synchronization word

#### SW1 (0x6196) - Primary Sync Word Byte 1

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD[15:8]	0x00	R/W	Contains bits 15:8 of the primary synchronization word

#### SW2 (0x6197) – Primary Sync Word Byte 2

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD[23:16]	0x00	R/W	Contains bits 23:16 of the primary synchronization word

#### SW3 (0x6198) - Primary Sync Word Byte 3

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD[31:24]	0x00	R/W	Contains bits 31:24 of the primary synchronization word

#### SW4 (0x61F8) – Secondary Sync Word Byte 0

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD2[7:0]	0x00	R/W	Contains bits 7:0 of the secondary synchronization word

# SW5 (0x61F9) – Secondary Sync Word Byte 1

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD2[15:8]	0x00	R/W	Contains bits 15:8 of the secondary synchronization word

#### SW6 (0x61FA) – Secondary Sync Word Byte 2

Bit No.	Name	Reset	R/W	Description
7:0	SYNC_WORD2[23:16]	0x00	R/W	Contains bits 23:16 of the secondary synchronization word

SW7 (0x61FB) – Secondary Sync Word Byte 3					
Bit No.	Name	Reset	R/W	Description	
7:0	SYNC_WORD2[31:24]	0x00	R/W	Contains bits 31:24 of the secondary synchronization word	

# FREQEST (0x6199) – Estimated RF Frequency Offset

Bit No.	Name	Reset	R/W	Description
7:0	FREQEST[7:0]	0x00	R	Signed value. Contains an estimate of the frequency offset between carrier and the receiver frequency. FOC_MODE controls when this estimate is updated.

### RXCTRL (0x619A) - Receive Section Tuning

Bit No.	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5:4	GBIAS_LNA2_REF[1:0]	10	R/W	Adjusts front-end LNA2/mixer PTAT current output (M = GBIAS_LNA2_REF[1:0] + 3), default: M = 5.
3:2	GBIAS_LNA_REF[1:0]	10	R/W	Adjusts front-end LNA PTAT current output (M = GBIAS_LNA_REF[1:0] + 3), default: M = 5.
1:0	MIX_CURRENT[1:0]	01	R/W	Control of the receiver mixers output current. The current increases with increasing setting.

#### FSCTRL (0x619B) – Frequency Synthesizer Tuning

Bit No.	Name	Reset	R/W	Description
7:6	PRE_CURRENT [1:0]	01	R/W	Prescaler current setting
5:4	LODIV_BUF_CURRENT_TX [1:0]	01	R/W	Adjusts current in mixer and PA buffers (lodiv_buf_current). Used when lle_tx_active = 1
3:2	LODIV_BUF_CURRENT_RX [1:0]	01	R/W	Adjusts current in mixer and PA buffers (lodiv_buf_current). Used when lle_tx_active = 0
1:0	LODIV_CURRENT [1:0]	01	R/W	Adjusts divider currents, except mixer and PA buffers.

#### LNAGAIN (0x61A0) - LNA Gain Setting

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved, read as 0
6:5	LNA1_CURRENT[1:0]	11	R/W	Gain setting LNA1 00: 0-dB gain (reference level) 01: 3-dB gain 10: Reserved 11: 6-dB gain
4:2	LNA2_CURRENT[2:0]	111	R/W	Gain setting LNA2 000: 0-dB gain (reference level) 001: 3-dB gain 010: 6-dB gain 011: 9-dB gain 100: 12-dB gain 101: 15-dB gain 110: 18-dB gain 111: 21-dB gain
1:0	LNA3_CURRENT[1:0]	11	R/W	Gain setting LNA3 00: 0-dB gain (reference level) 01: 3-dB gain 10: 6-dB gain 11: 9-dB gain



#### Registers

# AAFGAIN (0x61A1) – AAF Gain Setting

-							
Bit No.	Name	Reset	R/W	Description			
7:2	-	0000 00	R0	Reserved. Read as zero			
1:0	AAF_GAIN[1:0]	11	R/W	Controls attenuation in AAF 00: 9-dB attenuation in AAF 01: 6-dB attenuation in AAF 10: 3-dB attenuation in AAF 11: 0-dB attenuation in AAF (reference level)			

# ADCTEST0 (0x61A2) – ADC Tuning

Bit No.	Name	Reset	R/W	Description
7:0	ADC_ADJ[7:0]	0x10	R/W	Adjust ADC gain

#### MDMTEST0 (0x61A5) – Modem Configuration

Bit No.	Name	Reset	R/W	Description
7:5	RSSI_ACC[2:0]	101	R/W	RSSI accuracy 000: 5.33-µs average window 001: Mean of two 5.33-µs average windows 010: Reserved 011: Mean of four 5.33-µs average windows 100: 21.3-µs average window 101: Mean of two 21.3-µs average windows 110: Reserved 111: Mean of four 21.3-µs average windows
4	-	0	R/W	Reserved, always write 0.
3:2	DC_BLOCK_LENGTH[1:0]	00	R/W	Controls the number of samples to be accumulated between each dump of the accumulate-and-dump filter used in dc removal. 00: 16 samples 01: 32 samples 10: 64 samples 11: 128 samples
1:0	DC_BLOCK_MODE[1:0]	01	R/W	Selects the mode of operation: 00 : Manual override mode 01 : Enable dc cancellation. Normal operation 10 : Freeze estimates of dc when sync is found. Start estimating dc again when searching for the next frame. 11 : Delayed dc offset estimate used. Delay set by MDMTEST1.DC_DELAY. Until the first estimate is ready, the manual override value is used.



Bit No.	Name	Reset	R/W	Description
7:6	DC_DELAY	00	R/W	Controls delay of dc estimate delayed dc block mode. Delay unit is set by MDMTEST0.DC_BLOCK_LENGTH 00: 5 delays 01: 6 delays 10: 7 delays 11: 8 delays
5	RX_IF	0	R/W	Controls mixer frequency in demodulator (not 2 Mbps) 0: 1 MHz 1: -1 MHz For 2 Mbps, always write 0. The receiver then operates at zero IF.
4:0	TX_TONE [ 4 : 0 ]	0 0000	R/W	Controls baseband frequency of transmission Note: If MDMCTRL0.PHASE_INVERT is 1, the sign of the frequency is inverted 0: -8 MHz 1: -6 MHz 2: -4 MHz 3: -3 MHz 4: -2 MHz 5: -1 MHz 6: -500 kHz 7: -250 kHz 8: -125 kHz 10: 0 Hz 11: 4 kHz 12: 125 kHz 13: 250 kHz 14: 500 kHz 15: 1 MHz 16: 2 MHz 17: 3 MHz 18: 4 MHz 19: 6 MHz 20: 8 MHz

# ATEST (0x61A9) – Analog Test Control

Bit No.	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved. Read as zero
5:0	ATEST_CTRL[5:0]	00 0000	R/W	Controls the analog test mode: 00 0000: Disabled 00 0001: Enables the temperature sensor (see also the TR0 register description in Chapter 12).Other values reserved.

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#### Registers

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as 0
6	RFC_OBS_POL0	0	R/W	The signal chosen by RFC_OBS_MUX0 is XORed with this bit
5:0	RFC_OBS_MUX0	00 0000		Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(0); see Section 7.9. 00 0111: rfc_sniff_data – Data from packet sniffer, see Section 25.11 00 1000: rfc_sniff_clk – Clock for packet sniffer data, see Section 25.11 00 1001: tx_active 00 1010: rx_active 00 1011: vco_on – VCO on Low: The VCO is powered down, so the next Rx or Tx operation must start and calibrate the synthesizer before transmitting or receiving High: The VCO is powered up. If the LLE is idle, it means the next task starts quickly if frequency programming is disabled (PRF_CHAN . FREQ = 127) 00 1100: sync_search – Rx search for sync Low: The modem is not ready to receive a packet High: The modem is in search for a sync word or receiving a packet 00 1101: lle_idle – Link-layer engine idle Low: The LLE is busy processing or finishing a command, or in reset High: The LLE is idle waiting for a command to start a new task 00 1110: wait_t2e1 – Indication on the LLE waiting for Timer 2 event 1 to start a task Low: Not waiting for Timer 2 event 1 High: Command processed, event 1 not yet received 00 1111: agc_lowgain – High if the AGC algorithm has reduced the front-end gain; low otherwise 01 0011: fsc_lock – High when PLL is in lock; low otherwise 01 1011: na_pd - Power amplifier power-down signal 10 1000: dem_sync_found - High when demodulator has detected a sync word. Stays high until end of packet. 11 0001: mod_sync_sent - High when modulator has sent a sync word. Stays high until end of packet.

# RFC_OBS_CTRL1 (0x61AF) – RF Observation Mux Control 1

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reseved. Read as 0
6	RFC_OBS_POL1	0	R/W	The signal chosen by RFC_OBS_MUX1 is XORed with this bit.
5:0	RFC_OBS_MUX1	00 0000	R/W	Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(1). See description of RFC_OBS_CTRL0.

#### RFC_OBS_CTRL2 (0x61B0) - RF Observation Mux Control 2

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reseved. Read as 0
6	RFC_OBS_POL2	0	R/W	The signal chosen by RFC_OBS_MUX2 is XORed with this bit.
5:0	RFC_OBS_MUX2	00 0000	R/W	Controls which observable signal from rf_core is to be muxed out to rfc_obs_sigs(2). See description of RFC_OBS_CTRL0.

# LLECTRL (0x61B1) – LLE Control

LLEC	IRL (0X61B1) – LLE CONT	01		
Bit No.	Name	Reset	R/W	Description
7:3	-	0	R0	Reseved. Read as 0
2:1	LLE_MODE_SEL	00	R/W	LLE mode. Changing this field has no effect unless LLE_EN is changed from 0 to 1. The value is reset to 00 when coming up from PM2 or PM3. 00: Proprietary mode (described in this chapter) 01: BLE mode (only for use by the BLE stack) Others: Reserved
0	LLE_EN	0	R/W	0: LLE held in reset 1: LLE enabled

#### TXFILTCFG (0x61BC) – TX Filter Configuration

Bit No.	Name	Reset	R/W	Description
7:4	-	0000	R0	Reserved
3:2	-	11	R/W	Reserved
1:0	FC	11	R/W	Sets TX anti-aliasing filter to appropriate bandwidth. Reduces spurious emissions close to signal. For the best value to use, see Table 25-24 and Table 25-25.

#### RFRND (0x61BF) - Random Data

Bit No.	Name	Reset	R/W	Description
7:0	RND	0x00	R	Random bits, provided analog part is in random number generation mode (receiver running without sync)

#### RFRAMCFG (0x61C0) – Radio RAM Configuration

Bit No.	Name	Reset	R/W	Description
7:3	-	0000 1	R	Reserved
2:0	PRE	000	R/W	Selects active memory page for RF core data memory

TEXAS INSTRUMENTS

#### Registers

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Bit No.	Name	Reset	R/W	Description
7:5	-	000	R	Reserved
4:0	DMAO	0x00	R/W	Generate a pulse on radio DMA trigger 0 (DMA trigger 19) when: 0x00: Never 0x01: A byte is read from Rx FIFO and more bytes remain or when a byte arrives in Rx FIFO and it was previously empty. 0x02: A byte is written to Rx FIFO and there is available space left or when there becomes available space when the Rx FIFO was full. 0x03: Rx FIFO is empty. 0x04: Rx FIFO is full. 0x05: Rx FIFO length equals RFRXFTHRS after a write to Rx FIFO. 0x06: Rx FIFO is read when its size equals RFRXFTHRS . 0x07: Rx FIFO is reset (see Table 25-2). 0x08: Rx FIFO is deallocated (see Table 25-2). 0x08: Rx FIFO is deallocated (see Table 25-2). 0x08: Rx FIFO is discarded (see Table 25-2). 0x08: Rx FIFO is discarded (see Table 25-2). 0x08: Rx FIFO is discarded (see Table 25-2). 0x08: Rx FIFO is committed (see Table 25-2). 0x08: Rx FIFO is discarded (never) 0x10: Never 0x11: A byte is read from Tx FIFO and more bytes remain or when a byte arrives in Tx FIFO and it was previously empty. 0x12: A byte is written to Tx FIFO and there is available space left or when there becomes available space when the Tx FIFO was full. 0x13: Tx FIFO is full. 0x15: Tx FIFO is full. 0x15: Tx FIFO is full. 0x15: Tx FIFO is read when its size equals RFTXFTHRS . 0x17: Tx FIFO is full. 0x16: Tx FIFO is read when its size equals RFTXFTHRS . 0x17: Tx FIFO is reset (see Table 25-2). 0x18: Tx FIFO is reset (see Table 25-2). 0x18: Tx FIFO is reset (see Table 25-2). 0x18: Tx FIFO is deallocated (see Table 25-2). 0x18: Tx FIFO is committed (see Table 25-2). 0x18: Tx FIFO is c

RFFDMA1, (0x61C4) – Radio DMA Trigger 1 Control

Bit No.	Name	Reset	R/W	Description
7:5	-	0x0	R	Reserved
4:0	DMA1	0x0	R/W	Condition for generating a pulse on radio DMA trigger 1 (DMA trigger 11). See RFFDMA0 for the list of conditions.

RFFSTATUS (0x61C5) – FIFO Status

Bit No.	Name	Reset	R/W	Description
7	TXAVAIL	0	R	0: No readable data in Tx FIFO 1: Readable data present in Tx FIFO
6	TXFEMPTY	1	R	0: Data present in Tx FIFO 1: Tx FIFO is empty
5	TXDTHEX	1	R	0: There is less data in Tx FIFO than the threshold amount given by RFTXFTHRS.1: There is more than or equal amount of data in Tx FIFO than the threshold amount given by RFTXFTHRS
4	TXFFULL	0	R	0: Tx FIFO has available space 1: Tx FIFO is full
3	RXAVAIL	0	R	0: No readable data in Rx FIFO 1: Readable data present in Rx FIFO
2	RXFEMPTY	1	R	0: Data present in Rx FIFO 1: Rx FIFO is empty
1	RXDTHEX	1	R	0: There is less data in Rx FIFO than the threshold amount given by RFRXFTHRS.1: There is more than or equal amount of data in Rx FIFO than the threshold amount given by RFRXFTHRS
0	RXFFULL	0	R	0: Rx FIFO has available space 1: Rx FIFO is full

RFFCFG (0x61C6) – FIFO Configuration

Bit No.	Name	Reset	R/W	Description
7:6	-	0	R	Reserved
5	TXAUTOCOMMIT	1	R/W	0: Commit Tx FIFO only on command 0x95 1: Always set RFTXSWP = RFTXWP
4	TXFAUTODEALLOC	0	R/W	0: Deallocate Tx FIFO only on command 0x92 1: Always set RFTXFSRP = RFTXFRP.
3:2	-	0	R	Reserved
1	RXAUTOCOMMIT	0	R/W	0: Commit Rx FIFO only on command 0x85 1: Always set RFRXSWP = RFRXWP
0	RXFAUTODEALLOC	1	R/W	0: Deallocate Rx FIFO only on command 0x82 1: Always set RFRXFSRP = RFRXFRP.

RFRXFLEN (0x61C8) - Rx FIFO Length

Bit No.	Name	Reset	R/W	Description
7:0	D	0x00	R	Amount of data present in Rx FIFO

RFRXFTHRS (0x61C9) – Rx FIFO Threshold

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Threshold value for Rx FIFO

RFRXFWR (0x61CA) - Rx FIFO Write Register

Bit No.	Name	Reset	R/W	Description
7:0	D	0x00	W	Data written to this register is written to the Rx FIFO address at offset RFRXFWP from the start of the Rx FIFO area (see Figure 25-1). RFRXFWP (and RFRXFSWP if RFFCFG.RXAUTODEALLOC = 1) is incremented by 1 modulo 0x80 unless the write fails.

Registers



Registers

RFRXF	RFRXFRD (0x61CB) – Rx FIFO Read Register					
Bit No.	Name	Reset	R/W	Description		
7:0	D	0x00	R	When this register is read, the data in Rx FIFO address offset RFRXFRP from the start of the Rx FIFO area is returned (see Figure 25-1). RFRXFRP (and RFRXFSRP if RFFCFG.RXAUTOCOMMIT = 1) is incremented by 1 modulo 0x80 unless the read fails.		

RFRXFWP (0x61CC) - Rx FIFO Write Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Rx FIFO write pointer. This is the offset into Rx FIFO the next write operation writes to.

RFRXFRP (0x61CD) - Rx FIFO Read Pointer

Bit No.	Name	Reset	R/W	Description
7	_	0	R	Reserved
6:0	D	000 0000	R/W	Rx FIFO read pointer. This is the offset into Rx FIFO the next read operation reads from.

RFRXFSWP (0x61CE) - Rx FIFO Start-of-Frame Write Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Rx FIFO start of written package. This is the point to which the write pointer can be reset if a discard command is issued.

RFRXFSRP (0x61CF) - Rx FIFO Start-of-Frame Read Pointer

Bit No.	Name	Reset	R/W	Description
7	_	0	R	Reserved
6:0	D	000 0000	R/W	Rx FIFO start of read package. This is the start of the allocated part of the Rx FIFO.

RFTXFLEN (0x61D0) – Tx FIFO Length

Bit No.	Name	Reset	R/W	Description
7:0	D	0x00	R	Amount of data present in Tx FIFO

RFTXFTHRS (0x61D1) – Tx FIFO Threshold

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Threshold value for Tx FIFO

RFTXFWR (0x61D2) - Tx FIFO Write Register

Bit No.	Name	Reset	R/W	Description
7:0	D	0x00	W	Data written to this register is written to the Tx FIFO address at offset RFTXFWP from the start of the Tx FIFO area (see Figure 25-1) is returned. RFTXFWP (and RFTXFSWP if RFFCFG.TXAUTODEALLOC = 1) is incremented by 1 modulo 0x80 unless the write fails.



RFTXF	RFTXFRD (0x61D3) – Tx FIFO Read Register					
Bit No.	Name	Reset	R/W	Description		
7:0	D	0x00	R	When this register is read, the data in Tx FIFO address offset RFTXFRP from the start of the Tx FIFO area is returned (see Figure 25-1). RFTXFRP (and RFTXFSRP if RFFCFG.TXAUTOCOMMIT = 1) is incremented by 1 modulo 0x80 unless the read fails.		

RFTXFWP (0x61D4) - Tx FIFO Write Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Tx FIFO write pointer. This is the offset into Tx FIFO the next write operation writes to.

RFTXFRP (0x61D5) – Tx FIFO Read Pointer

Bit No.	Name	Reset	R/W	Description
7	_	0	R	Reserved
6:0	D	000 0000	R/W	Tx FIFO read pointer. This is the offset into Tx FIFO the next read operation reads from.

RFTXFSWP (0x61D6) - Tx FIFO Start-of-Frame Write Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	000 0000	R/W	Tx FIFO start of written package. This is the point to which the write pointer can be reset if a discard command is issued.

RFTXFSRP (0x61D7) – Tx FIFO Start-of-Frame Read Pointer

Bit No.	Name	Reset	R/W	Description
7	-	0	R	Reserved
6:0	D	0x00	R/W	Tx FIFO start-of-read package. This is the start of the allocated part of the Tx FIFO.

BSP_P0 (0x61E0) - CRC Polynomial Byte 0

Bit No.	Name	Reset	R/W	Description
7:0	P[7:0]	0x00	R/W	Bits 7:0 of p register in CRC sub-module

BSP_P1 (0x61E1) - CRC Polynomial Byte 1

Bit No.	Name	Reset	R/W	Description
7:0	P[15:8]	0x5B	R/W	Bits 15:8 of p register in CRC sub-module

BSP_P2 (0x61E2) – CRC Polynomial Byte 2

Bit No.	Name	Reset	R/W	Description
7:0	P[23:16]	0x06	R/W	Bits 23:16 of p register in CRC sub-module

BSP_P3 (0x61E3) – CRC Polynomial Byte 3

Bit No.	Name	Reset	R/W	Description
7:0	P[31:24]	0x00	R/W	Bits 31:24 of p register in CRC sub-module

Registers

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BSP_D	BSP_D0 (0x61E4) – CRC Value Byte 0						
Bit No.	Name	Reset	R/W	Description			
7:0	D[7:0]	0x00	R/W	Bits 7:0 of d register in CRC sub-module			

BSP_D1 (0x61E5) - CRC Value Byte 1

Bit No.	Name	Reset	R/W	Description
7:0	D[15:8]	0x5B	R/W	Bits 15:8 of d register in CRC sub-module

BSP_D2 (0x61E6) – CRC Value Byte 2

Bit No.	Name	Reset	R/W	Description
7:0	D[23:16]	0x06	R/W	Bits 23:16 of d register in CRC sub-module

BSP_D3 (0x61E7) – CRC Value Byte 3

Bit No.	Name	Reset	R/W	Description
7:0	D[31:24]	0x00	R/W	Bits 31:24 of d register in CRC sub-module

BSP_W (0x61E8) – Whitener Value

Bit No.	Name	Reset	R/W	Description
7	W_PN9_RESET	0	R0	When a 1 is written to this bit, the CC2500-compatible whitener is reset, and all bits in the s and b registers are set to 1.
6:0	W[6:0]	110 0101	R/W	Write: Writes all whitening registers. w_6 is set to BSP_W[0], w_5 is set to BSP_W[1] and so on up to w_1 is set to BSP_W[5]. w_0 is set to 1. Read: Reads back w register. BSP_W[0] is set to w_6 , BSP_W[1] is set to w_5 and so on up to BSP_W[6] is set to w_0 .

BSP_MODE (0x61E9) – Bit Stream Processor Configuration

Bit No.	Name	Reset	R/W	Description
7	-	0	R0	Reserved. Read as zero
6	CP_BUSY	0	R	Coprocessor mode busy. Goes to 1 after a byte has been written to BSP_DATA. Goes to 0 when a byte is ready to be read back from BSP_DATA
5	CP_READOUT	0	R/W	Coprocessor mode readout
4	CP_END	0	R/W	Endianness of data in coprosessor mode. 0: LSB processed first 1: MSB processed first
3:2	CP_MODE[1:0]	00	R/W	Coprocessor mode 00: Coprocessor disabled 01: Coprocessor receive mode 10: Reserved 11: Coprocessor transmit mode
1	W_PN9_EN	0	R/W	Enable CC2500-compatible PN9 whitener
0	W_PN7_EN	1	R/W	Enable PN7 whitener

BSP_DATA (0x61EA) – Bit Stream Processor Coprocessor Data

Bit No.	Name	Reset	R/W	Description
7:0	BSP_DATA[7:0]	0x00	R/W	When BSP_MODE.CP_BUSY = 0: Write: Provide byte to be processed in coprocessor mode Read: Read processed byte



DC_I_	DC_I_L (0x61FC) – In-Phase DC Offset Estimate, Low Byte						
Bit No.	Name Reset R/W Description						
7:0	DC_I[7:0]	0x00	R*/W	When running dc estimation, this register reflects the 8 LSBs of the dc estimate in the I channel. When manual dc override is selected, the override value is written to this register.			

DC_I_H (0x61FD) – In-Phase DC Offset Estimate, High Byte

Bit No.	Name	Reset	R/W	Description	
7:0	DC_I[15:8]	estimate		When running dc estimation, this register reflects the 8 MSBs of the dc estimate in the I channel. When manual dc override is selected, the override value is written to this register.	

DC_Q_L (0x61FE) – Quadrature-Phase DC Offset Estimate Low Byte

Bit No.	Name	Reset	R/W	Description	
7:0	DC_Q[7:0]	0x00	R*/W	When running dc estimation. this register reflects the 8 LSBs of the dc estimate in the Q channel. When manual dc override is selected, the override value is written to this register.	

DC_Q_H (0x61FF) – Quadrature-Phase DC Offset Estimate High Byte

Bit No.	Name	Reset	R/W	Description	
7:0	DC_Q[15:8]	0x00	R*/W	When running dc estimation, this register reflects the 8 MSBs of the dc estimate in the Q channel. When manual dc override is selected, the override value is written to this register.	

IVCTRL (0x6265) – Analog Control Register

Bit No.	Name	Reset	R/W	Description
7:6	-	00	R0	Reserved
5:4	TX_MIX_LOAD	01	R/W	Controls load capacitor in Tx mixer 00: Minimum load (Intermediate loads) 11: Maximum load
3	LODIV_BIAS_CTRL	0	R/W	Controls bias current to LODIV 0: IVREF bias 1: PTAT bias
2	TXMIX_DC_CTRL	0	R/W	Controls dc bias in TXMIX
1:0	PA_BIAS_CTRL	11	R/W	Controls bias current to PA 00: IREF bias 01: IREF and IVREF bias 10: PTAT bias 11: Increased PTAT slope bias



Voltage Regulator

The digital voltage regulator is used to power the digital core. The output of this regulator is available on the DCOUPL pin and requires capacitive decoupling to function properly (see, e.g., the CC2530 reference design).

The voltage regulator is disabled in power modes PM2 and PM3 (see Chapter 4). When the voltage regulator is disabled, register and RAM contents are retained while the unregulated 2 V to 3.6 V power supply is present

NOTE: The voltage regulator should not be used to provide power to external circuits.



Available Software

This chapter presents the various available software solutions relevant to the CC253x/CC2540/CC2541 family. They are all available free of charge on the TI Web site at www.ti.com/lprf when used with TI LPRF devices.

As shown in Table 0-1 in the Preface, the members of the CC253x/CC2540/CC2541 family have different flash/RAM sizes; hence, they are not equally well suited for the different software offerings in the sections below. For example, a user designing a ZigBee device should use the CC2530F256 as the Z-Stack[™] requires in most cases more than 128 KB of flash and needs the 8-KB RAM.

Topic

27.1	SmartRF [™] Software for Evaluation (www.ti.com/smartrfstudio)	358
27.2	RemoTI™ Network Protocol (www.ti.com/remoti)	358
27.3	SimpliciTI [™] Network Protocol (www.ti.com/simpliciti)	359
27.4	TIMAC Software (www.ti.com/timac)	359
27.5	Z-Stack™ Software (www.ti.com/z-stack)	360
27.6	BLE Stack Software	360



27.1 SmartRF[™] Software for Evaluation (www.ti.com/smartrfstudio)

Texas Instruments' SmartRF Studio can be used for radio performance and functionality evaluation and is great for exploring and gaining knowledge about the RF-IC products. This software helps the designers of radio systems to evaluate the RF-ICs easily at an early stage in the design process. It is especially useful for generation of the configuration data and for finding optimized external component values.

SmartRF Studio software runs on Microsoft[™] Windows[™] 98, Windows 2000, Windows XP, Windows Vista (32 bit) and Windows 7 (32 bit). SmartRF Studio software can be downloaded from the Texas Instruments Web page: www.ti.com/smartrfstudio (http://www.ti.com/litv/zip/swrc046m).

Features

- Link tests. Send and receive packets between nodes
- Packet error-rate (PER) tests
- Communication with evaluation boards through the USB port or the parallel port
- Up to eight USB devices are supported on a single computer.
- · Normal view with preferred register settings
- Register view with possibilities to read and write each individual register. Each register given with detailed information
- Save/Open configuration data from file
- · Save/Load register settings from file
- Export/Import register values from text file
- Exports register settings into a C-compatible software structure

27.2 RemoTI[™] Network Protocol (www.ti.com/remoti)

Most existing remote controls use infrared technology to communicate commands to consumer electronics devices. However, radio frequency (RF) remote controls enable non-line-of-sight operation and provide more advanced features based on bidirectional RF communication.

ZigBee Radio Frequency for Consumer Electronics (RF4CE) is the result of a recent agreement between the ZigBee Alliance and the RF4CE Consortium (<u>http://www.zigbee.org/rf4ce</u>) and has been designed to be deployed in a wide range of remotely-controlled audio/visual consumer electronics products, such as TVs and set-top boxes. ZigBee RF4CE promises you:

- Richer communication and increased reliability
- Enhanced features and flexibility
- Interoperability
- No line-of-sight barrier

The RemoTI network protocol is Texas Instruments' implementation of the ZigBee RF4CE standard. It is a complete solution offering hardware and software support for TI's low-power RF product portfolio. With the RemoTI network protocol we provide:

- The industry-leading RF4CE-compliant stack featuring the interoperable CERC profile support, a simple API, easy-to-understand sample application code, full development kits and reference designs, and much more.
- Operation on our best-in-class IEEE 802.15.4 compliant System-on-Chip, the CC2530, with excellent RF co-existence and RF performance. The four flexible power modes include the lowest-current-consumption power-down mode for long battery in life low-duty-cycle applications.
- Extensive worldwide support and tools to ensure that development of ZigBee RF4CE-based products is simple, fast, and can be completed at minimal cost.
- The RemoTI network protocol is a Golden Unit platform; i.e., it is used for testing other implementations of the ZigBee RF4CE standard for standard compliance.

For more information on TI's RemoTI network protocol, see the Texas Instruments RemoTI network protocol Web site <u>www.ti.com/remoti</u>.



27.3 SimpliciTI[™] Network Protocol (www.ti.com/simpliciti)

The SimpliciTI network protocol is a low-power RF protocol (for sub-1 GHz, 2.4 GHz, and IEEE 802.15.4 RF ICs) targeting simple, small RF networks. This open-source software is an excellent start for building a network with battery-operated devices using a TI low-power RF System-on-Chip (SoC). The SimpliciTI network protocol was designed for easy implementation and deployment out-of-the-box on several TI RF platforms. It provides several sample applications.

Key Applications

- Alarm and security: occupancy sensors, light sensors, carbon monoxide sensors, glass-breakage detectors
- Smoke detectors
- Automatic meter reading: gas meters, water meters, e-meters
- Active RFID applications

Key Features

- Low power: A TI-proprietary low-power network protocol
- Flexible:
 - Direct device-to-device communication
 - Simple star with access point for store and forward to end device
 - Range extenders to increase range to four hops
- Simple: uses a five-command API
- Low data rate and low duty cycle
- Ease of use

For more information about the SimpliciTI network protocol, see the Texas Instruments SimpliciTI network protocol Web site <u>www.ti.com/simpliciti</u>.

27.4 TIMAC Software (www.ti.com/timac)

TIMAC software is an IEEE 802.15.4 medium-access-control software stack for TI's IEEE 802.15.4 transceivers and System-on-Chips.

You can use TIMAC when you:

- Need a wireless point-to-point or point-to-multipoint solution; e.g., multiple sensors reporting directly to a master
- Need a standardized wireless protocol
- Have battery-powered and/or mains-powered nodes
- · Need support for acknowledgement and retransmission
- Have low data-rate requirements (around 100-kbps effective data rate)

Features

- Support for IEEE 802.15.4 standard
- Support for beacon-enabled and non-beaconing systems
- Multiple platforms
- Easy application development

The TIMAC software stack is certified to be compliant with the IEEE 802.15.4 standard. TIMAC software is distributed as object code free of charge. There are no royalties for using TIMAC software.

For more information about TIMAC software, see the Texas Instruments TIMAC Web site <u>www.ti.com/timac</u>.



27.5 Z-Stack[™] Software (www.ti.com/z-stack)

The Z-Stack software is TI's ZigBee-compliant protocol stack for a growing portfolio of IEEE 802.15.4 products and platforms. The Z-Stack software stack is compliant with the ZigBee-2007 specification, supporting both the ZigBee and ZigBee PRO features sets. The Z-Stack software includes implementation of two ZigBee application profiles – SmartEnergy and HomeAutomation. Other application profiles can easily be implemented by the user.

Z-Stack software notables include:

- A fully compliant ZigBee and ZigBee PRO feature set
- A range of sample applications including support for the ZigBee Smart Energy and ZigBee Home Automation profiles
- Over-the-air download and serial boot loader support
- Can be used together with the RF front ends, <u>CC2590</u> and <u>CC2591</u>, which support 10 dBm and 20 dBm output power, respectively, and improved receive sensitivity.

The Z-Stack software has been awarded the ZigBee Alliance's golden-unit status for both the ZigBee and ZigBee PRO stack profiles and is used by ZigBee developers worldwide.

Z-Stack software is well suited for:

- Smart energy (AMI)
- Home automation
- Commercial building automation
- Medical, assisted living, or personal health and hospital care
- Monitoring and control applications
- · Wireless sensor networks
- Alarm and security
- Asset tracking
- Applications that require interoperability

For more information about Z-Stack software, see the Texas Instruments Z-Stack software Web site <u>www.ti.com/z-stack</u>.

27.6 BLE Stack Software

TI's single-mode *Bluetooth* low energy stack has been certified according to the *Bluetooth* 4.0 low energy specification. Key features:

- Supports all BLE roles
- Range of example applications
- Multi-role capabilities

For more information about TI's BLE stack software, visit Texas Instruments *Bluetooth* low energy stack software Web site at <u>www.ti.com/blestack</u>.



Abbreviations

Abbreviations used in this user's guide:

AAF	Anti-aliasing filter
ACK	Acknowledge
ADC	Analog-to-digital converter
AES	Advanced Encryption Standard
AGC	Automatic gain control
AGC	Association of Radio Industries and Businesses
BCD	Binary-coded decimal
BER	Bit error rate
BLE	Bluetooth low-energy
BOD	Brownout detector
BOM BSP	Bill of materials
	Bit-stream process
CBC CBC-MAC	Cipher block chaining
	Cipher block chaining message authentication code
CCA	Clear channel assessment
CCM	Counter mode + CBC-MAC
CFB	Cipher feedback
CFR	Code of Federal Regulations
CMRR	Common-mode rejection ratio
CPU	Central processing unit
CRC	Cyclic redundancy check
CSMA-CA	Carrier sense multiple access with collision avoidance
CSP	CSMA/CA strobe processor
CTR	Counter mode (encryption)
CW	Continuous wave
DAC	Digital-to-analog converter
DC	Direct current
DMA	Direct memory access
DSM	Delta-sigma modulator
DSSS	Direct-sequence spread spectrum
ECB	Electronic code book (encryption)
EM	Evaluation module
ENOB	Effective number of bits
ETSI	European Telecommunications Standards Institute
EVM	Error vector magnitude
FCC	Federal Communications Commission
FCF	Frame control field
FCS	Frame check sequence
FFCTRL	FIFO and frame control



Appendix A

FIFO	First in, first out
FS	Full scale
GPIO	General-purpose input/output
HF	High frequency
HSSD	High-speed serial data
I/O	Input/output
I/Q	In-phase/quadrature-phase
IEEE	Institute of Electrical and Electronics Engineers
IF	Intermediate frequency
IOC	I/O controller
IRQ	Interrupt request
IR	Infrared
ISM	Industrial, scientific and medical
ITU-T	International Telecommunication Union – Telecommunication
IV	Initialization vector
KB	1024 bytes
kbps	Kilobits per second
LFSR	Linear feedback shift register
LLE	Link-layer engine
LNA	Low-noise amplifier
LO	Local oscillator
LQI	Link quality indication
LSB	Least-significant bit/byte
MAC	Medium access control
MAC	Message authentication code
MCU	Microcontroller unit
MFR	MAC footer
MHR	MAC header
MIC	Message integrity code
MISO	Master in, slave out
MOSI	Master out, slave in
MPDU	MAC protocol data unit
MSB	Most-significant bit/byte
MSDU	MAC service data unit
MUX	Multiplexer
NA	Not applicable/available
NC	Not connected
OFB	Output feedback (encryption)
O-QPSK	Offset – quadrature phase-shift keying
PA	Power amplifier
PC	Program counter
PCB	Printed circuit board
PER	Packet error rate
PHR	PHY header
PHY	Physical layer
PLL	Phase-locked loop
PM1, PM2,	Power mode 1, 2, and 3
PM3	
PMC	Power management controller
	4

POR	Power-on reset
PSDU	PHY service data unit
PWM	Pulse-width modulator
RAM	Random access memory
RBW	Resolution bandwidth
RC	Resistor-capacitor
RCOSC	RC oscillator
RF	Radio frequency
RSSI	Receive signal strength indicator
RTC	Real-time clock
RX	Receive
SCK	Serial clock
SFD	Start of frame delimiter
SFR	Special function register
SHR	Synchronization header
SINAD	Signal-to-noise and distortion ratio
SPI	Serial peripheral interface
SRAM	Static random-access memory
ST	Sleep timer
T/R	Tape and reel
T/R	Transmit/receive
THD	Total harmonic distortion
TI	Texas Instruments
ТХ	Transmit
UART	Universal asynchronous receiver/transmitter
USART	Universal synchronous/asynchronous receiver/transmitter
VCO	Voltage-controlled oscillator
VGA	Variable-gain amplifier
WDT	Watchdog timer
XOSC	Crystal oscillator



Additional Information

Texas Instruments offers a wide selection of cost-effective, low-power RF solutions for proprietary and standard-based wireless applications for use in industrial and consumer applications. Our selection includes RF transceivers, RF transmitters, RF front ends and System-on-Chips as well as various software solutions for the sub-1 and 2.4-GHz frequency bands.

In addition, Texas Instruments provides a large selection of support collateral such as development tools, technical documentation, reference designs, application expertise, customer support, third-party and university programs.

The Low-Power RF E2E Online Community provides you with technical support forums, videos and blogs, and the chance to interact with fellow engineers from all over the world.

With a broad selection of product solutions, end application possibilities, and the range of technical support, Texas Instruments offers the broadest low-power RF portfolio. **We make RF easy!**

The following subsections point to where to find more information.

Topic

B.1	Texas Instruments Low-Power RF Web Site	366
B.2	Low-Power RF Online Community	366
B.3	Texas Instruments Low-Power RF Developer Network	366
B.4	Low-Power RF eNewsletter	366

Page



B.1 Texas Instruments Low-Power RF Web Site

Texas Instruments' Low-Power RF Web site has all our latest products, application and design notes, FAQ section, news and events updates, and much more. Just go to <u>www.ti.com/lprf</u>.

B.2 Low-Power RF Online Community

- Forums, videos, and blogs
- RF design help
- E2E interaction Posting one's own and reading other users' questions

Join us today at www.ti.com/lprf-forum

B.3 Texas Instruments Low-Power RF Developer Network

Texas Instruments has launched an extensive network of low-power RF development partners to help customers speed up their application development. The network consists of recommended companies, RF consultants, and independent design houses that provide a series of hardware module products and design services, including:

- RF circuit, low-power RF and ZigBee design services
- Low-power RF and ZigBee module solutions and development tools
- · RF certification services and RF circuit manufacturing

Need help with modules, engineering services or development tools?

Search the Low-Power RF Developer Network to find a suitable partner! www.ti.com/lprfnetwork

B.4 Low-Power RF eNewsletter

The Low-Power RF eNewsletter keeps you up to date on new products, news releases, developers' news, and other news and events associated with low-power RF products from TI. The Low-Power RF eNewsletter articles include links to get more online information.

Sign up today on www.ti.com/lprfnewsletter.



References

References and other useful material:

- 1. IEEE Std. 802.15.4-2006: Wireless Medium Access Control (MAC) and Physical Layer (PHY) specifications for Low Rate Wireless Personal Area Networks (LR-WPANs) http://standards.ieee.org/getieee802/download/802.15.4-2006.pdf
- 2. CC2530 Data Sheet (SWRS081)
- 3. CC2531 Data Sheet (SWRS086)
- 4. CC2533 Data Sheet (SWRS087)
- 5. CC2540 Data Sheet (SWRS084)
- 6. CC2541 Data Sheet (SWRS110)
- 7. Bluetooth® Core Technical Specification document, version 4.0 https://www.bluetooth.org/technical/specifications/adopted.htm
- 8. Universal Serial Bus Revision 2.0 specification http://www.usb.org/developers/docs/usb_20_101111.zip

Appendix C



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Revision History

Changes from B Revision (September 2010) to C Revision

Page

		_
•	Added /41 to all CC2540creating globally:CC2540/41	1
•	Added version information for CC2541	
•	Added 3rd paragraph in the section About This Manual	. 15
•	Added last column in table under Devices section	. 16
•	Added Figure 1-3	
•	Added BLE; also mention of proprietary modes	. 25
•	Added CC2541 data to table entries	. 35
•	Changed interrupt no. 0 description	. 44
•	Changed register IEN0 bit 0 description	. 46
•	Changed register TCON bit 1 description	. 48
•	Changed description for bit 4 in the Debug Status table	. 59
•	Added CC2541 to CHIPID description	. 62
•	Added CC2541 to PM3 section of Power Management Introduction	. 64
•	Added CC2541 to SRCRC register description table	. 66
•	Made SRCRC register bit 7 CC2533 only	. 66
•	Corrected part number: CC25540 \rightarrow CC2540	. 69
•	Added to the list of registers that do not retain contents in PM2 and PM3	. 72
•	Added CC2541 to last paragraph of chapter introduction	
•	Added CC2541 to end of (CC2530/CC2531/CC2540) in several places	. 75
•		. 83
•		. 86
•	Changed R/W0 to R/W for T3CTL bit 3	128
•	Changed R/W0 to R/W for T4CTL bit 3	131
•	Updated description for Watchdog Timer Control.	161
•	Included relevant parts of CC2541 Timer 2 (Radio Timer) chapter in this chapter and then deleted that one	207
•	Added CC2541 information to MAC timer description	207
•	Inserted Long Compare section for CC2541	209
•	Added two new interrupts for CC2541	209
•	Added CC2541 content, including two new events, to the Event Outputs section	210
•	Changed name of Timer 2 event output configuration register	212
•	Relegated existing T2IRQF register to CC253X/CC2540; added new T2IRQF register for CC2541	
•	Relegated existing T2IRQM register to CC253X/CC2540; added new T2IRQM register for CC2541	215
•	Relegated existing T2EVTCFG register to CC253X/CC2540; added new T2EVTCFG register for CC2541	216
•	Changed chapter title and added CC2541 information to intro paragraph	287
•	Added new chapter "CC2541 Radio)"	291

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

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